

DEC-TR
471

TECHNICAL DETAILS OF FOCAS
(Fiber Optic Cable Simulator)

DEC - TR 471

Ann B. Ewalt

96-029/024/10

March 11, 1987

DIGITAL MARLBORO CO

DEC-TR
471

*** RESTRICTED DISTRIBUTION ***

TECHNICAL DETAILS OF FOCAS
(Fiber Optic Cable Simulator)

DEC - TR 471

Ann B. Ewalt
96-029/024/110
March 11, 1987

DIGITAL MARLBORO CORPORATE LIBRARY

Keywords: FOCAS, fiber optics, optical cable, optical fiber,
fiber optic communications

ABSTRACT

This document is intended to provide the technical information necessary to support and/or modify the simulation tool FOCAS. FOCAS is used to simulate the physical behavior of fiber optic communications links. This document assumes that the reader is familiar with the "FOCAS User's Guide", DEC TR-470.

Distributed Systems Technical Office
Digital Equipment Corporation
110 Spitbrook Road
Nashua, New Hampshire 03602

(c) 1987 by DIGITAL EQUIPMENT CORPORATION

This is an unpublished work which contains Confidential and Secret information which is protected under the Copyright laws. The existence of the Copyright notice is not to be construed as an admission or presumption that publication has occurred. Unauthorized copying is strictly prohibited. All rights reserved.

CONTENTS

I.	INTRODUCTION	1
II.	BASIC NETWORK ANALYSIS ALGORITHMS	2
III.	OPTICAL FIBER MODEL & ENHANCEMENTS	5
IV.	NETWORK SUBSECTION MODELS		
	A. GEN - trapezoidal generator & overlapping 1's correction	10
	B. SINE - sine wave generator	13
	C. TEE - TEE network with series L, R and shunt C		14
	D. FIBER - fiber optic cable driven by source with Gaussian spectral distribution	15
	E. HPF - high pass filter	17
	F. EAMP - voltage dependent voltage generator	18
	G. IAMP - voltage dependent current generator	19
	H. LINE - rectangular stripline	20
	I. FILTER - response $F(w) = \{\cos wT_0\}^{*N}$	22
	J. SHUNT - shunt impedance	23
	K. SERIES - series impedance	24
	L. DELTA - standard PI circuit with series L and shunt C	25
	M. BPF - bandpass filter	26
	N. SRC - interface to user defined source data file	28
V.	FOCAS PROGRAMS		
	A. Data Flow	31
	B. Calling Structures	36
	C. Parameter Map for Array RLC	44
	D. Listings	47
VI.	ADDING A NEW NETWORK TYPE	110

VII. ENHANCEMENTS

A.	Complex Filters & Resonant Circuits	. . .	112
B.	Optical Cable Splices	. . .	113
C.	Reduction of Timing Uncertainty	. . .	113
D.	Fiber Models	. . .	114
E.	Graphics Support on MicroVax	. . .	114

VIII. APPENDICES

Appendix I.	LED/CABLE Reference Data for Figures 1 & 2	. . .	117
Appendix II.	Fiber Response Function with Constant D	. . .	119
Appendix III.	Fiber Response Function with $D = SO*(L-L_c)$. . .	120
Appendix IV.	Integrals & Mathematical Identities	. . .	123
Appendix V.	CIRCUIT_GRAPH.FOR Listing	. . .	125

IX. ACKNOWLEDGEMENTS

. 129

X. BIBLIOGRAPHY

. 130

XI. FIGURES

. 131

Figure 1. Cable 1 Rise & Fall Time Data

Figure 2. Cable 2 Rise & Fall Time Data

Figure 3. Source Gaussian Spectrum

Figure 4. Overshoot with Leading & Trailing 1's

Figure 5. Corrected Waveform with Leading & Trailing 1's

Figure 6. Overlapping 1's Correction: $PER+D_o < D1+P1$ Figure 7. Overlapping 1's Correction
($D1+P1 < PER+D_o < D1+P1+F$) .and. ($D1+P1+F > PER+D_o+R$)

INTRODUCTION

FOCAS is a SIMULATION tool for analyzing the physical behavior of point-to-point fiber optic communications links. All circuit analysis is performed in the frequency domain, then transformed into the time domain. Hence, all input waveforms are assumed to be periodic in time.

FOCAS contains a collection of network subsections (circuits) from which the simulation models are constructed. These models represent the behavior of a circuit or signal generator in the frequency domain. Three voltage sources are included: GEN, an asymmetric periodic trapezoidal wave source; SINE, a periodic sine wave source; and SRC, a digitized waveform with voltage specified at equally spaced time intervals. These source models all return the component of the source's frequency spectrum at the desired angular frequency. All of the other circuit type models (TEE, FIBER, HPF, EAMP, IAMP, LINE, FILTER, SHUNT, SERIES, DELTA, BPF) return the general circuit parameters (A, B, C, D) for the particular circuit type at the specified angular frequency. A simulation circuit is composed of a series of these cascaded two-terminal-pair networks.

Once FOCAS is installed on a system, the user creates a simulation model by simply creating the model in a file, e.g. "MODELXYZ.". At the system prompt, RUN FOCAS invokes the simulator. FOCAS prompts to the user for the simulation model name as the only input required.

Output voltages are based on terminal pair differentials. FOCAS provides output in both graphical and tabular form. The graphics files may be displayed directly on REGIS terminals. Output data files provide an interface to graphing routines for various other display devices.

FOCAS creates up to seven output files as specified by control statements:

```

MODELXYZ.TG = time domain graph in REGIS (always created)
MODELXYZ.TD = time domain data file (always created)

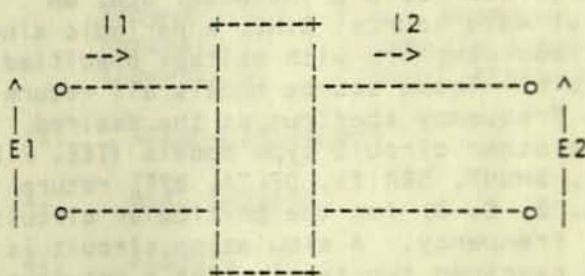
MODELXYZ.FG = frequency domain graph in REGIS (created when
              control line .FDG specified)
MODELXYZ.FD = frequency domain data file (created when
              control line .FDG specified)

MODELXYZ.PD = phase vs. frequency data file (created when
              control line .PDG specified)
MODELXYZ.PG = phase vs. frequency graph in REGIS (created
              when control line .PDG specified)

MODELXYZ.DAT = times at which output voltage waveform
               crosses specified voltage levels (created
               when there is crossover or extremum data
  
```

11. BASIC NETWORK ANALYSIS ALGORITHM

FOCAS performs all network calculations in the frequency domain, then translates the results back into the time domain by use of the fast Fourier transform (subroutine FFT842). Complex networks are broken down into simple sections for which the chain matrices can be calculated (Ref. Weinberg, pp.23-26). The parameters: A, B, C, and D in the chain matrix completely describe the response of the network subsection as shown below.



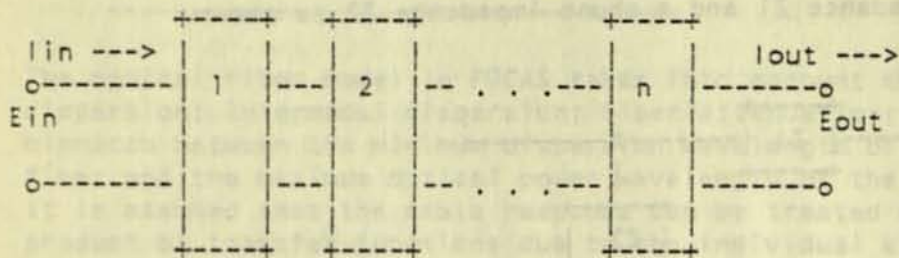
The chain matrix is defined by:

$$(1) \quad \begin{vmatrix} A & B \\ C & D \end{vmatrix} \cdot \begin{vmatrix} E2 \\ I2 \end{vmatrix} = \begin{vmatrix} E1 \\ I1 \end{vmatrix}$$

where the general circuit parameters are given by:

- (2) $A = E1/E2$ when $I2 = 0$ (open output)
- (3) $B = E1/I2$ when $E2 = 0$ (output shorted)
- (4) $C = I1/E2$ when $I2 = 0$ (open output)
- (5) $D = I1/I2$ when $E2 = 0$ (output shorted).

Chain matrices have the desirable property that the chain matrices of cascaded two-terminal-pair subsections can be multiplied together to determine the behavior of the composite network. The illustration on the top of the following page shows how this property is applied.



$$(6) \quad \begin{vmatrix} A_{tot} & B_{tot} \\ C_{tot} & D_{tot} \end{vmatrix} \cdot \begin{vmatrix} E_{out} \\ I_{out} \end{vmatrix} = \begin{vmatrix} E_{in} \\ I_{in} \end{vmatrix}$$

where:

$$(7) \quad \begin{vmatrix} A_{tot} & B_{tot} \\ C_{tot} & D_{tot} \end{vmatrix} = \begin{vmatrix} A_1 & B_1 \\ C_1 & D_1 \end{vmatrix} \begin{vmatrix} A_2 & B_2 \\ C_2 & D_2 \end{vmatrix} \dots \begin{vmatrix} A_n & B_n \\ C_n & D_n \end{vmatrix}$$

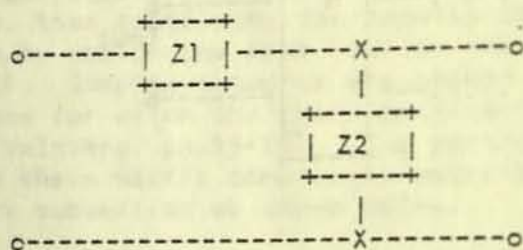
A more useful form of these equations relates the output voltage and current to the input voltage and current:

$$(8) \quad \begin{vmatrix} D_{tot} & B_{tot} \\ C_{tot} & A_{tot} \end{vmatrix} \cdot \begin{vmatrix} E_{in} \\ -I_{in} \end{vmatrix} = \begin{vmatrix} E_{out} \\ -I_{out} \end{vmatrix}$$

It should be noted that the term "chain matrix" refers to the form given in (1):

$$\begin{vmatrix} A & B \\ C & D \end{vmatrix}$$

As an illustration, consider a network comprised of a series impedance Z_1 and a shunt impedance Z_2 as shown below:



This configuration is described by the product of two chain matrices: $M_1 * M_2$ where M_1 represents the series impedance and M_2 the shunt impedance.

$$(9) \quad M_1 = \begin{vmatrix} 1 & Z_1 \\ 0 & 1 \end{vmatrix} \quad \text{and} \quad M_2 = \begin{vmatrix} 1 & 0 \\ 1/Z_2 & 1 \end{vmatrix}$$

If Z_1 represents a Capacitor, C , and Z_2 a resistor, R , the network is a simple high pass filter. The individual chain matrices are:

$$M_1 = \begin{vmatrix} 1 & 1/sC \\ 0 & 1 \end{vmatrix} = \begin{vmatrix} 1 & -j/\omega C \\ 0 & 1 \end{vmatrix}; \quad s = j\omega$$

and

$$M_2 = \begin{vmatrix} 1 & 0 \\ 1/R & 1 \end{vmatrix}$$

The chain matrix for the network, M_{tot} , is:

$$M_{tot} = \begin{vmatrix} (1+Z_1/Z_2) & Z_1 \\ 1/Z_2 & 1 \end{vmatrix} = \begin{vmatrix} (1 + 1/sRC) & 1/sC \\ 1/R & 1 \end{vmatrix}$$

This is precisely the result shown for the HPF (Section IV) with the inductance, L , equal to zero.

III. OPTICAL FIBER MODELS & ENHANCEMENTS

The optical fiber model in FOCAS takes into account chromatic dispersion; intermodal dispersion; fiber attenuation; and the mismatch between the minimum dispersion wavelength of the fiber and the maximum optical power wavelength of the source. It is assumed that the cable response can be treated as the product of transfer functions due to the individual effects:

$$(10) \quad G_{\text{tot}}(w, L) = P_{\text{LED}}(w, L) * G_c(w, L) * G_{\text{IM}}(w) * G_a(w)$$

$$(11) \quad G_{\text{tot}}(w, L) = \text{cable response at angular frequency } w \text{ and wavelength } L$$

$$(12) \quad P_{\text{LED}}(w, L) = \text{optical source output power at angular frequency } w \text{ and wavelength } L$$

$$(13) \quad G_c(w, L) = \text{fiber chromatic dispersion transfer function at angular frequency } w \text{ and wavelength } L$$

$$(14) \quad G_{\text{IM}}(w) = \text{fiber intermodal dispersion transfer function at angular frequency } w$$

$$(15) \quad G_a(w) = \text{fiber attenuation transfer function at angular frequency } w$$

The fiber optic cable is modelled by a single network subsection with characteristics of both the optical fiber and the optical source. Mathematical details are given in Appendices II & III. Each of the individual physical effects is discussed separately below.

A. SOURCE SPECTRUM

The normalized optical transmitter power spectrum is approximated by a Gaussian wavelength distribution.

$$(16) \quad P_{\text{LED}}(w, L) = \frac{1}{Z * \text{SQRT}(\text{PI})} \exp\{ - (L - L_c)^2 / Z^2 \} \quad \text{where}$$

$$(17) \quad Z = F / [2 * \text{sqrt}(\ln 2)] = \text{standard deviation of the source spectrum}$$

F = width of the spectral source at half power in nanometers

B. CHROMATIC DISPERSION

The FIBER model approximates the dispersion coefficient of the fiber with a constant based on the Sellmeier equation. The dispersion coefficient, D in ns/(nm-km), is given by:

$$(18) \quad D = 1.176 * \text{SQRT} \left[D_o^{**2} + \frac{(0.85 * S_o * \text{FWHM})^{**2}}{8} \right]$$

where one has:

L_o = minimum dispersion wavelength of the fiber

L_c = peak power wavelength of the source

S_o = slope of the dispersion curve in ns/(nm**2-km)

FWHM = source half max. spectral width

$$(19) \quad D_o = \frac{S_o}{4} * \left[\frac{L_c - L_o^{**4}}{L_c^{**3}} \right]$$

The expression in equation (19) is the classical representation of the Sellmeier equation. Equation (18) was suggested by Siecor's Dieter Schickentanz and his treatment of chromatic dispersion, then modified by empirical data. The value of D given by equation (18) can be thought of as an average dispersion value weighted by the source's spectral distribution.

An experiment was run in which seven different LED's were used to excite two different optical fiber cable plants, and the 10-90% rise and fall times of the output waveforms compared with predictions. Appendix 1 contains the specifications of the cables and LED's used in this experiment. Figures 1 and 2 show the results.

The SIECOR points in the Figures correspond to the values predicted using Schickentanz's dispersion and bandwidth summing equations. These relationships assume that all fiber frequency responses follow Gaussian distributions and can be summed accordingly:

$$(20) \quad \frac{1}{\text{Bwtot}^{**2}} = \frac{1}{\text{BW1}^{**2}} + \dots + \frac{1}{\text{Bwn}^{**2}}$$

SIMER points use the same cable model as FOCAS, with the exception that the value of D is not multiplied by the factor of 1.176 as in equation (18).

It is evident from the Figures that the FOCAS model provided the best prediction of the output rise and fall times of the models available. However, the discrepancies with the measured values are of some concern. Subsequent review of the data showed some lack of repeatability in the measurements due to changes in the fiber pigtails through which measurements were made. Plans have been made to repeat these tests and perform others.

In addition, the simulation runs all assumed simple rounded trapezoidal edge shapes without ringing for the LED's, while observation showed that, particularly for those LED's with large differences between predicted and measured results, the transistors actually had a good deal of ringing. There are presently experiments being run to determine the importance of exactly modelling source turnon and turnoff characteristics.

It is anticipated that the chromatic dispersion model in FOCAS will undergo further refinement as more data become available. Appendix III contains the derivation of an alternate optical fiber model which includes first order dependence of dispersion on wavelength:

$$(21) \quad D(L) = D(L_0) + D'(L) * (L - L_0) + D''(L) * [(L - L_0) ** 2] / 2 + \dots$$

which reduces to:

$$(22) \quad D(L) = S_0 * (L - L_0)$$

because dispersion varies linearly with wavelength near L_0 , and L_0 is defined as the wavelength at which $D(L_0)$ is zero and $D'(L_0)$ is equal to S_0 (Ref. 2).

Modifications which simply require that a different expression be used to compute D can be made directly in SUBROUTINE PRECALC. The dispersion calculation is labelled in the calculations for the Type 7 Circuit. More complex models, showing the explicit wavelength dependence of D can be added as new network types. Appendix IV contains basic integrals and identities to assist in future modifications to the fiber model(s).

The effect of chromatic dispersion is to delay wavelength component L by an amount, T , proportional to fiber length, X . This delay results in a phase shift as shown in equation (24).

$$(23) \quad T = D * (L - L_c) * (X ** Q_c) = \text{delay introduced to spectral component at wavelength } L \text{ with respect to } L_c$$

$$Q_c = \text{chromatic length scale factor}$$

$$(24) \quad G_c(w) = \exp\{jw * T\} = \text{phase shift introduced to spectral component at wavelength } L \text{ with respect to a component at the central wavelength, } L_c$$

$$= \exp\{jw * D * (L - L_c) * (X ** Q_c)\}$$

C. FIBER ATTENUATION

The fiber attenuation is assumed to be exponential with effective fiber length, $X^{**}Q_a$

$$(25) \quad G_a(w) = \exp\{-K*A\} = \text{fiber transfer function due to attenuation}$$

$$(26) \quad A = [r_o + r_w*(L-L_c)]*(X^{**}Q_a) = \text{fiber attenuation in dB}$$

$$(27) \quad K = (\ln 10)/10 = 0.2306 = \text{dB conversion factor}$$

r_o = DC optical power attenuation of the fiber in dB/km

r_w = wavelength dependent optical power attenuation of the fiber in dB/(nm-km)

Q_a = attenuation length scale factor, assumed equal to Q_c (chromatic length scale factor) in present FIBER model, but not in alternate model, Appendix III

D. INTERMODAL DISPERSION

The intermodal dispersion of the fiber is described by a Gaussian distribution of effective optical path lengths, ($X_{eff} = X^{**}Q_m$).

$$(28) \quad G_{IM}(w) = \exp\{-(d^{**}2)\} \quad \text{where}$$

$$(29) \quad d = (0.1325 * w * X_{eff})/F_1$$

$$(30) \quad B_m = F_1/X_{eff} = \text{modal bandwidth of the fiber in Gigahertz}$$

F_1 = cable bandwidth due to intermodal dispersion in GHz-km

$X_{eff} = X^{**}Q_m$ = effective modal cable length

X = fiber physical length in kilometers

Q_m = intermodal length scale factor

E. FIBER RESPONSE

Combining all of the physical effects, one obtains the fiber response function, $Pe(w)$:

$$(31) \quad Pe(w) = \int_{-\infty}^{\infty} G_{tot}(w,L) dL \quad \text{with } G_{tot}(w,L) \text{ defined in (10).}$$

Detailed evaluation of integral (31) is given in Appendix II for the case in which equation (18) is used for dispersion; Appendix III for the case in which equation (22) is used for dispersion.

The general circuit parameters for a two terminal pair fiber model are then given by:

$$(32) \quad A = D = 1 / Pe(w)$$

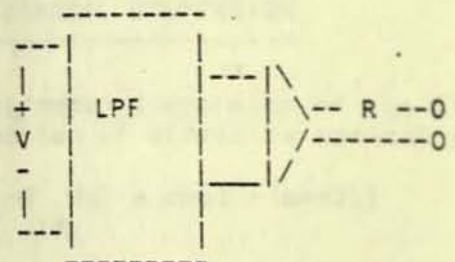
$$(33) \quad B = C = 0.$$

NETWORK SUBSECTION MODELS

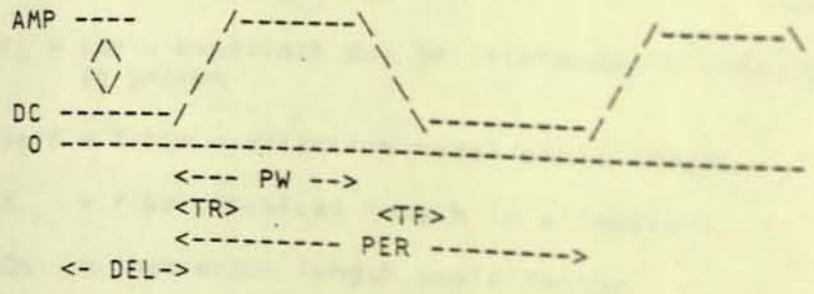
The network subsection types are as follows:

- GEN - A trapezoidal waveform voltage generator
- SINE - A sinewave generator
- EAMP - Voltage dependent voltage generator
- IAMP - Voltage dependent current generator
- TEE - The standard T circuit with series L and shunt C
- DELTA - The standard PI circuit with series L and shunt C
- LINE - A transmission line, skin effect for stripline
- FIBER - Fiber optic cable dispersion model
- HPF - High Pass Filter
- XFORMER - Transformer
- FILTER - Cosine frequency domain filter
- SHUNT - Polynomial expression for a shunt impedance
- SERIES - Polynomial expression for a series impedance
- BPF - Bandpass filter
- SRC - User defined input waveform

A. GEN Network



The GEN is a periodic trapezoidal waveform generator followed by low pass filter and a unity voltage gain buffer with an output impedance equal to R.



Parameters:

- PW = "ideal" pulse width, i.e. the time between the decisions to turn on and turn off
- TR = 10-90 % risetime (leading edge) in nanoseconds.
- TF = 10-90 % falltime (trailing edge) in nanoseconds. (default=TR)
- PER = pulse repetition period in nanoseconds
- AMP = pulse amplitude in volts, defaults to one.
- DEL = pulse delay in nanoseconds, measured from the leading edge. Defaults to zero.
- BW = filter bandwidth in gigahertz, defaults to 50 divided by the risetime.
- DC = DC voltage offset in volts, defaults to zero
- R = generator output resistance in ohms, defaults to zero.

If several GEN sections are cascaded together to form a data stream, FOCAS automatically sets some conditions to simplify user input. FOCAS compares the risetimes of the first two GEN sections. If they are not equal, and the second is nonzero (i.e. not to be set equal to the first), a dual slope leading edge is assumed. If they are equal, or if the second risetime is not specified (i.e. initially zero), a single sloped leading edge is assumed. For the single slope case, TR, TF, and AMP are all set to the values specified for the first GEN. For the dual slope case, all odd numbered GEN's are assigned the TR, TF, and AMP values of the first GEN; all even numbered GEN's receive the values specified by the user for the second GEN.

In all cases, only the first GEN need specify PER and BW. Whenever any of the input parameters is missing from a subsequent GEN specification, FOCAS automatically determines whether to use the single or dual slope model.

Similarly, FOCAS automatically checks if the first and last bits in a periodic data stream are one's. If there are potentially overlapping one's, i.e., the optical transmitter does not have time to completely turn off before receiving another signal to turn on, FOCAS will combine the first and last GEN's while preserving the data integrity, without driving the transmitter output beyond its maximum level.

Without this "overlapping 1's" checking, nonphysical overshoot behavior as shown in Figure 4 would occur. Subroutine PRECALC performs the overlapping 1's checking and correction to produce the waveform shown in Figure 5 by comparing the last GEN's output pulse with the second repetition of the pulse generated by the first GEN in the string.

The correction algorithm works by checking for the conditions which result in overshoot:

$$(34) \quad \text{PER} + \text{Do} < \text{D1} + \text{P1} \quad (\text{See Figure 6})$$

-

or

$$(35) \quad (\text{D1} + \text{P1} < \text{PER} + \text{Do} < \text{D1} + \text{P1} + \text{F}) \text{ .and. } (\text{D1} + \text{P1} + \text{F} > \text{PER} + \text{Do} + \text{R})$$

(See Figure 7)

where: D1 = pulse delay of the last (NGEN-th) GEN of the input simulation model

P1 = ideal pulse width of the last GEN of th input model

Do = pulse delay of the first GEN of the input model

Po = ideal pulse width of the first GEN of the input model

PER = period of the data stream

R = 0-100% rise time of the source (same for all GEN's)

F = 0-100% fall time of the source (same for all GEN's)

When either of these conditions is encountered, subroutine PRECALC modifies the input simulation model by replacing the last GEN with a GEN with corrected pulse width, PL, and then dropping the first GEN. The corrected pulse width is calculated to combine the input bits represented by the original model's first and last GEN's.

$$(36) \quad \text{PL} = \text{Po} + \text{CORR}$$

$$(37) \quad \text{CORR} = |(\text{PER} + \text{Do} - \text{D1})|$$

Invoking the overlapping 1's correction results in a model with (NGEN-1) generators, where NGEN is the number of GEN's in the original simulation model. Figure 6 illustrates the original and corrected GEN pulses corresponding to condition (34). Figure 7 shows the original and corrected pulses corresponding to condition (35).

Checking only occurs between the first and last GEN sections. It is assumed that the user has exercised care in defining all internal GEN sections to eliminate the potential for signals that overshoot the maximum HIGH level.

B. SINE Network

SINE is sinewave generator which in the frequency domain is a delta function at the frequency f_0 . It has an output impedance R . In the time domain:

$$F(t) = \text{AMP} * \text{SIN} (2 \text{ PI } F t + \text{PHI}) + \text{DC}$$

Parameters:

AMP = Peak sinewave amplitude, defaults to one.

F = sinewave frequency in GHz, must be an exact multiple of the simulation reciprocal period.

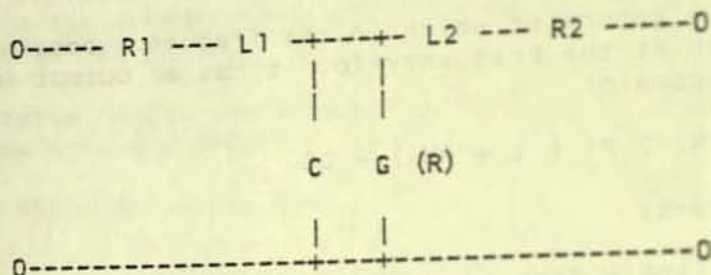
DEL = delay in nanoseconds, defaults to zero.
(PHI = $-W * \text{DEL}$)

DC = dc voltage offset in volts, defaults to zero.

R = generator output resistance in ohms, defaults to zero.

There are no general circuit parameters for SINE. As a source generator, it is represented by its frequency spectrum.

C. TEE Network



Parameters:

R_1, R_2 = series resistances in ohms, default to zero. Will have a skin effect component if F_1 or F_2 are specified, respectively.

L_1, L_2 = series inductances in nanohenries, default to zero. Will have an internal inductance component if F_1 or F_2 are specified respectively.

C = shunt capacitance in nanofarads, defaults to zero.

$G (R)$ = shunt conductance in mhos, defaults to zero. Either G or R (in ohms) may be specified. If both are specified by mistake, the R value supercedes the G value.

F_1, F_2 = The skin effect break frequencies in Gigahertz; Defined as the frequency at which the skin effect resistance is equal to the dc resistance. Skin effect resistance and internal inductance are assumed to be square-root frequency dependent. Default to infinity (no skin effect).

The skin effect model is:

$$R_{ac} = \text{conductor resistance} = R_1 \text{ SQRT} \left(1 + \left(\frac{F}{F_1} \right)^2 \right)^{1/4}$$

for conductor 1 (R_1, L_1) where F is the frequency.

$$X_{int} = \text{internal conductor inductive reactance} = R_{ac} - R_1$$

for conductor 1. Appropriate substitutions are made for conductor 2 (R_2, L_2).

D. FIBER Network

FIBER models the chromatic and intermodal dispersion, and attenuation losses in an optical cable. The model assumes intermodal dispersion based on a gaussian distribution of path lengths and chromatic dispersion based on a source with a gaussian spectral density and a linear material dispersion coefficient for the cable. The fiber cable delay is not included. The frequency response, $P(w)$, of the cable is given by:

$$P(w) = e^{-a + b w^2 - c w^2 - d w - j 2 b c w}$$

$a = K A_0 X^{Q_c}$ = Static attenuation term

$b = K K' A_w \text{FWHM} X^{Q_c}$ = Wavelength dependent attenuation

$c = w K' \text{FWHM} D X^{Q_c}$ = Wavelength dependent time delay

$d = 0.1325 w X^{Q_m} / F1$ = Intermodal dispersion

$$K = \ln(10)/10$$

$$K' = 1./ (4 \text{ SQRT}(\ln 2))$$

Parameters:

A_0 = DC optical power attenuation of cable in db/km
Defaults to zero. (Same as r_p in models, section III)

A_w = wavelength dependent optical power attenuation in db/(nanometer-kilometer). Defaults to zero. Note that A_w is positive when attenuation increases with wavelength. (Same as r_w in models, Section III)

FWHM = optical source half power full bandwidth in nanometers. Defaults to zero.

LC = optical source center wavelength in nanometers; used in calculating D when LC specified with L0 and S0

L0 = cable minimum dispersion wavelength in nm.; used in calculating D when LC , L0 and S0 specified

S0 = slope of cable dispersion curve at L0 in nanoseconds per (nanometer**2)-kilometer; used in calculating D when S0 , LC , and L0 specified

FIBER NETWORK (cont.)

FI = cable bandwidth in GHz-Km due to intermodal dispersion. Defaults to infinity.

X = cable length in Km. Defaults to zero.

Qm = Intermodal length scale factor (default=1)

Qc = Chromatic length scale factor (default=1)

D = fiber material dispersion coefficient in nanoseconds per nanometer-kilometer; if D is zero or not specified, it is calculated from input parameters LC, L0, S0 using the equations below:

$$D = 1.176 * \left\{ \text{SQRT} \left[D_0^{**2} + \frac{(0.85 * S_0 * \text{FWHM})^{**2}}{8} \right] \right\}$$

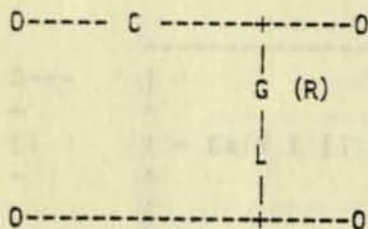
$$D_0 = \frac{S_0}{4} \left(LC - \frac{L_0^{**4}}{LC^{**3}} \right) \quad \text{Sellmeir's Equation}$$

The chain matrix parameters for the FIBER model are given by:

$$A = 1/P(w) \quad B = 0 \quad C = 0 \quad D = 1/P(w)$$

E. HPF Network

A high pass filter or ac coupler.



Parameters:

C = capacitance in nanofarad, must be present.

G = shunt conductance in mhos, either G or R must be specified.

R = shunt resistance in ohms, if present it overrides an input G value and $G = 1/R$.

L = shunt inductance in nanohenries, defaults to zero.

The general circuit parameters for the HPF are given by:

$$A = A_r + j * A_i$$

$$A_r = 1 - \frac{L}{C [R^{**2} + (wL)^{**2}]}$$

$$A_i = \frac{-R}{wC [R^{**2} + (wL)^{**2}]}$$

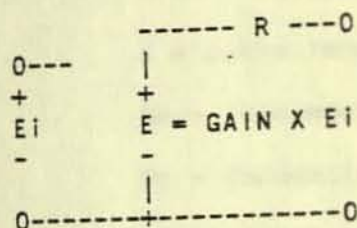
$$B = -j / (wC)$$

$$C = 1 / (R + wL)$$

$$D = 1$$

F. EAMP Network

EAMP is a voltage dependent voltage generator with a specified bandwidth and output resistance.



Parameters:

GAIN = voltage gain, defaults to one

BW = filter bandwidth in gigahertz, defaults to 50 divided by the minimum risetime.

R = generator output resistance in ohms, defaults to zero.

The filter does not change the infinite input impedance of IAMP and has the low pass filter frequency domain response:

$$F(\omega) = \frac{1}{1 + j \omega / (2 \text{ PI } \text{ BW})}$$

The general circuit parameters are given by the following relations:

$$\omega_0 = \text{PI}2 * \text{BW} \quad (\text{PI}2 = 6.28 \dots)$$

$$d = \frac{1}{\text{GAIN}} * [1 + j * \omega / \omega_0]$$

$$A = d$$

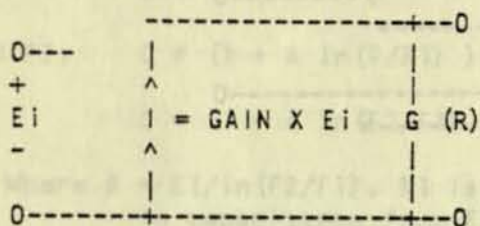
$$B = R * d = \frac{R}{\text{GAIN}} * [1 + j * \omega / \omega_0]$$

$$C = (d/R) * 10.D-9 = \frac{10^{**(-9)}}{R * \text{GAIN}} * [1 + j * \omega / \omega_0]$$

$$D = d * 10^{**9} = \frac{10^{**(-9)}}{\text{GAIN}} * [1 + j * \omega / \omega_0]$$

G. IAMP Network

IAMP is a voltage dependent current generator with a specified bandwidth and output conductance.



Parameters:

GAIN = current gain in mhos, defaults to one

BW = filter bandwidth in gigahertz, defaults to 50 divided by the minimum risetime.

G = generator output conductance in mhos, defaults to zero.

R = optional output resistance in ohms, if present
 $G = 1/R$; overriding an input G value.

The filter does not change the infinite input impedance of IAMP and has the low pass filter frequency domain response:

$$F(\omega) = \frac{1}{1 + j \omega / (2 \pi \text{ BW})}$$

The general circuit parameters are given by the relationships:

$$\omega_0 = \pi \cdot 2 \cdot \text{BW} \quad (\pi \cdot 2 = 6.28 \dots)$$

$$d = \frac{1}{\text{GAIN}} * [1 + j * \omega / \omega_0]$$

$$A = G * d$$

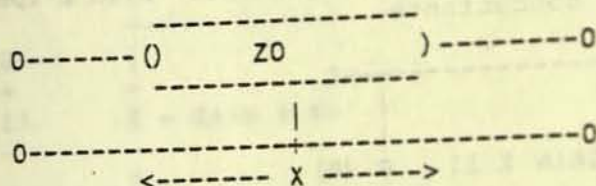
$$B = d$$

$$C = G * d * (10^{** -6})$$

$$D = d * (10^{** -6})$$

H. LINE Network

Transmission line model for a rectangular stripline of length X and characteristic impedance Z_0 .



Parameters:

Z_0 = characteristic impedance in ohms

TD = propagation delay in nanoseconds per unit X

X = line length

R = series line resistance in ohms per unit X .
Defaults to zero.

TC = stripline conductor thickness in cm, used to calculate skin effect. Defaults to zero; no skin effect.

WC = stripline conductor width in cm, this optional parameter will calculate R in ohms/cm. If present, will over-ride an input value for R . Skin effect assumes $WC \gg TC$ and a symmetric stripline.

Q = relative conductor resistivity compared to copper. Defaults to one.

D = dielectric loss tangent, assumed to be independent of frequency, i.e., the real part of the shunt impedance has the same frequency dependence as the imaginary part (capacitive reactance).

$F1, F2, F3, K1, K2$ = generate a variation in line capacitance as a function of frequency.
Default to zero.

Conductor losses will not be calculated if both R and WC are zero or absent.

The frequency variable capacitance is A piece-wise linear model with log frequency as follows:

The value of line C is computed from the input values for Z0 and TD, and the capacitance per unit length is:

$$\begin{aligned}
 F < F_1, & \quad C \\
 F_1 < F < F_2, & \quad C * (1 + A \ln(F/F_1)) \\
 F_2 < F & \quad C * (1 + A \ln(F_2/F_1) + B \ln(F/F_2))
 \end{aligned}$$

Where $A = K_1/\ln(F_2/F_1)$, K_1 is the fractional change in capacitance from F_1 to F_2

$B = K_2/\ln(F_3/F_2)$, K_2 is the fractional change in capacitance from F_2 to F_3

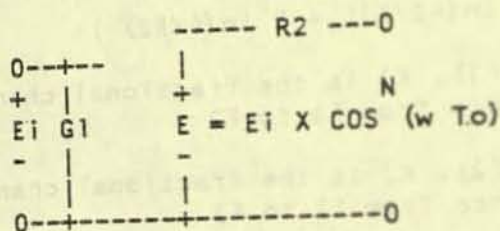
The chain matrix parameters are:

$$\begin{aligned}
 A &= 1 & B &= Z \\
 C &= 1/Z & D &= 1
 \end{aligned}$$

I. FILTER Network

Computes the response: $F(\omega) = \cos^N(\omega T_0)$ if F_0 not specified
and

$$F(\omega) = \cos^N(\omega/4 * F_0) \text{ if } F_0 \text{ is specified}$$



Parameters:

$G1$ = Shunt input conductance, mhos. Defaults to zero.

$R1$ = Shunt input resistance, ohms. If specified, replaces $G1$ where $G1 = 1/R1$.

$R2$ = Series output resistance, ohms. Defaults to zero.

T_0 = Function period, nanoseconds. Must be present.

N = Power for raised cosine, defaults to 1.

F_0 = Location of first zero in response in GHz. If present, overrides T_0 , where:

$$T_0 = 1 / (4 F_0).$$

The chain matrix parameters are given by:

$$A = 1/F(\omega)$$

$$B = R2 / F(\omega)$$

$$C = G1 / F(\omega)$$

$$D = 0$$

J. SHUNT Network

Provides a shunt impedance expressed as a polynomial:

$$Z = K \frac{A_0 + A_1*S + A_2*S^2 + A_3*S^3 + A_4*S^4}{B_0 + B_1*S + B_2*S^2 + B_3*S^3 + B_4*S^4}$$

Where: S = j w; the complex radian frequency

Units must be consistent, and a scaling in the frequency domain will scale in the output time domain.

To avoid dividing by zero, the real part of the numerator

is not allowed (by the program) to be less than 10⁻¹⁶.

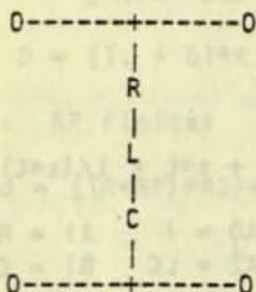
In addition, if none of the Bi coefficients are specified, B0 is automatically set equal to one.

Input parameters are:

A0, A1, A2, A3, A4, B0, B1, B2, B3, and B4 as the coefficients described above. Each defaults to zero (with the exception of B0 as described above).

K = scaling parameter, defaults to 1

EXAMPLE:



$$Z = R + s*L + 1/(s*C)$$

$$A_0 = 1$$

$$A_1 = RC$$

$$A_2 = LC$$

The chain matrix parameters are:

$$A = 1$$

$$B = 0$$

$$C = 1/Z$$

$$D = 1$$

K. SERIES Network

Provides a series impedance expressed as a polynomial:

$$Z = K \frac{A_0 + A_1*S + A_2*S^2 + A_3*S^3 + A_4*S^4}{B_0 + B_1*S + B_2*S^2 + B_3*S^3 + B_4*S^4}$$

Where: $S = j \omega$; the complex radian frequency

Units must be consistent, and a scaling in the frequency domain will scale in the output time domain.

To avoid dividing by zero, the real part of the numerator is not allowed (by the program) to be less than 10^{-16} . In addition, if none of the B_i coefficients are specified, B_0 is automatically set equal to one.

Input parameters are: $A_0, A_1, A_2, A_3, A_4, B_0, B_1, B_2, B_3, B_4,$ and K .

$A_0, A_1, A_2, A_3, A_4, B_0, B_1, B_2, B_3,$ and B_4 as the coefficients described above. Each defaults to zero (with the exception of B_0 as described above).

K = scaling parameter, defaults to 1

EXAMPLE:

0---- R -- L -- C ----0

0-----0

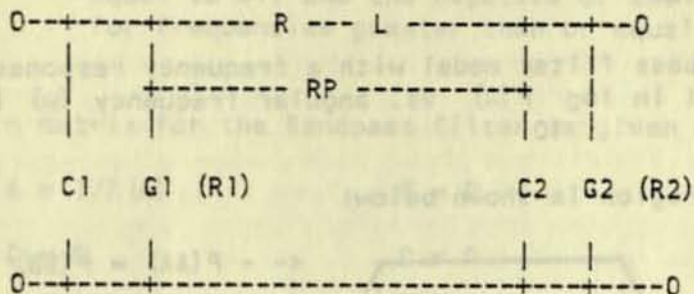
$$Z = R + s*L + 1/(s*C)$$

$$\begin{matrix} A_0 = 1 & A_1 = RC \\ A_2 = LC & B_1 = C \end{matrix}$$

The chain matrix parameters are:

$$\begin{matrix} A = 1 & B = Z \\ C = 0 & D = 1 \end{matrix}$$

L. DELTA Network



Parameters:

R = series resistance in ohms, defaults to zero.

L = series inductance in nanohenries, defaults to zero.

RP = parallel series resistance in ohms, defaults to infinity.

C1, C2 = shunt capacitances in nanofarads, default to zero.

G1 (R1), G2, (R2) = shunt conductances in mhos, default to zero. Either G or R (in ohms) may be specified. If both are specified by mistake, the R value supersedes the G value.

The general circuit parameters are given by the following relationships.

Case I: RP defaults to infinity:

$$A = \{1. + G2 * R - (w ** 2) * C2 * L\} + j * \{wL * G2 + wR * C2\}$$

$$B = R + j * wL$$

$$C = [G1 + G2 + R * \{G1 * G2 - C1 * C2 * (w ** 2)\} - XL * \{wG1 * C2 + wG2 * C1\}] + j * \{wC1 + wC2 + R * (wG1 * C2 + wG2 * C1) + wL * (G1 * G2 - C1 * C2 * (w ** 2))\}$$

$$D = \{1. + G1 * R - (w ** 2) * C1 * L\} + j * \{wL * G1 + wR * C1\}$$

Case II: RP finite:

Let $d = \{(R + RP) ** 2\} + \{(w *) L ** 2\}$; $a = RP * (R * (R + RP) + wL ** 2) / d$; and

$b = wL * RP ** 2 / d$, then:

$$A = (1. + G2 * a - w * b * C2) + j * (G2 * b + w * a * C2)$$

$$B = a + j * b$$

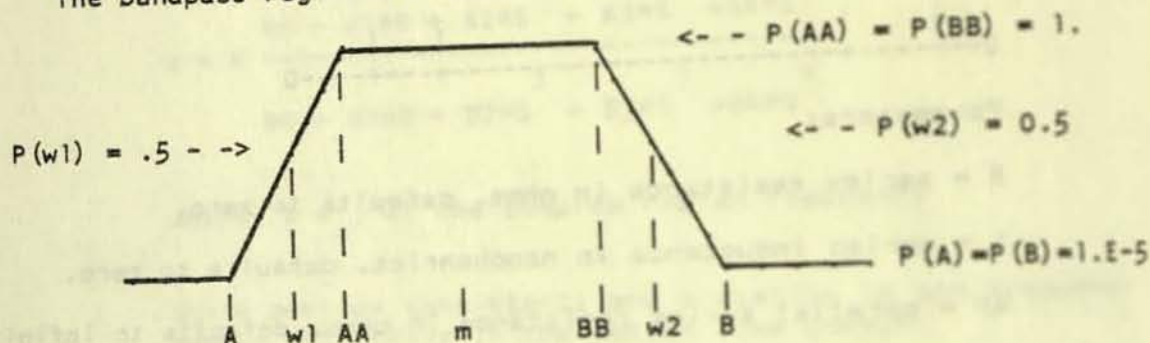
$$C = \{G1 + G2 + a * (G1 * G2 - C1 * C2 * (w ** 2)) - b * (wG1 * C2 + wG2 * C1)\} + j * \{wC1 + wC2 + a * (wG1 * C2 + wG2 * C1) + b * \{G1 * G2 - C1 * C2 * (w ** 2)\}\}$$

$$D = \{1. + G1 * a - w * b * C1\} + j * \{G1 * b + w * a * C1\}$$

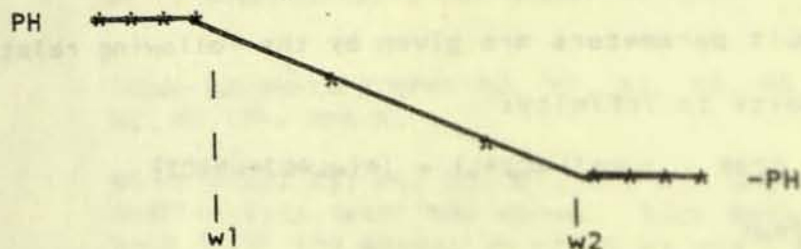
M. BPF Network

BPF is a bandpass filter model with a frequency response which is trapezoidal in $\log_{10} P(\omega)$ vs. angular frequency (ω) in octaves.

The bandpass region is shown below:



The phase shift as a function of frequency is a constant, PH, for $\omega < w_1$, and constant equal to $-PH$ for $\omega > w_2$. Between w_1 and w_2 the phase shift is linear with frequency, and zero at the midpoint, m , between w_1 and w_2 :



Input parameters are:

- F1 = frequency in Gigahertz at the low frequency 3 dB point of the trapezoid (corresponding to w_1)
- F2 = frequency in Gigahertz at the high frequency 3 dB point of the trapezoid (corresponding to w_2)
- S1 = slope of the low frequency edge of the trapezoid in dB/octave (default = 6 dB/octave)
- S2 = slope of the high frequency edge of the trapezoid in dB/octave (default = 6 dB/octave)
- K = gain (default = 1.)

PH = constant phase shift for frequencies less than or equal to w_1 , and the negative of the phase shift for frequencies greater than or equal to w_2

The chain matrix for the Bandpass filter is given by:

$$\begin{aligned} A &= 1/F(w) & B &= 0 \\ C &= 0 & D &= 0 \end{aligned}$$

The transfer function, $F(w)$, for the bandpass filter is

$$F(w) = P(w) \cdot \exp\{j \cdot \text{PHI}\}$$

where $P(w)$ is given by:

$$P(w) = \begin{cases} 0 & \text{for } w < A \text{ or } w > B \\ 1 & \text{for } AA < w < BB \\ 0.5 * \left\{ 10^{s_1} * \left[\frac{w}{w_1} * \left(\frac{w}{w_1} - 1 \right) \right] \right\} & \text{for } A < w < AA \\ 10^{s_2} * \left\{ - \frac{w}{BB} * \left(\frac{w}{BB} - 1 \right) \right\} & \text{for } BB < w < B \end{cases}$$

and the phase shift, PHI, is given by:

$$\text{PHI} = \begin{cases} \text{PH} & w < w_1 \\ P_s * \{w - m\} & w_1 < w < w_2 \\ -\text{PH} & w > w_2 \end{cases}$$

The values of A and B are found by setting $P(A) = 10^{s_1 - 4}$ and $P(B) = 10^{s_2 - 4}$.

$$AA = w_1 * (1 + 3/s_1)$$

$$BB = w_2 / (1 + 3/s_2)$$

$$A = w_1 * \{1 - 46.99/s_1\}$$

$$B = BB * \{(50/s_2) + 1\}$$

N. SRC Network

SRC is not a true network section in the sense that it models circuit behavior. Rather, it provides the interface between FOCAS, and the user defined file, SOURCE.DTA which contains a digitized input data stream of interest. The purpose of providing this interface is to allow the user to investigate complex edge behavior (turnon and turnoff ringing) of real optical sources.

SOURCE.DTA may contain either the time or frequency domain representation of the input waveform. The number of points, NT, in SDOURCE.DTA must be a power of 2, less than or equal to 8192. For time domain data, the input format is FORTRAN F20.3. For frequency domain data, it is FORTRAN ZE16.2. (See User's Guide, section 11-A-3 for examples.)

In addition to requiring that a data file named SOURCE.DTA be in the directory, the explicit input parameters are:

- NT = number of points in SOURCE.DTA; must be a power of two, less than or equal to 8192
- DT = time interval between consecutive data points (ns) or angular frequency interval between consecutive points in Gigaradians/sec
- PER = period of waveform in ns
- TYPE = 1 if SOURCE.DTA contains frequency domain representation
2 if SOURCE.DTA contains time domain representation
- R = source output resistance in ohms, defaults to zero

FOCAS consists of two FORTRAN routines: FOCAS.FOR and F2784.F.
A 30,000-page page file must be prepared to link FOCAS.

4) DATA FLOW

FOCAS_ANALYZE is the analyzer of FOCAS. It calls all
of the subroutines to read and interpret the simulation
model, perform the network analysis, and write the
output files.

a) SIMULATION MODEL DESCRIBED BY DATA STRUCTURES:

V. FOCAS PROGRAMS

Data about the network subcircuits, circuit parameter values
is passed in the COMMON block RLC, which is created in subroutine
INPUT. In addition to setting up circuit parameter matrix RLC,
INPUT calls subroutine CONTROL, which interprets the control
statements in the simulation model, generates output pointers
NPTS, NFF, and NPF, and defines structures in control of the
voltage level, crossing calculations as described below.

INPUT writes the simulation model's filename to \$FILE,
and generates names for the output files based on \$FILE.
For a model in file MODELAYZ, the other files are:

\$FILE + MODELAYZ1.TM = time domain graph output file (RECIS)
\$FILE + MODELAYZ2.TM = ASCII time domain output data file
\$FILE + MODELAYZ3.TM = power spectral graph output file (RECIS)
\$FILE + MODELAYZ4.TM = ASCII power spectral data file
\$FILE + MODELAYZ5.TM = phase angle graph output file (RECIS)
\$FILE + MODELAYZ6.TM = ASCII phase angle data file
\$FILE + MODELAYZ7.TM = ASCII file with voltage crossing times
and voltage times and values; timing jitter

INPUT also creates the vector XMB, which is a pointer to the
circuit type for each of the network subcircuits. FORD(N) is
used to define the network type for the N-th subcircuit:

```
XMB(N) = 1 -> RLC, etc. (see Parameter Map for RLC for  
further detail) - Next section)
```

The period of the waveform, PER, is also passed in COMMON,
and is taken to be the maximum of the periods of all the
generators in the model.

The generator output impedance is taken to be the sum, RSTN,
of the internal resistances of all the generators in the
simulation model, and is passed in COMMON.

05 2000

01 2000

02 2000 ... the number that ... provides the ... data ... of interest ... allow the ... and ...

03 2000 ... frequency ... number of ... data ...

04 2000 ... data ...

05 2000 ... data ...

06 2000 ... data ...

07 2000 ... data ...

08 2000 ... data ...

09 2000 ... data ...

FOCAS consists of two FORTRAN routines: FOCAS.FOR and FFT842.
A 20,000-page page file quota is required to link FOCAS.

A. DATA FLOW

FOCAS_ANALYZE is the skeleton of FOCAS. It calls all of the subroutines to read and interpret the simulation model, perform the network analysis, and write the output files.

a) SIMULATION MODEL DESCRIPTION DATA STRUCTURES:

Data about the network subsections' circuit parameter values is passed in the COMMON array RLC, which is created in subroutine INPUT. In addition to setting up circuit parameter matrix RLC, INPUT calls subroutine CONTROL which interprets the control statements in the simulation model; generates output pointers NEYE, NFF, and NPP; and defines structures to control the voltage level crossing calculations as described below.

INPUT writes the simulation model's filename to AFILE, and generates names for the output files based on AFILE. For a model in file MODELXYZ., these other files are:

BFIL = MODELXYZ.TG = time domain graph output file (REGIS)
DFIL = MODELXYZ.TD = ASCII time domain output data file
FGLE = MODELXYZ.FG = power spectrum graph output file (REGIS)
FDLE = MODELXYZ.FD = ASCII power spectrum data file
PGLE = MODELXYZ.PG = phase angle graph output file (REGIS)
PDLE = MODELXYZ.PD = ASCII phase angle data file
CRLE = MODELXYZ.DAT = ASCII file with voltage crossing times
and extrema times and values; timing jitter

INPUT also creates the vector KIND, which is a pointer to the circuit type for each of the network subsections. KIND(N) is used to define the network type for the N-th subsection:

KIND(N) = 1 <---> GEN
 2 <---> TEE, etc. (see Parameter Map for RLC for
 further detail - next section)

The period of the waveform, PER, is also passed in COMMON, and is taken to be the maximum of the periods of all the generators in the model.

The generator output impedance is taken to be the sum, RGEN, of the internal resistances of all the generators in the simulation model, and is passed in COMMON.

b) CONTROL DATA STRUCTURES

The following structures have their values defined in subroutine CONTROL, and are passed in COMMON:

- NOUT = vector of pointers to locations of .OUT statements; NOUT(K) is the number of the circuit element after which the K-th .OUT statement appears; NOUT is initially set in SUBROUTINE CONTROL, and is later corrected in SUBROUTINE PRECALC if overlapping 1's case was encountered between 1st and last input GEN
- MOUT = number of .OUT statements in the simulation model; MOUT must be less than or equal to 5
- KD = vector of pointers to those .OUT statements which specify that extrema also be calculated; KD(J)=1 if the J-th .OUT also requires extrema, else KD(J)=0
- IFF = pointer indicating that the number of frequency terms, NF, is to be limited to 8192 to speed up calculations

NCROSS(5,3) is a matrix of pointers to those .OUT statements which request that crossing times for V1, V2, or V3 be computed. NCROSS(K,J) is equal to one if Vj (j=1,2,3) is specified in the K-th .OUT statement in the simulation model.

VOLT(5,3) contains the values of the crossing levels; VOLT(K,J) indicates that the K-th .OUT statement (which is located after network subsection NOUT(K)) specifies that FOCAS is to calculate the times at which $V_j = VOLT(K,J)$.

Additional control parameters that are defined in subroutine CONTROL and passed as parameters in the subroutine calling sequence are:

- NEYE = 1 if an EYE pattern is required
- NFF = 1 if power spectrum calculation required
- NPP = 1 if phase shift vs. frequency calculation required
- DC = vector of DC waveform offsets for the various .OUT statements
- VOFF = vertical axis offset; set in the .END or .EYE statement to shift location of zero volts in the time domain graph
- TOFF = vector of horizontal shifts for each of the .OUT statements
- NC = number of circuit subsections in the model
- TDIS = time interval to be graphed along the horizontal axis in time domain graphs
- ES = full scale vertical axis length for time domain graph

Definition of the model's circuit and control descriptions is complete when subroutine INPUT is exited. Next, subroutine PRECALC does some initial setup calculations and data massaging of circuit parameters. Intermediate calculations which need be performed only once because they are not frequency dependent are performed in PRECALC. An example of such a computation is the multiplication of resistance times capacitance, $R \cdot C$, which is needed in each frequency calculation, but does not change, so can be done only once for all, then passed through RLC. PRECALC also checks and corrects for overlapping leading and trailing ones in the input data stream. PRECALC also returns:

NGEN = number of generator in the model,
after overlapping 1's have been
eliminated

NF = number of frequency terms to be calculated

c) NETWORK CALCULATIONS AND DATA FLOW

After all the initial setup, FOCAS is ready to perform the detailed network analysis. This is all performed in the frequency domain, then inverse Fourier Transformed back into the time domain.

The frequency domain analysis is performed in subroutine TRANSFORM. At each frequency, TRANSFORM will calculate the network response at each of the MOUT .OUT statements. The treatment of the internal generators and the circuit subsections is somewhat different. The user specifies the generator's time domain behavior, but FOCAS automatically converts this waveform into its frequency spectrum. This is possible because the specific generator models in FOCAS all have simple, standard fourier series representations.

For additional generator models, which must exhibit more detailed source behavior, one defines a function, SOURCE, which would Fourier Transform the input waveform using subroutine FFT842, then return the frequency component when called by subroutine TRANSFORM.

A circuit's behavior is determined by its chain matrix, which is a simple algebraic function of the circuit parameters. The A, B, C, D parameters of the chain matrix are calculated in the subroutines or functions called by the subsection type name, e.g. GEN, TEE, BPF. They are then stored in COMMON array ABCD.

ABCD = array of A, B, C, D parameters for each two terminal-pair network

$ABCD(N,1,1) = A$ for N-th network
 $ABCD(N,1,2) = B$ for N-th network
 $ABCD(N,2,1) = C$ for N-th network
 $ABCD(N,2,2) = D$ for N-th network

TRANSFORM works through the entire network, one angular frequency at a time, starting at the DC term ($w=0$), and stepping by an amount equal to $(2*PI/PER)$ up to the maximum angular frequency: $(NF-1)*(2*PI/PER)$.

For each frequency, TRANSFORM will calculate the response of each subsection, then multiply the subsection responses together to find the entire network's response at the particular frequency. The frequency domain response thus calculated is returned in the COMMON array EW(5,8192). The first index of EW points to which of the MOUT .OUT waveforms is described. The second index indicates the frequency term. Thus,

$EW(K,IX)$ = complex frequency response of the network
 after the NOUT(K)-th circuit subsection,
 (i.e. after the K-th .OUT statement)

If an overflow condition is detected during the complex multiplications in TRANSFORM, IFLAG is set, and the frequency calculations will be terminated before all NF terms have been computed. An error message is then printed to alert the user.

Once the complex frequency domain response of the network is known, the power spectrum, phase shift, and time domain computations are performed directly in FOCAS_ANALYZE. These waveforms are all calculated at one network location at a time. At the K-th .OUT statement (after subsection NOUT(K)), the real and imaginary parts of the frequency response are found:

$RX(IW)$ = real part of $EW(K,IW)$
 $RY(IW)$ = imaginary part of $EW(K,IW)$.

The power spectrum is then simply the complex magnitude of EW and is passed to GRAPHICS in the COMMON array FOUT.

$EOUT(K,IW) = \text{SQRT}((RX(IW)**2) + (RY(IW)**2))$

The phase angle is the arctan($RY(IW)/RX(IW)$) and is also passed through COMMON array EOUT.

To find the time domain response of the network after the K-th .OUT, subroutine FFT842 must inverse fourier transform the frequency response after the NOUT(K)-th subsection. The input waveform and its transform are passed to and from FFT842 in arrays RX and RY in its calling sequence.

The time domain response is the real part of the inverse fourier transform of the frequency response. It is passed to support routines CROSS and GRAPHICS in COMMON array EOUT(5,8192).

Subroutines CROSS and GRAPHICS operate on the waveforms stored in array EOUT. Thus, when graphing the power spectrum rather than time domain response on the network, FOUT must be copied into EOUT. Similarly, when graphing the phase angle output, GOUT must be copied into EOUT.

FUNCTIONS AND SUBROUTINES REFERENCED: CROSS, GRAPHICS, EOUT, FOUT, GOUT, COMMON, ARRAY, EOUT(5,8192), SUBROUTINES CROSS AND GRAPHICS OPERATE ON THE WAVEFORMS STORED IN ARRAY EOUT. THUS, WHEN GRAPHING THE POWER SPECTRUM RATHER THAN TIME DOMAIN RESPONSE ON THE NETWORK, FOUT MUST BE COPIED INTO EOUT. SIMILARLY, WHEN GRAPHING THE PHASE ANGLE OUTPUT, GOUT MUST BE COPIED INTO EOUT.

2. CALLING STRUCTURES

FOCAS_ANALYZE

FOCAS_ANALYZE is the skeleton of FOCAS. It calls all of the subroutines to read and interpret the simulation model, perform the network analysis, and write the output files.

VARIABLES: CHAR AFILE, ANF, CHAR BELL, CHAR BFILE, CHAR CRLE, CHAR DFILE, ERR, CHAR ESC, FDIS, CHAR FDLE, CHAR FGLE, FS, I, ID, IERR, IFF, IK, IRR, K, MOUT, NC, NEYE, NF, NF1, NFF, NGEN, NPP, CHAR PDLE, PER, CHAR PGLE, PI2, RGEN, TDIS, VOFF, CHAR W132, CHAR W80, ZE

ARRAYS: ABCD, DC, EDGE, EOUT, EW, KD, KIND, NCROSS, NOUT, RX, RY, TOFF, VOLT

FUNCTIONS AND SUBROUTINES REFERENCED: CROSS, FFT842, GRAPHICS, INPUT, MTH\$ATAN2, MTH\$SQRT, PRECALC, TRANSFORM

SUBROUTINE BPF(N,W)

Returns the A, B, C, D parameters for a Bandpass Filter modelled as a trapeziod in Log P(w) vs w.
(TYPE 15 CIRCUIT)

VARIABLES: A, AA, B, BB, F, FF, K, N, PHI, PI2, S1, S2, W, W1, W2, ZERO

ARRAYS: ABCD, RLC

FUNCTIONS AND SUBROUTINES REFERENCED: MTH\$COS, MTH\$SIN

FUNCTION BRACKET(T,X)

* KEEPS THE TRUNCATED VALUE OF T BETWEEN 0 AND X

VARIABLES: T, X

SUBROUTINE CONTROL(M,NC,ITEM,TDIS,TOFF,DC,FS,VOFF,IERR,NEND,NEYE,NFF,NPP)

* PERFORMS CONTROL STATEMENT PROCESSING

VARIABLES: FS, IERR, IFF, J, K, M, MOUT, NC, NCONT, CHAR ND, NEYE, NFF, NJ, CHAR NOW, NPP, NVAR, RGEN, CHAR SPA, CHAR TAB, TDIS, VOFF

ARRAYS: DC, CHAR ITEM, KD, KIND, CHAR NAME, NCROSS, CHAR ND, NOUT, TOFF, VOLT

FUNCTIONS AND SUBROUTINES REFERENCED: DIGIT

SUBROUTINE CROSS (AFILE, CRLE, NF, TDIS, TOFF, NEYE)

* COMPUTES THE CROSSOVER TIMING, MAXIMA & MINIMA, TIMING JITTER, AND PER CENT NOISE MARGIN OF WAVEFORMS PASSED IN COMMON ARRAY EOUT. WRITES RESULTS TO FILE CRLE

VARIABLES: CHAR AFILE, AJ, AN1, AN2, ANF, AX1, AX2, CHAR CRLE, DELT, DELTT, I, IC1, IC2, IN, INN, INO, INO2, INX, IVH, IVL, IVLO, IVM, IXN, IXX, J, J1, J2, JH, JL, JLO, JP, JT, K, KDER, M, MAX, MAXO, MOUT, N, N1, N2, NDER, NEYE, NF, NM, PER, RGEN, TDIS, TJ1, TJ2, TO, V1, V2, VO

ARRAYS: DERMAX, DERMIN, EOUT, EW, IVN, IVX, KD, KIND, MAXMIN, NCROSS, NMAT, NN, NOUT, NX, RLC, RN, TD, TMAT, TOFF, VMAT, VOLT, XTD

FUNCTIONS AND SUBROUTINES REFERENCED: FORSCLOSE, FORSOPEN

SUBROUTINE DELTA (N,W)

* SOLVES FOR THE A,B,C,D PARAMETERS FOR THE PI CIRCUIT FOR THE Nth NETWORK ELEMENT AT FREQUENCY W (TYPE 3 CIRCUIT)

VARIABLES: A, B, BC1, BC2, D, G1, G2, N, PI2, R, RP, W, XL, ZERO

ARRAYS: ABCD, RLC

FUNCTION DIGIT (M, ITEM, IERR)

* CONVERTS A CHARACTER STRING TO A NUMBER

VARIABLES: A, I, IERR, J, K, L, M, CHAR N, PLUS

ARRAYS: CHAR INTEG, CHAR ITEM, TEMP

SUBROUTINE EAMP (N,W)

* RETURNS THE A, B, C, D PARAMETERS FOR A VOLTAGE DEPENDENT VOLTAGE SOURCE WITH EXTERNAL RESISTANCE AND A DEFINED BANDWIDTH (TYPE 6 CIRCUIT)

VARIABLES: D, N, PI2, W, WO, ZERO

ARRAYS: ABCD, RLC

SUBROUTINE FFT842(IN, N, X, Y)

* THIS PROGRAM REPLACES THE VECTOR $Z=X+iy$ BY ITS FINITE DISCRETE COMPLEX FOURIER TRANSFORM IF $IN=0$ (FREQUENCY DOMAIN TO TIME DOMAIN). THE INVERSE TRANSFORM IS CALCULATED FOR $IN=1$. IT PERFORMS AS MANY BASE 8 ITERATIONS AS POSSIBLE AND THEN FINISHES WITH A BASE 4 ITERATION OR A BASE 2 ITERATION IF NEEDED.

VARIABLES: F1, FN, I, IJ, IN, IPASS, J, J1, J10, J11, J12, J13, J14, J2, J3, J4, J5, J6, J7, J8, J9, J1, L1, L10, L11, L12, L13, L14, L15, L2, L3, L4, L5, L6, L7, L8, L9, LENGT, M, N, N2POW, N8POW, NT, NTHPO, NXTLT, P7, P12, R

ARRAYS: L, X, Y

FUNCTIONS AND SUBROUTINES REFERENCED: ATAN, SQRT, R2TX, R4TX, R8TX

SUBROUTINE FIBER(N,W)

* OPTICAL FIBER MODEL - Returns A, B, C, D parameters (TYPE 7 CIRCUIT)

VARIABLES: AO, AW, H, N, P, PHI, P12, W, WTC, WTM, ZERO

ARRAYS: ABCD, RLC

FUNCTIONS AND SUBROUTINES REFERENCED: MTHSDCOS, MTHSDEXP, MTHSDSIN

SUBROUTINE FILTER(N,W)

* DELAY LINE FILTER MODEL (TYPE 12 CIRCUIT)

VARIABLES: M, N, P, P12, Q, W, X, ZERO

ARRAYS: ABCD, RLC

FUNCTIONS AND SUBROUTINES REFERENCED: MTHSDCOS

FUNCTION GEN(N,W)

* RETURNS THE VOLTAGE GENERATOR FREQUENCY COMPONENT FOR THE Nth NETWORK ELEMENT AT FREQUENCY W (TYPE 1 CIRCUIT)

VARIABLES: AF, AMP, AP, AR, BETA, C, DF, DN, DNC, DO, N, PER, PHI, P12, PW, TFM, TRM, W, WO, ZERO

ARRAYS: ABCD, RLC

FUNCTIONS AND SUBROUTINES REFERENCED: MTHSDCOS, MTHSDSIN

SUBROUTINE GRAPHICS (AFILE,BFILE,DFILE,NF,TDIS,TOFF,DC,FS,VOFF,
1 NEYE,ID)

* WRITES TO THE OUTPUT FILES

VARIABLES: CHAR AFILE, AFS, AMP, ANF, AO, AVOFF, CHAR BFILE,
DELT, CHAR DFILE, DT, EO, EOFF, ESC, FS, FWO, I,
ID, IFF, IN, J, K, MM, MOUT, NEYE, NF, NP, NT, NT1,
OUT, P2, PER, RGEN, T, TC, TDIS, TG, TG2, TGO,
TINC, TO, TOUT, TS, VOFF, Y

ARRAYS: DC, EOUT, EW, KD, KIND, NOUT, ORD, OUTPUT, TOFF

FUNCTIONS AND SUBROUTINES REFERENCED: BRACKET, FOR\$CLOSE, FOR\$OPEN

FUNCTION GSIN(N,W)

* RETURNS THE SINEWAVE GENERATOR FREQUENCY COMPONENT
FOR THE Nth NETWORK ELEMENT AT FREQUENCY W (TYPE 8 CIRCUIT)

VARIABLES: DN, N, PER, PHI, PI2, W, WO, ZERO

ARRAYS: ABCD, RLC

FUNCTIONS AND SUBROUTINES REFERENCED: MTH\$DCOS, MTH\$DSIN

SUBROUTINE HPF(N,W)

* SOLVES FOR THE A,B,C,D PARAMETERS FOR THE HPF CIRCUIT
FOR THE Nth NETWORK ELEMENT AT FREQUENCY W (TYPE 10 CIRCUIT)

VARIABLES: AI, AR, B, D, GC, GL, GLW, N, PI2, W, ZERO

ARRAYS: ABCD, RLC

SUBROUTINE IAMP(N,W)

* A VOLTAGE DEPENDENT CURRENT SOURCE WITH EXTERNAL RESISTANCE
AND A DEFINED BANDWIDTH (TYPE 9 CIRCUIT)

VARIABLES: D, N, PI2, W, WO, ZERO

ARRAYS: ABCD, RLC

SUBROUTINE INFILTER (NVAR,K,ITEM)

* Special Handling for FILTER Network, places the implied after the Sine, Cosine, or Gaussian statements

VARIABLES: I, K, NVAR

ARRAYS: CHAR ITEM

SUBROUTINE INPUT (AFIL,BFIL,DFIL,FGL,FDL,PGL,PDL,NC,TDIS,TOFF,DC,CRL,FS,VOFF,NEYE,ID,nff,npp)

VARIABLES: CHAR AFIL, CHAR BFIL, CHAR CRL, CROSS_SIZE, CHAR DFIL, CHAR FDL, CHAR FGL, FREQ_DATA_SIZE, FREQ_GRAPH_SIZE, FS, I, CHAR ICDAT, ID, IEIERR, IFF, IT, J, K, LT, MOUT, NC, CHAR ND, NEND, NEYE, NFF, NKind, CHAR NOUT, NPP, NVAR, CHAR PDL, PER, CHAR PGL, PHASE_DATA_SIZE, PHASE_GRAPH_SIZE, PI2, RGEN, SIZE, CHAR SPA, STATUS, CHAR TAB, TDIS, TIME_DATA_SIZE, TIME_GRAPH_SIZE, VO

ARRAYS: ABCD, DC, CHAR ITEM, KD, KIND, CHAR NAME, CHAR NDATA, NOUT, RLC, TOFF

FUNCTIONS AND SUBROUTINES REFERENCED: CONTROL, DIGIT, FORSCLOSE, FORSOPEN, INFILTER, LIBSSIGNAL, STRSTRIM

SUBROUTINE LINE (N,W)

* SOLVES FOR THE A,B,C,D PARAMETERS FOR A TRANSMISSION LINE FOR THE Nth NETWORK ELEMENT AT FREQUENCY W (TYPE 4 CIRCUIT)

VARIABLES: A, B, BC, C, CON, D, GAM, GAMJ, N, PI2, Q, QO, R, RT, W, W1, W2, X, XL, Y, Z, ZERO, ZO

ARRAYS: ABCD, RLC

FUNCTIONS AND SUBROUTINES REFERENCED: MTHSALOG, MTHSCDCOS, MTHSCDSIN, MTHSCDSQRT, MTHSCOS, MTHSCOSH, MTHSSIN, MTHSSINH, MTHSSQRT

SUBROUTINE MABCD (A,M,IFLAG)

* SOLVES THE MATRIX MULTIPLICATION $A(I,J) = ABCD(M,I,J)$

VARIABLES: C, I, IFLAG, J, K, M, N, PI2, ZERO

ARRAYS: A, ABCD, B, RLC

FUNCTIONS AND SUBROUTINES REFERENCED: MTHSCDABS

SUBROUTINE MOVE (A,M)

* MOVES ABCD (M,I,J) TO A(I,J)

VARIABLES: I, J, M, N, PI2, ZERO

ARRAYS: A, ABCD, B, RLC

SUBROUTINE PRECALC (NC,NF,TDIS,TOFF,NGEN)

* CALCULATION OF CERTAIN CIRCUIT PARAMETERS TO REDUCE
RECURRENT CALCULATIONS, INITIALIZATIONS, & OVERLAPPING 1
CORRECTION

VARIABLES: A, ALC, ANF, B, BT, BW, C, CORR, DEL, DEL1, DN, DN1,
DP, DT, ERR, F1, F2, I, IBG, ICK, ICT, IDF, IFF, IGEN,
II, J, K, MOUT, N1, NC, NDGEN, NF, NGEN, NSRC, NT, P,
P1, P2, PER, PI2, PN, PN1, R1, R2, RGEN, T, TDF50,
TDIS, TDR50, TDR51, TDR52, TF, TR, TR1, TR2, TRM, TRO,
TYPE, WO, X, XC, XM, ZERO

ARRAYS: ABCD, DC1,EDGE,KD, KIND, NOUT, RLC, TOFF, XX, YY

FUNCTIONS AND SUBROUTINES REFERENCED: FFT842, FOR\$CLOSE, FOR\$OPEN,
MTH\$ALOG, MTH\$ALOG10, MTH\$DSQRT, MTH\$SQRT, RISE

FUNCTION RISE (WGEN,TR,A)

* COMPUTES THE DELAY FOR THE INPUT WAVEFORM
WGEN IS THE GENERATOR FILTER 3DB FREQUENCY (RAD/SEC).
TR IS THE 20 % TO 80 % GENERATOR RISE TIME
A IS THE FRACTIONAL AMPLITUDE

VARIABLES: A, C, DF, F, I, TR, WGEN, X, Y

FUNCTIONS AND SUBROUTINES REFERENCED: MTH\$EXP

SUBROUTINE R2TX (NTHPO, CRO, CR1, CIO, C11)

RADIX 2 ITERATION SUBROUTINE OF FFT842

VARIABLES: F11, K, NTHPO, R1

ARRAYS: CIO, C11, CRO, CR1

SUBROUTINE R4TX (NTHPO, CRO, CR1, CR2, CR3, C10, C11, C12, C13)

* RADIX 4 ITERATION SUBROUTINE OF FFT842

VARIABLES: F11, F12, F13, F14, K, NTHPO, R1, R2, R3, R4

ARRAYS: C10, C11, C12, C13, CRO, CR1, CR2, CR3

SUBROUTINE R8TX (NXTLT, NTHPO, LENGT, CRO, CR1, CR2, CR3, CR4, CR5, CR6, CR7, C10, C11, C12, C13, C14, C15, C16, C17)

* RADIX 8 ITERATION SUBROUTINE OF FFT842

VARIABLES: A10, A11, A12, A13, A14, A15, A16, A17, ARO, AR1, AR2, AR3, AR4, AR5, AR6, AR7, ARG, B10, B11, B12, B13, B14, B15, B16, B17, BRO, BR1, BR2, BR3, BR4, BR5, BR6, BR7, C1, C2, C3, C4, C5, C6, C7, J, K, LENGT, NTHPO, NXTLT, P7, P12, S1, S2, S3, S4, S5, S6, S7, SCALE, T1, TR

ARRAYS: C10, C11, C12, C13, C14, C15, C16, C17, CRO, CR1, CR2, CR3, CR4, CR5, CR6, CR7

FUNCTIONS AND SUBROUTINES REFERENCED: COS, SIN

SUBROUTINE SERIES (N,W)

* POLYNOMIAL MODEL FOR A SERIES IMPEDANCE
UP TO FOUR ZEROS AND FOUR POLES (TYPE 14 CIRCUIT)

VARIABLES: N, P1, P12, PR, W, W2, W3, W4, ZERO, Z1, ZR

ARRAYS: ABCD, RLC

SUBROUTINE SHUNT (N,W)

* POLYNOMIAL MODEL FOR A SHUNT IMPEDANCE
UP TO FOUR ZEROS AND FOUR POLES (TYPE 13 CIRCUIT)

VARIABLES: N, P1, P12, PR, W, W2, W3, W4, ZERO, Z1, ZR

ARRAYS: ABCD, RLC

FUNCTION SOURCE (N,W)

* RETURNS THE SOURCE GENERATOR FREQUENCY COMPONENT
FOR THE Nth NETWORK ELEMENT AT FREQUENCY W
TYPE 16 CIRCUIT - SRC, experimental data

VARIABLES: AX, IFOLD, IX, N, PER, P12, W, ZERO

ARRAYS: ABCD, EDGE, RLC

SUBROUTINE STUB (N,W)

* SOLVES FOR THE A,B,C,D PARAMETERS FOR A TRANSMISSION LINE
STUB FOR THE Nth ELEMENT AT FREQUENCY W (TYPE 5 CIRCUIT)

VARIABLES: A, BC, C, CON, D, GAM, GAMJ, N, PI2, Q, QO, R,
RT, W, X, XL, Y, YL, Z, ZERO, ZO

ARRAYS: ABCD, RLC

FUNCTIONS AND SUBROUTINES REFERENCED: MTH\$ALOG, MTH\$CDCOS, MTH\$CDSIN,
MTH\$CDSQRT, MTH\$COS, MTH\$COSH, MTH\$SSIN, MTH\$SSINH,
MTH\$SQRT

SUBROUTINE TEE (N,W)

* SOLVES FOR THE A,B,C,D PARAMETERS FOR THE TEE CIRCUIT
FOR THE Nth NETWORK ELEMENT AT FREQUENCY W (TYPE 2 CIRCUIT)

VARIABLES: BC, G, N, PI2, R1, R2, RAC, RT, W, W1, W2, XL1,
XL2, ZERO

ARRAYS: ABCD, RLC

SUBROUTINE TRANSFORM (NC,NF,NGEN)

* COMPUTES THE FREQUENCY DOMAIN RESPONSE OF THE NETWORK

VARIABLES: EG, I, IFF, IFLAG, K, KOUT, M, MM, MOUT, N1, N2,
NC, NF, NGEN, NS, PER, PI2, RGEN, W, WO, ZERO

ARRAYS: A, ABCD, B, EOUT, EW, KD, CHAR NODENAME, NOUT, RLC

FUNCTIONS AND SUBROUTINES REFERENCED: BPF, DELTA, EAMP, FIBER,
FILTER, GEN, GSIN, HPF, IAMP, LINE, MABCD, MOVE, SERIES,
SHUNT, SOURCE, STUB, TEE, XFORMER

SUBROUTINE XFORMER (N,W)

* TRANSFORMER MODEL (TYPE 11 CIRCUIT)

VARIABLES: M, N, PI2, Q, W, ZERO

ARRAYS: ABCD, RLC

B. PARAMETER MAP FOR THE ARRAY RLC

The array RLC in FOCAS passes circuit parameters between the various subroutines. Element RLC(n,p) is the p-th circuit parameter of n-th subsection of the simulation model. RLC is used in conjunction with array KIND(14) to completely specify a subsection. The numerical value of KIND(n) identifies the subsection circuit type. For example if KIND(53) = 7, the fifty-third element in the simulation model is an optical fiber. RLC(53,8) is the length (X) of the fiber as shown below.

KIND = Type P	1 GEN	2 TEE	3 DELTA	4 LINE	5 STUB	6 EAMP	7 FIBER	8 SIN	9 IAM
1	PW	R1	G1	Z0	Z0	GAIN	AO	F	GAIN
2	TR	R2	G2	TD	TD	R	AW	AMP	R
3	PER	G	C1	X	X	F	FWHM	DC	G
4	AMP	L1	C2	R	R		F1	DEL	F
5	DEL	L2	L	TC	TC		D	R	
6	BW	C	R	D	D		QM		
7	DC	F1	R1	K1			QC		
8	R	F2	R2	K2			X		
9	TF	R	RP	WC	GL		LC		
10				Q	RL		LO		
11				F1	CL		SO		
12				F2	WC				
13				F3					
14									
15									
16	WO	W1		CONST	CONST	WO	A	WO	
17		W2					TC		
18							A'		
19	TFM			L	L		TM		
20	TRM			C	C				

1-13 INPUT PARAMETERS
 14-20 PARAMETERS COMPUTED IN SUBROUTINE PRECALC

PARAMETER MAP FOR THE ARRAY RLC (cont.)

ND	10	11	12	13	14	15	16
type	HPF	XFORMER	FILTER	SHUNT	SERIES	BPF	SRC

C	L1	TO	A0	A0	F1	EF
L	L2	N	A1	A1	F2	PW
G	M	R2	A2	A2	S1	PER
R	K	G1	A3	A3	S2	AMP
	N	R1	A4	A4	K	DEL
		FO	B0	B0	PH	NT
			B1	B1		DT
		SINE	B2	B2		TYPE
		COS	B3	B3		R
		GAUS	B4	B4		
			K	K		

					PS
	L1/M				AA
	L2/M				BB
	M-L1L2/M				A
G/C					B
G*L		TYPE			M

1-13 INPUT PARAMETERS
 14-20 PARAMETERS COMPUTED IN SUBROUTINE PRECALC

TYPE = 1, SINE (not implemented)
 2, COSINE
 3, GAUSSIAN (not implemented)

TABLE

CONTENTS

CC

+

F

A

C

M

CC

CC

CC

FOCAS_ANALYZE is the skeleton of FOCAS. It calls all of the subroutines to read and interpret the simulation model, perform the network analysis, and write the output files.

1. MODEL DESCRIPTION DATA STRUCTURES:

Data about the network subsections' circuit parameter values is passed in the COMMON array RLC, which is created in subroutine INPUT. In addition to setting up circuit parameter matrix RLC, INPUT calls subroutine CONTROL which interprets the control statements in the simulation model; generates output pointers NEYE, NFF, and NPP; and defines structures to control the voltage level crossing calculations as described below.

INPUT writes the simulation model's filename to AFILE, and generates names for the output files based on AFILE. For a model in file MODELXYZ., these other files are:

BFILE = MODELXYZ.TG = time domain graph output file (REGIS)
 DFILE = MODELXYZ.TD = ASCII time domain output data file
 FGLE = MODELXYZ.FG = power spectrum graph output file (REGIS)
 FDLE = MODELXYZ.FD = ASCII power spectrum data file
 PGLE = MODELXYZ.PG = phase angle graph output file (REGIS)
 PDLE = MODELXYZ.PD = ASCII phase angle data file
 CRLE = MODELXYZ.DAT = ASCII file with voltage crossing times and extrema times and values; timing jitter

INPUT also creates the vector KIND, which is a pointer to the circuit type for each of the network subsections. KIND(N) is used to define the network type for the N-th subsection:

KIND(N) =

1	<--->	GEN
2	<--->	TEE
3	<--->	DELTA
4	<--->	LINE
5	<--->	STUB
6	<--->	EAMP
7	<--->	IAMP
8	<--->	FIBER
9	<--->	SINE
10	<--->	HPF
11	<--->	XFORMER
12	<--->	FILTER
13	<--->	SHUNT
14	<--->	SERIES
15	<--->	BPF
16	<--->	SRC

The period of the waveform, PER, is also passed in COMMON, and is taken to be the maximum of the periods of all the generators in the model. The generator output impedance is the sum, RGEN, of the internal resistances of all the generators in the simulation model, and is passed in COMMON.

2. CONTROL DATA STRUCTURES

The following structures have their values defined in subroutine CONTROL, and are passed in COMMON:

- NOUT = vector of pointers to locations of .OUT statements; NOUT(K) is the number of the circuit element after which the K-th .OUT statement appears; NOUT is initially set in SUBROUTINE CONTROL, and is later corrected in SUBROUTINE PRECALC if overlapping 1's case was encountered between 1st and last input GEN
- MOUT = number of .OUT statements in the simulation model; MOUT must be less than or equal to 5
- KD = vector of pointers to those .OUT statements which specify that extrema also be calculated; KD(J)=1 if the J-th .OUT also requires extrema, else KD(J)=0
- IFF = pointer indicating that the number of frequency terms, NF, is to be limited to 8192; used to speed up calculations

NCROSS(5,3) is a matrix of pointers to those .OUT statements which request that crossing times for V1, V2, or V3 be computed. NCROSS(K,J) is equal to one if Vj (j=1,2,3) is specified in the K-th .OUT statement in the simulation model.

VOLT(5,3) contains the values of the crossing levels; VOLT(K,J) indicates that the K-th .OUT statement (which is located after network subsection NOUT(K)) specifies that FOCAS is to calculate the times at which $V_j = VOLT(K,J)$.

Additional control parameters that are defined in subroutine CONTROL and passed as parameters in the subroutine calling sequence are:

- NEYE = 1 if an EYE pattern is required
- NFF = 1 if power spectrum calculation required
- NPP = 1 if phase shift vs. frequency calculation required
- DC = vector of DC waveform offsets for the various .OUT statements
- VOFF = vertical axis offset; set in the .END or .EYE statement to shift location of zero volts in the time domain graph
- TOFF = vector of horizontal shifts for each of the .OUT statements
- NC = number of circuit subsections in the model
- TDIS = time interval to be graphed along the horizontal axis in time domain graphs
- FS = full scale vertical axis length for time domain graph

Definition of the model's circuit and control descriptions is complete when subroutine INPUT is exited. Next, subroutine PRECALC does some initial setup calculations and data massaging of circuit parameters. Intermediate calculations which need be performed only once because they are not frequency dependent are performed in PRECALC. An example of such a computation is the multiplication of resistance times capacitance, $R \times C$, which is needed in each frequency calculation, but does not change, so can be done only once for all, then passed through RLC. PRECALC also checks and corrects for overlapping leading and trailing ones in the input data stream.

PRECALC also returns:

```

NGEN = number of generator in the model,
      after overlapping 1's have been
      eliminated
NF   = number of frequency terms to be calculated
  
```

3. NETWORK CALCULATIONS AND DATA FLOW

After all the initial setup, FOCAS is ready to perform the detailed network analysis. This is all performed in the frequency domain, then inverse Fourier Transformed back into the time domain.

The frequency domain analysis is performed in subroutine TRANSFORM. At each frequency, TRANSFORM will calculate the network response at each of the MOUT .OUT statements. The treatment of the internal generators and the circuit subsections is somewhat different. The user specifies the generator's time domain behavior, but FOCAS automatically converts this waveform into its frequency spectrum. This is possible because the specific generator models in FOCAS all have simple, standard fourier series representations.

For additional generator models, which must exhibit more detailed source behavior, one defines a function, SOURCE, which would Fourier Transform the input waveform using subroutine FFT842, then return the frequency component when called by subroutine TRANSFORM.

A circuit's behavior is determined by its chain matrix, which is a simple algebraic function of the circuit parameters. The A, B, C, D parameters of the chain matrix are calculated in the subroutines or functions called by the subsection type name, e.g. GEN, TEE, BPF. They are then stored in COMMON array ABCD.

ABCD = array of A, B, C, D parameters for each two terminal-pair network;

```

ABCD(N,1,1) = A for N-th network
ABCD(N,1,2) = B for N-th network
ABCD(N,2,1) = C for N-th network
ABCD(N,2,2) = D for N-th network
  
```



```

CHARACTER*1 ESC,BELL
CHARACTER*4 W80,W132
CHARACTER*32 AFILE,BFILE,DFILE,FGLE,FDLE,PGLE,PDLE,CRLE
DIMENSION RX(32768),RY(32768),TOFF(5),DC(5)
DOUBLE PRECISION PER,P12
COMPLEX*16 ABCD(500,2,2),EW(5,32768),ZERO
complex*16 edge(8192)
COMMON/NET/RLC(500,20),ABCD,P12,ZERO
COMMON/MISC/KIND(500),RGEN,NOUT(5),MOUT,KD(5),iff
COMMON/WAVES/EW,EOUT(5,32768)
COMMON/TIME/PER
common/expt1/edge
COMMON/MATCH/VOLT(5,3),NCROSS(5,3)
DATA P12/6.28318530717958/, PER/0.0/, RGEN/0.0/, VOFF/0.0/
DATA W80/'[?31'/', W132/'[?3h'/'
ESC=CHAR(27)
BELL=CHAR(7)
ZERO=DCMPLX(0.0,0.0)
iff=0 !Fourier transform speed up by limiting NF
TYPE *,' Fiber Optic Cable Simulator FOCAS'
TYPE *,' Revised 11 MARCH 1987'
TYPE *

CALL INPUT(AFILE,BFILE,DFILE,FGLE,FDLE,PGLE,PDLE,NC,TDIS,TOFF,DC,
1 CRLE,FS,VOFF,NEYE,ID,nff,npp)

ik=kind(1)
if((ik.eq.1).or.(ik.eq.8).or.(ik.eq.16))go to 686
type *,' '
type *,' *** A SOURCE IS REQUIRED IN THE MODEL *** '
TYPE *,BELL

686 IF(NEYE.EQ.1)TYPE *,' FREQ. DOMAIN RESPONSE WILL NOT BE
1 CALCULATED -- EYE PATTERN REQUESTED'
IF(NEYE.EQ.1)TYPE *,' '

IF(NEYE.EQ.1)ID=0

CALL PRECALC(NC,NF,TDIS,TOFF,NGEN)
nfl=nf ! Used to indicate abnormal end
CALL TRANSFORM(NC,NF,NGEN)
type *,' '
TYPE *,'Number of frequency terms computed =',NF
anf=nf
err=per/anf
ierr=1000*err
irr=(err/2.)*1000
if(ik.eq.16)irr=irr+500*rlc(1,7) !Cannot be more accurate
if(ik.eq.16)ierr=2*irr !than input data
type *,' '
type *,' Time co-ord uncertainty = ',irr,' ps'
if(neye.eq.1)type *,' Timing jitter uncertainty = ',ierr,' ps'
type *,' '
if(nfl.eq.nf)go to 667 ! O.K.

```

```
TYPE *, ' '
type *, ' *** WARNING - reduced accuracy ***'
TYPE *, ' '
IF (NF.GE.128) GO TO 667
TYPE *, ' NOT ENOUGH TERMS TO USE RESULTS '
TYPE *, BELL
STOP

DO K=1, MOUT
DO I=1, NF          !Find time domain response
RX(I)=REAL(EW(K,I)) !Setup for FFT842
RY(I)=-DIMAG(EW(K,I)) !Setup for FFT842
ENDDO

CALL FFT842(0, NF, RX, RY) !Inverse Fourier Transform

DO I=1, NF          !Time domain response
EOUT(K,I)=RX(I)
ENDDO
ENDDO
CALL CROSS(AFILE, CRLE, NF, TDIS, TOFF, NEYE)

ID=ID-1            !Get ready for time domain graph
CALL GRAPHICS(AFILE, BFILE, DFILE, NF, TDIS, TOFF, DC, FS, VOFF, NEYE, ID)
ID=ID+1            !Reset ID to chk for freq domain outputs
IF (ID.EQ.0.0) GO TO 11

if (nff.eq.0.and.npp.eq.0) go to 11 !Done if no power spectra or
!phase angle outputs required

do i=1,5           !Setup offsets to zero for freq. domain outputs
toff(i)=0.
enddo

do k=1,mout        !Setup for power spectrum outputs
do i=1,nf
rx(i)=real(ew(k,i))
ry(i)=dimag(ew(k,i))
eout(k,i)=(rx(i)**2)+(ry(i)**2)
eout(k,i)=sqrt(eout(k,i))
enddo
enddo

FS=amax1(eout(1,1),eout(1,2)) !Rqrd for freq. domain scaling

do k=1,mout        !Find maximum of power spectra for normalization
do i=1,nf          !on .fg graphs
FS=AMAX1(FS,EOUT(K,I))
enddo
enddo

do k=1,mout        !Renormalize power spectrum to its maximum value
do i=1,nf
eout(k,i)=eout(k,i)/fs
enddo
enddo
```

```

do k=1,mout !Don't displace freq. or phase response outputs
dc(k)=0.
enddo
fs=1.

```

```

ANF=NF
FDIS=ANF/(6.*PER) !Max freq to be displayed in spectrum
if(nf.gt.8192) fdis=(8192./(6.*per)) !Limit max. freq displayed
if(nff.eq.0) go to 12
CALL GRAPHICS(AFILE,FGLE,FDLE,NF,fdis,TOFF,DC,FS,0.0,NEYE,10)

```

```

12 if(npp.eq.0) go to 11
do k=1,mout !Setup for phase vs. freq graph
do i=1,nf
eout(k,i)=atan2(ry(i),rx(i))
enddo
enddo
fs=pi2

```

```

CALL GRAPHICS(AFILE,PGLE,PDLE,NF,fdis,TOFF,DC,fs,pi2/2.,NEYE,10)

```

```

11 TYPE *,BELL
STOP
END .1

```

! BPF

SUBROUTINE BPF (N,W)

CC
Returns the A, B, C, D parameters for a Bandpass Filter modelled as a trapeziod in Log P(w) vs w.
(TYPE 15 CIRCUIT)
2/2/87
CC

COMPLEX*16 ABCD (500,2,2),ZERO,FF
DOUBLE PRECISION P12,F
COMMON/NET/RLC (500,20),ABCD,P12,ZERO
phi=0.

w1=rlc (n,1)
W2=RLC (N,2)
S1=RLC (N,3) !Slope in dB/octave
S2=RLC (N,4) !Slope in dB/octave
k=rlc (n,5) !Gain
AA=rlc (n,16)
BB=rlc (n,17)
a=rlc (n,18)
b=rlc (n,19)

IF (W.LE.A.OR.W.GE.B) F=0.0
IF (AA.LE.W.AND.BB.GE.W) F=10.0

IF (A.LT.W.AND.AA.GT.W) go to 20
IF (BB.LT.W.AND.B.GT.W) go to 30

go to 10
f=(w/w1)-1
f=s1*f/10.
f=10**f
f=5*f
go to 10

f=(w/bb)-1
f=s2*f/10.
f=1-f
f=10**f

IF (F.LT.1.E-16) F=1.E-16
f=f/10. !Normalize F btwn 0 and 1
f=k*f !Multiply by gain

if (w.lt.w1) phi=rlc (n,6)
if (w.gt.w2) phi=-rlc (n,6)
if ((w.ge.w1).and.(w.le.w2)) go to 40
go to 50
phi=w-rlc (n,20) !w-midpt btwn w1 and w2
phi=rlc (n,14)*phi
FF=F*DCMPLX (COS (PHI),SIN (PHI))
FF=1./FF

ABCD(N,1,1)=FF
ABCD(N,1,2)=ZERO
ABCD(N,2,1)=ZERO
ABCD(N,2,2)=ZERO

RETURN
END

FUNCTION BRACKET(T,X)

```
CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC  
C  
C   KEEPS THE TRUNCATED VALUE OF T BETWEEN 0 AND X  
C  
CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC  
    IF (T.GT.0.0) GO TO 1  
    BRACKET=T+X  
    RETURN  
1   IF (JINT(T).GT.X) GO TO 2  
    BRACKET=T  
    RETURN  
2   BRACKET=T-X  
    RETURN  
    END  
!
```

! CONTROL

SUBROUTINE CONTROL (M,NC,ITEM,TDIS,TOFF,DC,FS,VOFF,IERR,NEND,NEYE,
1 nff, npp)

CONTROL STATEMENT PROCESSING

CHARACTER*1 ITEM(130),SPA,TAB
CHARACTER*2 NDATA(6,10),ND
CHARACTER*3 NAME(10),NOW
DIMENSION TOFF(4),DC(4)
COMMON/MISC/KIND(500),RGEN,NOUT(5),MOUT,KD(5),iff
COMMON/MATCH/VOLT(5,3),NCROSS(5,3)

DATA NAME/

1 2 3 4 5 6 7 8 9 10
1 'OUT','END','EYE','TIT','FDG','PDG','TIT','TIT','TIT','TIT'/'

DATA NDATA/

1 2 3 4 5 6
1 'V1','V2','V3','TO','DC','DV',
2 'TX','DC','FS','FS','FS','FS',
3 'TX','DC','FS','FS','FS','FS',
4 ' ',' ',' ',' ',' ',' ',
5 'FF',' ',' ',' ',' ',' ',
6 ' ',' ',' ',' ',' ',' ',
7 ' ',' ',' ',' ',' ',' ',
8 ' ',' ',' ',' ',' ',' ',
9 ' ',' ',' ',' ',' ',' ',
1 ' ',' ',' ',' ',' ',' ' /

DATA NCROSS/15*0/, NCONT/6/, SPA/' ', TAB/' '

NOW(1:1)=ITEM(M+1)

NOW(2:2)=ITEM(M+2)

NOW(3:3)=ITEM(M+3)

M=M+4

DO NJ=1,NCONT !NCONT= number of types of control statements

IF (NOW.EQ.NAME(NJ)) GO TO (100,200,200,400,500,600),NJ

ENDDO

TYPE *, '***** IMPROPER CONTROL CHARACTERS *****'

IERR=1

RETURN

IF (MOUT.LT.5) GO TO 110 !.OUT

TYPE *, '***** TOO MANY .OUTs SPECIFIED - LAST ONE IGNORED *****'

RETURN

! CONTROL

```

110      MOUT=MOUT+1
        NOUT(MOUT)=NC
        GO TO 205
200      NEND=1
205      DO K=M,130
        IF (ITEM(K).EQ.'!') RETURN
        IF (ITEM(K).EQ.SPA) GO TO 210
        ENDDO
210      M=K

        DO 220 J=1,6
        DO 230 K=M,130
        IF (ITEM(K).EQ.'!') RETURN
        IF (ITEM(K).EQ.SPA) GO TO 230
        ND(1:1)=ITEM(K)
        ND(2:2)=ITEM(K+1)
        DO NVAR=1,6
        IF (ND.EQ.NDATA(NVAR,NJ)) GO TO 240
        ENDDO
230      CONTINUE
        IF (K.GE.130) RETURN
        TYPE *, '***** UNDEFINED VARIABLE NAME *****'
        IERR=1
        RETURN

240      M=K+1
        IF (NJ.EQ.1.AND.NVAR.EQ.6) GO TO 260
        DO K=M,130
        IF (ITEM(K).EQ.SPA.AND.ITEM(K+1).NE.SPA) GO TO 250
        ENDDO
        TYPE *, '***** INCORRECT DATA FORMAT *****'
        IERR=1
        RETURN

250      M=K+1
        GO TO (260,245,245,400),NJ
245      GO TO (251,252,253),NVAR
251      TDIS=DIGIT(M,ITEM,IERR)
        GO TO 255
252      IF (NVAR.EQ.2) VOFF=DIGIT(M,ITEM,IERR)
        GO TO 255
253      FS=DIGIT(M,ITEM,IERR)
255      IF (NJ.NE.3) GO TO 290
        NEYE=1
        IF (MOUT.EQ.1) GO TO 290
        TYPE *, '***** ONLY ONE OUTPUT ALLOWED FOR EYE *****'
        IERR=1
        RETURN

```

! CONTROL

```

GO TO (280,280,280,270,275,300),NVAR      !OUT statement
TOFF (MOUT)=DIGIT (M, ITEM, IERR)
GO TO 290
DC (MOUT)=DIGIT (M, ITEM, IERR)
GO TO 290
NCROSS (MOUT,NVAR)=1      !V1,V2,V3
VOLT (MOUT,NVAR)=DIGIT (M, ITEM, IERR)
IF (IERR.LT.2) GO TO 220      !Bad input
TYPE *, '***** VARIABLE VALUE ERROR *****'
RETURN
KD (MOUT)=1
CONTINUE
RETURN      !Title Line
nff=1      !Frequency domain outputs

DO K=M,130      !Read rest of line
IF (ITEM (K) .EQ. '!') RETURN
IF (ITEM (K) .EQ. SPA) GO TO 211
ENDDO
M=K

DO 221 J=1,6
DO 231 K=M,130
IF (ITEM (K) .EQ. '!') RETURN
IF (ITEM (K) .EQ. SPA) GO TO 231
ND (1:1)=ITEM (K)
ND (2:2)=ITEM (K+1)
DO NVAR=1,6
IF (ND.EQ.NDATA (NVAR,NJ)) GO TO 241
ENDDO
CONTINUE
IF (K.GE.130) RETURN
TYPE *, '***** UNDEFINED VARIABLE NAME *****'
IERR=1
RETURN

M=K+1
IF (NJ.EQ.5.AND.NVAR.EQ.1) IFF=1
DO K=M,130
IF (ITEM (K) .EQ. SPA.AND.ITEM (K+1) .NE.SPA) GO TO 250
ENDDO
TYPE *, '***** INCORRECT DATA FORMAT *****'
IERR=1
continue
return
npp=1      !Phase shift outputs
return

END

```



```

SUBROUTINE CROSS(AFILE,CRLE,NF,TDIS,TOFF,NEYE)
CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC
C
C      COMPUTES THE CROSSOVER TIMING, MAXIMA & MINIMA,
C      TIMING JITTER, AND PER CENT NOISE MARGIN
C
C      CROSS computes the crossover & extrema timings; and the
C      extrema values of the waveforms passed in array
C      EOUT(5,32768) through COMMON. CROSS writes this
C      data into output file CRLE.
C
C      TMAT = array of cross over timings in nanoseconds
C      DERMAX = array with times of maxima in nanoseconds
C      DERMIN = array with times of minima in nanoseconds
C      IVX = array with maxima values in millivolts
C      IVN = array with minima values in millivolts
C      KD = vector of pointers to .OUT statements requiring
C          extrema calculations
C      RN = vector of percent noise margin for .OUT
C          locations with DV specified, KD(K).ne.0
C      MAXMIN = vector of pointers to .OUT statement locations
C      XTD = TOFF = vector of time offsets defined by user
C
CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC

```

```

CHARACTER*32 AFILE,CRLE
COMPLEX*16 EW(5,32768)
DOUBLE PRECISION PER
DIMENSION TMAT(5,100),NMAT(5),VMAT(5),TOFF(5),TD(20),
1 DERMAX(5,100),DERMIN(5,100),MAXMIN(5),XTD(5)
dimension ivx(5,100),ivn(5,100),rn(5),nn(5),nx(5)
COMMON/NET/RLC(500,20),ABCD,P12,ZERO
COMMON/MISC/KIND(500),RGEN,NOUT(5),MOUT,KD(5)
COMMON/WAVES/EW,EOUT(5,32768)
COMMON/TIME/PER
COMMON/MATCH/VOLT(5,3),NCROSS(5,3)

```

```

DATA KDER/O/,NDER/O/,IN/O/,MAX/O/,INO/O/,MAXO/O/
DATA IVX/500*O/,IVN/500*O/

```

```

C      KDER      Derivative output indicator
C      NDER      Max number of maxima or minima
ANF=NF      !Number of time increments computed
DELT=PER/ANF      !Time delay between samples
DELT=DELT
if(kind(1).eq.16) deltt=deltt+rlc(1,7)
C      IN      Number of Outputs counter, not more than 10
C      MAX      Crossover counter for all outputs
C      N      N is the number of crossovers detected for each
              output, not more than 100

```

```

DO 50 K=1,MOUT
TO=-TOFF(K)      !User input time delay
DO 100 I=1,3
IF(NCROSS(K,I).EQ.0)GO TO 100 !Check if crossing calc. reqrd
VO=VOLT(K,I)    !Crossing level voltage
N=0
IN=IN+1
NMAT(IN)=NOUT(K) !Location of IN-th .OUT
VMAT(IN)=VO
TD(IN)=TOFF(K)
J1=-TO/DELT
IF(J1.LT.1)J1=J1+NF
IVL=1000*(EOUT(K,J1)-VO) !Round off to millivolts
j1o=j1-1
if(j1o.eq.0)j1o=nf
ivlo=1000*(eout(k,j1o)-vo)
JL=J1+1
JH=J1+NF

DO 150 J=JL,JH !Find crossing locations by sign
J2=J !change btwn (IVH-IV0) and (IVL-IV0)
IF(J2.GT.NF)J2=J-NF
iVH=1000*(EOUT(K,J2)-VO)

IF(IVL)10,120,20
IF(IVH)121,200,200
IF(IVH)200,200,121
IF(N.EQ.100)GO TO 240
N=N+1
AJ=J2-2
TMAT(IN,N)=AJ*DELT+TO+DELT*IVL/(IVL-IVH)
go to 122
if(ivlo.ne.0)go to 120 !Check if at corner of constant
if(n.eq.100)go to 240 !region equal to VO
n=n+1
aj=j2-3 !Time index of ivlo
tmat(in,n)=aj*delt+to !Time co-ord of ivlo
ck to make sure this isn't the previous crossing value instead
a corner that would have been missed otherwise:

if(n.eq.1)go to 122 !First point so it can't be a repeat
if(tmat(in,n).ne.tmat(in,n-1))go to 122 !Not a repeat
tmat(in,n)=0. !Reset incorrect crossing value
n=n-1 !Reset crossing counter

IF(NEYE.EQ.0)GO TO 120
DO M=1,100
IF(TMAT(IN,N).LE.TDIS)GO TO 120
TMAT(IN,N)=TMAT(IN,N)-TDIS
ENDDO

IVLO=IVL
IVL=IVH
CONTINUE

```

```

240 IF (N.GT.MAX) MAX=N
    IF (IN.EQ.10) GO TO 300
100 CONTINUE
300 IF (KD(K).EQ.0) GO TO 50      !Extrema calculations setup
    N1=0
    N2=0
    INO=INO+1
    J1=-TO/DELT
    IF (J1.LT.1) J1=J1+NF
    JT=J1+1
    IF (JT.GT.NF) JT=JT-NF
    IVM=EOUT(K,JT)*1000          !Round off slight variations
    IVL=EOUT(K,J1)*1000        !Round off
    j1o=j1-1
    if (j1o.eq.0) j1o=nf
    iv1o=eout(k,j1o)*1000
    jp=0                          !Index of previous extremum

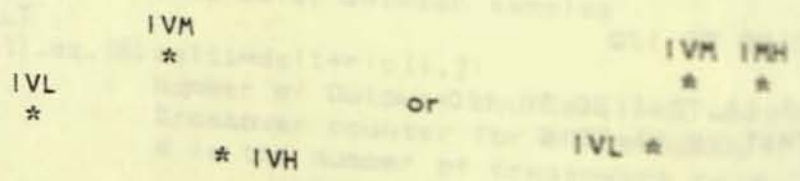
    JL=J1+2
    jh=j1+nf

    DO 350 J=JL,JH                !Find maxima & minima
    J2=J
    IF (J2.GT.NF) J2=J-NF
    IVH=EOUT(K,J2)*1000          !Round off slight variations

```

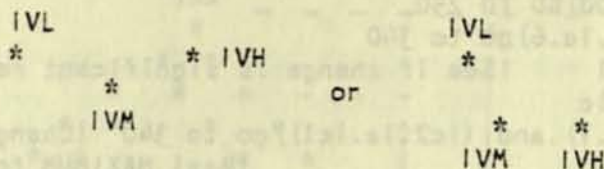
C Extrema are found by sorting through IVL, IVM,
C and IVH, where IVL corresponds to the voltage
C value with low index number of the triplet (L,M,H)
C IVM has the middle index value, and IVH the highest
C of the three index values. When IVM is greater
C than IVL and IVH a maximum is found; when IVM is
C less than IVL and IVH a minimum has been found.
C
C The values of IVL, IVM, and IVH are rounded off to
C an integral number of millivolts to eliminate slight
C variations in values due to numerical roundoffs.
C Similarly, when a local extremum is found, the
C differences between IVL and IVM, and between IVL and
C IVLO (the point preceding IVL) are checked to make sure
C that they are significant (i.e. greater than 1 mv).

C Statement 360 corresponds to treatment of a simple
C maximum of the forms:



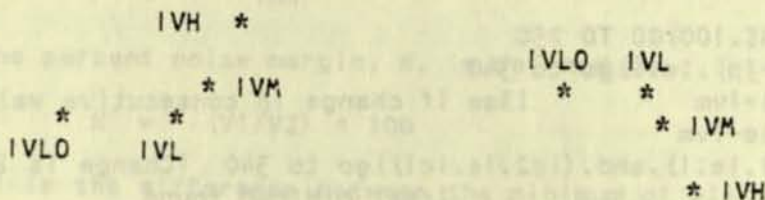
!

Similarly, statement 370 corresponds to treatment of a simple minimum of the forms:



For cases in which a corner (i.e. change in value after a constant region) is encountered, and additionally the low index point from the previous pass through the loop is required to identify the corner. This is saved as IVL0.

Statements 339 and 341 correspond to corner minima and maxima respectively:



```

IF (IVL-IVM) 310,340,320
IF (IVH-IVM) 360,360,339
IF (IVH-IVM) 341,370,370

```

```

if (ivl0.ne.ivl) go to 340 !Check to make sure have a corner min.
if ((j2-jp).le.5) go to 340 !Slow changes or close extrema rejected
if (n2.ge.100) go to 250
n2=n2+1
nn(ino)=n2 !Count number of min for this .OUT
aj=j2-2
dermin(ino,n2)=aj*delt+to !Min. at IVL
maxmin(ino)=nout(k)
ivn(ino,n2)=ivl !Minimum in millivolts

jp=j2-2
go to 340

```

```

if (ivl0.ne.ivl) go to 340 !Check for corner maximum
if ((j2-jp).le.5) go to 340 !Reject slow changes<->close extrema
if (n1.ge.100) go to 250
n1=n1+1
nx(ino)=n1 !Count number of max for this .OUT
aj=j2-2
dermax(ino,n1)=aj*delt+to !Corner max. at IVL
maxmin(ino)=nout(k)
ivx(ino,n1)=ivl !Maximum in millivolts

```

```

jp=j2-2
go to 340

```

! CROSS

```

360  IF (N1.GE.100) GO TO 250
      IF ((j2-jp).le.6) go to 340
      ic1=ivm-iv1      !See if change is significant relative
      ic2=ivm-ivlo
      IF ((ic1.le.1).and.(ic2.le.ic1)) go to 340 !Change insignificant
      !Real MAXIMUM found
      N1=N1+1
      nx(ino)=n1      !Count number of maxima for this .OUT
      AJ=J2-2
      DERMAX(INO,N1)=AJ*DELT+T0+DELT*(IVL-IVH)/(IVH+IVL-2*IVM)/2.
      MAXMIN(INO)=NOUT(K)
      ivx(ino,n1)=IVM      !Maximum in millivolts
      XTD(INO)=TOFF(K)
      jp=j2-1
      GO TO 340

370  IF (N2.GE.100) GO TO 250
      IF ((j2-jp).le.6) go to 340
      ic1=iv1-ivm      !See if change in consecutive values signific
      ic2=ivlo-ivm
      IF ((ic1.le.1).and.(ic2.le.ic1)) go to 340 !Change is insignificant
      !real MINIMUM found
      N2=N2+1
      nn(ino)=n2      !Count number minima for this .OUT
      AJ=J2-2
      DERMIN(INO,N2)=AJ*DELT+T0+DELT*(IVL-IVH)/(IVH+IVL-2*IVM)/2.
      MAXMIN(INO)=NOUT(K)
      ivn(ino,n2)=ivm      !Minimum in millivolts

      jp=j2-1

340  ivlo=iv1
      IVL=IVM
      IVM=IVH

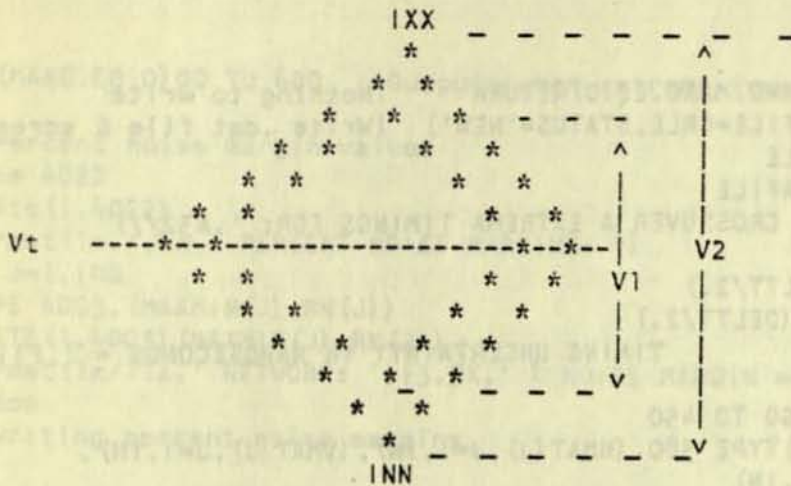
350  CONTINUE

250  IF (N.GT.MAX) MAX=N
      IF (N1.GT.MAXO) MAXO=N1
      IF (N2.GT.MAXO) MAXO=N2
      IF (INO.GE.5) GO TO 270

C!

```

Noise margin calculations, based on EYE pattern:



The percent noise margin, N, is defined as:

$$N = (V1/V2) * 100$$

V1 is the difference between the minimum of all the maxima, INX, and the maximum of all the minima, IXN: $V1 = INX - IXN$.

V2 is the difference between the maximum of all the maxima, IXX, and the minimum of all the minima, INN: $V2 = IXX - INN$.

Although it is easier to interpret N with an EYE pattern, the calculation can be run wherever there has been an extremum calculation.

```

do j=1,ino      !Find noise margin wherever DV specified
  ixm=ivx(j,1) !Initialize ixm, ixm, ixn, ixn
  inn=ivn(j,1)
  ixn=ivn(j,1)
  ixm=ivx(j,1)
do i=2,nx(j)   !Find max & min of maxima values
  ixm=jmax0(ixm,ivx(j,i))
  ixn=jmin0(ixn,ivx(j,i))
enddo
do i=2,nn(j)   !Find max & min of minima values
  ixn=jmax0(ixn,ivn(j,i))
  inn=jmin0(inn,ivn(j,i))
enddo
nd V1 and V2 for the .OUT statement corresponding
this value of j
v1=ixn-ixn      !V1
v2=ixm-inn     !V2
rn(j)=100.*(v1/v2) !Per cent noise margin
enddo
nd of noise margin calculations

```

```

50      CONTINUE
270     IF (MAX.EQ.0.AND.MAX0.EQ.0) RETURN      !Nothing to write
        OPEN (UNIT=1,FILE=CRLE,STATUS='NEW') !Write .dat file & screen
        TYPE 880,AFILE
        WRITE (1,880) AFILE
880     FORMAT (1X/' CROSSOVER & EXTREMA TIMINGS FOR: '.A32//)

        TYPE 890, (DELTT/2.)
        WRITE (1,890) (DELTT/2.)
890     FORMAT (1X,'          TIMING UNCERTAINTY IN NANoseconds = '.F11.3/)

        IF (MAX.EQ.0) GO TO 450
        if (neye.ne.1) TYPE 380, (NMAT (J), J=1, IN), (VMAT (J), J=1, IN),
        1 (TD (J), J=1, IN)
380     FORMAT (1X/' CROSSOVER TIMING IN NANoseconds, '//
        1 ' NETWORK', <IN> (7X, 13, 2X) /
        2 ' VOLTAGE ', <IN> (F11.3, 1X) /
        3 ' DELAY   ', <IN> (F11.3, 1X) /
        WRITE (1, 380) (NMAT (J), J=1, IN), (VMAT (J), J=1, IN),
        1 (TD (J), J=1, IN)
        DO I=1, MAX
        if (neye.ne.1) TYPE 400, (TMAT (J, I), J=1, IN)
        WRITE (1, 400) (TMAT (J, I), J=1, IN)
400     FORMAT (9X, <IN> (F11.3, 1X))
        ENDDO

```

C Timing jitter calculation - executed only when EYE pattern requested

```

        IF (NEYE.NE.1) GO TO 450
        NM=JMOD (MAX, 2)
        IC1=MAX/2
        AX1=TMAT (1, 2)
        AN1=AX1
        AX2=TMAT (1, 1)
        AN2=AX2
        DO I=1, IC1
        IC2=1+2*I
        AX1=AMAX1 (TMAT (1, 2*I), AX1)
        AN1=AMIN1 (TMAT (1, 2*I), AN1)
        IF ((NM.EQ.0).AND.(I.EQ.IC1)) GO TO 4002
        AX2=AMAX1 (TMAT (1, IC2), AX2)
        AN2=AMIN1 (TMAT (1, IC2), AN2)
4002     ENDDO
        TJ1=AX1-AN1
        TJ2=AX2-AN2
        TYPE 4001, (TJ1, TJ2)
        WRITE (1, 4001) (TJ1, TJ2)
4001     FORMAT (1X, /, 9X, 'TIMING JITTER = '.F8.4, 4X, F8.4, ' NS')
        !

```

! CROSS

50 IF (MAX0.EQ.0) GO TO 600 !Outputs when extrema found

Write Percent noise margin values

type 4022

write (1,4022)

022 format (1x/,6x,' PERCENT NOISE MARGINS: ')

DO J=1,INO

TYPE 4003, (MAXMIN(J),RN(J))

WRITE (1,4003) (MAXMIN(J),RN(J))

003 format (1x/11x,' NETWORK: ',13,5X,' % NOISE MARGIN = ',F9.4)

enddo

Done writing percent noise margins

Write extrema values and locations

INO2=2*INO

if (neye.ne.1) TYPE 500, (MAXMIN(J),J=1,INO), (XTD(J),J=1,INO)

000 FORMAT (1X/' MAXIMA AND MINIMA TIMING IN NANoseconds'//

1 ' NETWORK ',<INO>(7X,13,14X)/

2 ' DELAY ',<INO>(F11.3,13X)/

3 ' ',<INO>(' MAXIMA MINIMA ')/

WRITE (1,500) (MAXMIN(J),J=1,INO), (XTD(J),J=1,INO)

DO I=1,MAX0

if (neye.ne.1) TYPE 550, (DERMAX(J,I),DERMIN(J,I),J=1,INO)

WRITE (1,550) (DERMAX(J,I),DERMIN(J,I),J=1,INO)

500 FORMAT (9X,<INO2>(F11.3,1X))

ENDDO

if (neye.ne.1) type 700

WRITE (1,700)

000 FORMAT (1X/' EXTREMA VALUES IN MILLIVOLTS '//

do j=1,ino

if (neye.ne.1) TYPE 750, (MAXMIN(J))

WRITE (1,750) (MAXMIN(J))

000 FORMAT (1X/' NETWORK: ',13)

do i=1,maxo

if (neye.ne.1) TYPE 760,1,IVX(J,I),IVN(J,I)

WRITE (1,760) 1,IVX(J,I),IVN(J,I)

000 FORMAT (9X,14,' MAX = ',16,5X,' MIN = ',16)

enddo

enddo

type *,' '

CLOSE (UNIT=1)

RETURN

END

! DELTA

```

SUBROUTINE DELTA (N,W)
CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC
C SOLVES FOR THE A,B,C,D PARAMETERS FOR THE PI CIRCUIT
C FOR THE Nth NETWORK ELEMENT AT FREQUENCY W (TYPE 3 CIRCUIT)
C 12/11/85
CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC
COMPLEX*16 ABCD (500,2,2),ZERO
DOUBLE PRECISION P12
COMMON/NET/RLC (500,20),ABCD,P12,ZERO

G1=RLC (N,1) ! G1
G2=RLC (N,2) ! G2
BC1=W*RLC (N,3) ! w*C1
BC2=W*RLC (N,4) ! w*C2
XL=W*RLC (N,5) ! w*L
R=RLC (N,6)
IF (RLC (N,9) .EQ.0.0) THEN ! RP default to infinity
ABCD (N,1,1)=DCMLPX ( (1.+G2*R-BC2*XL), (XL*G2+R*BC2) ) !A
ABCD (N,1,2)=DCMLPX (R,XL) !B
ABCD (N,2,1)=DCMLPX ( (G1+G2+R* (G1*G2-BC1*BC2) -XL* (G1*BC2+G2*BC1)),
1 (BC1+BC2+R* (G1*BC2+G2*BC1)+XL* (G1*G2-BC1*BC2)) ) !C
ABCD (N,2,2)=DCMLPX ( (1.+G1*R-BC1*XL), (XL*G1+R*BC1) ) !D
ELSE
RP=RLC (N,9) ! Finite RP case
D= (R+RP)**2+XL**2
A=RP* (R* (R+RP)+XL**2)/D
B=XL*RP**2/D
ABCD (N,1,1)=DCMLPX ( (1.+G2*A-BC2*B), (G2*B+BC2*A) ) !A
ABCD (N,1,2)=DCMLPX (A,B) !B
ABCD (N,2,1)=DCMLPX ( (G1+G2+A* (G1*G2-BC1*BC2) -B* (G1*BC2+G2*BC1)),
1 (BC1+BC2+A* (G1*BC2+G2*BC1)+B* (G1*G2-BC1*BC2)) ) !C
ABCD (N,2,2)=DCMLPX ( (1.+G1*A-BC1*B), (G1*B+BC1*A) ) !D
ENDIF
RETURN
END
    
```

! DIGIT

FUNCTION DIGIT(M, ITEM, IERR)

CC

CONVERTS A CHARACTER STRING TO A NUMBER

C

C

C

CC

CHARACTER*1 ITEM(130), INTEG(15), N

DIMENSION TEMP(130)

DATA INTEG/'1','2','3','4','5','6','7','8','9','0','.','-',

' ','K','M','U'/'

DIGIT=0.

K=0

!Decimal Point Location

PLUS=1.0

!Sign and Exponent

IF (ITEM(M) .NE. '-') GO TO 10

PLUS=-1.0

M=M+1

IF (ITEM(M) .EQ. '+') M=M+1

!Ignore +

DO 50 I=M, 130

N=ITEM(I)

DO J=1, 15

!Match Character

IF (N.EQ. INTEG(J)) GO TO 30

ENDDO

IERR=1

!Error, no match

RETURN

TEMP(I)=0.

IF (J.EQ.11.AND.K.NE.0) GO TO 20

IF (J.EQ.11) K=1 !A period

IF (J.EQ.12) GO TO 60 !A blank

IF (J.LT.10) TEMP(I)=J

IF (J.EQ.13) PLUS=PLUS*1.E3 !K, M, and U exponents

IF (J.EQ.14) PLUS=PLUS*1.E-3

IF (J.EQ.15) PLUS=PLUS*1.E-6

CONTINUE

IF (K.EQ.0) K=1

IF (ABS(PLUS) .EQ. 1.0) THEN

IF (K.EQ.0) K=1

L=I-1

ELSE

IF (K.EQ.0) K=I-1

L=I-2

ENDIF

DO 70 J=M, L

IF (J.EQ.K) GO TO 70

A=10.** (K-J)

IF (J.LT.K) A=0.1*A

DIGIT=DIGIT+TEMP(J)*A

CONTINUE

M=I+1

DIGIT=PLUS*DIGIT

RETURN

END

!

! EAMP
! FIBER

```

SUBROUTINE EAMP (N,W)
CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC
C RETURNS THE A, B, C, D PARAMETERS FOR A
C A VOLTAGE DEPENDENT VOLTAGE SOURCE WITH EXTERNAL RESISTANCE
C AND A DEFINED BANDWIDTH (TYPE 6 CIRCUIT)
C 1/20/86
CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC
COMPLEX*16 ABCD (500,2,2),ZERO,D
DOUBLE PRECISION PI2
COMMON/NET/RLC (500,20),ABCD,PI2,ZERO

WO=RLC (N,16)
D=DCMPLX (1.0,W/WO) /RLC (N,1)
ABCD (N,1,1)=D
ABCD (N,1,2)=RLC (N,2)*D
ABCD (N,2,1)=1.D-9*D/RLC (N,2)
ABCD (N,2,2)=1.D-9*D

RETURN
END

```

```

SUBROUTINE FIBER (N,W)
CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC
C OPTICAL FIBER MODEL - Returns A, B, C, D parameters
C (TYPE 7 CIRCUIT)
C 2/14/86
CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC
COMPLEX*16 ABCD (500,2,2),ZERO,P
DOUBLE PRECISION PI2,AO,AW,WTC,WTM,H,PHI
COMMON/NET/RLC (500,20),ABCD,PI2,ZERO

AO=RLC (N,16)
WTC=W*RLC (N,17)
AW=RLC (N,18)
WTM=W*RLC (N,19)
H=AO+WTC**2+WTM**2
IF (H.GT.20.) H=20.
P=DEXP (H)
IF (AW.EQ.0.0) GO TO 100 !Simplify if no wavelength dependent losses
PHI=2.*AW*WTC
P=P*DCMPLX (DCOS (PHI),DSIN (PHI))
100 ABCD (N,1,1)=P
ABCD (N,1,2)=ZERO
ABCD (N,2,1)=ZERO
ABCD (N,2,2)=P

RETURN
END

```

!

FILTER

SUBROUTINE FILTER(N,W)

DELAY LINE FILTER MODEL
(TYPE 12 CIRCUIT)

4/11/86

COMPLEX*16 ABCD(500,2,2),ZERO,Q
DOUBLE PRECISION P12,P,X
COMMON/NET/RLC(500,20),ABCD,P12,ZERO

X=W*RLC(N,1)
M=RLC(N,2)
P=(DCOS(X))**M
IF(DABS(P).LT.1.D-20)P=1.D-20
Q=1./P
ABCD(N,1,1)=Q
ABCD(N,1,2)=Q*RLC(N,3)
ABCD(N,2,1)=Q*RLC(N,4)
ABCD(N,2,2)=ZERO

RETURN
END

1 GEN

FUNCTION GEN (N,W)

```

CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC
C RETURNS THE VOLTAGE GENERATOR FREQUENCY COMPONENT
C FOR THE Nth NETWORK ELEMENT AT FREQUENCY W (TYPE 1 CIRCUIT)
C 12/11/85

```

```

CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC
COMPLEX*16 ABCD (500,2,2),ZERO,DNC,GEN,DF
DOUBLE PRECISION DN,DO,AP,AR,AF,PHI,PER,PI2,BETA,C
COMMON/NET/RLC (500,20),ABCD,PI2,ZERO
COMMON/TIME/PER

```

```

PW=RLC (N,1)           !Pulse width
AMP=RLC (N,4)          !Amplitude
WO=RLC (N,16)          !Gen Bandwidth rad/s
TRM=RLC (N,20)         !0% to 100% risetime
AR=0.5*TRM*W
AP=0.5*W*PW
C=2.*AMP
DN=C*PW/PER
IF (W.EQ.0.0) DN=0.5*DN+RLC (N,7)           !DC Component
IF (AP.GT.1.E-5) DN=DN*DSIN (AP) /AP
IF (AR.GT.1.E-5) DN=DN*DSIN (AR) /AR

```

C GENERATOR DELAY

```

PHI=-W*RLC (N,5)
DNC=DN*DCMLX (DCOS (PHI),DSIN (PHI))       !Delay

```

C ASSYMETRICAL TRAPEZOID

```

TFM=RLC (N,19)
AF=0.5*TFM*W
IF (TFM.EQ.TRM.OR.AF.EQ.0.0.OR.AR.EQ.0.0) GO TO 100
BETA=PHI-AP+PI2/4.0
DO=AF
IF (AR.NE.0.0) DO=DO*DSIN (AR) /AR
DF=C*(DSIN (AF)-DO)*DCMLX (DCOS (BETA),DSIN (BETA)) / (AF*W*PER)
DNC=DNC+DF

```

C GENERATOR LOW PASS FILTER

```

100 GEN=DNC/DCMLX (1.0,(W/WO))
RETURN
END

```

! GRAPHICS

SUBROUTINE GRAPHICS (AFILE,BFILE,DFILE,NF,TDIS,TOFF,DC,FS,VOFF,
1 NEYE,ID)

CC

WRITES TO THE OUTPUT FILES 12/29/86

- AFILE = name of simulation model file; used in labelling
- BFILE = name of graphics output file
- DFILE = name of data output file
- NF = number of terms to be graphed
- TDIS = horizontal axis scale
- TOFF = horizontal axis offset for time domain
- DC = vector of vertical offsets for various outputs, maximum of 5 allowed
- FS = vertical axis scale
- VOFF = vertical axis offset
- NEYE = 1 for EYE pattern; otherwise 0
- ID = 1 for frequency domain output; otherwise 0

CC

CHARACTER*32 AFILE,BFILE,DFILE
 COMPLEX*16 EW(5,32768)
 DOUBLE PRECISION PER,P2
 INTEGER ESC ! This is the graphics escape character
 DIMENSION TOFF(5),DC(5),ORD(11),OUTPUT(350,11)
 COMMON/MISC/KIND(500),RGEN,NOUT(5),MOUT,KD(5),iff
 COMMON/TIME/PER
 COMMON/WAVES/EW,EOUT(5,32768)
 DATA ESC/027/ ! This is its octal equivalent
 P2=6.28318530717985
 FWD=1./PER
 AFS=FS
 AVOFF=VOFF !Vertical offset
 IF (FS.EQ.0.0) AFS=2.0 !Vertical scale factor for time
 AMP=1./AFS
 AD=450.*AMP*AVOFF+20. !Vertical offset

TINC=TDIS/350. !Horizontal display time increment
 TG=35.*TINC !Horizontal axis labeling increment
 TGO=TG !Horizontal axis initial offset
 TS=TG !Waveform time offset

IF (NEYE.EQ.0) GO TO 10
 TINC=2.*TINC !Eye pattern values
 TG=35.*TINC
 TGO=175.*TINC
 TS=87.5*TINC
 ANF=NF !Number of time samples

!

I GRAPHICS

```

DELT=PER/ANF           !Time delay per waveform sample
IF (ID.EQ.1) DELT=FWO  !Freq delay per waveform output
Y=TINC/DELT           !Ratio of display sample to waveform sample
TO=1.-TS/DELT-Y       !Display offset in number of waveform samples
DO I=1,100
IF (ABS(TO).LT.ANF) GO TO 20
TO=TO+ANF
ENDDO

20  OPEN (UNIT=21,FILE=BFILE,STATUS='NEW',CARRIAGECONTROL='LIST')
    WRITE (21,90) ESC           !CLEAR SCREEN
90  FORMAT (1X,A1,'[2J')
    WRITE (21,100) ESC         !ENTER GRAPHICS
100 FORMAT (1X,A1,'Pps (a[0,479][767,0])')

C *****
C *****DRAW GRID AND TICK MARKS*****
C *****

110 WRITE (21,110) !MAKE GRID LIGHTER THAN THE WAVEFORM PLOT
    FORMAT (1X,'w(i2)')
    J=-20
    K=-25
    DO 130 I=1,11 !DRAW GRID LINES
        J=J+70
        K=K+45
        WRITE (21,120) J,J,K,K
120  FORMAT (1X,'p['13',20]v['13',470]p[50,'13']v[750,'13']')
130  CONTINUE
    J=50
    K=20
    DO 140 I=1,50 !DRAW TICK MARKS
        J=J+14
        K=K+9
        WRITE (21,150) J,J,K,K
150  FORMAT (1X,'p['13',30]v['13',20]p[50,'13']v[60,'13']')
140  CONTINUE

C *****
C *****PLOT WAVEFORMS HERE*****
C *****
C *****

190 WRITE (21,190)
    FORMAT (1X,'w(i3)')

    DO 300 K=1,MOUT
    EOFF=DC(K)
    TC=0
    T=TO+TOFF(K)/DELT
    NP=0
310  !Return to normal intensity

    !EYE pattern sample counter
    !Total time offset in number of waveform samples
    !Loop to here for EYE pattern

210 WRITE (21,210)
    FORMAT (1X,'p[48,0]')
    TOUT=-TINC-TG

```

! GRAPHICS

```

DO 400 I=1,350
IF (K.EQ.1.) OUTPUT (I,1)=I*TINC+TOUT
T=Y+T                                !Increment sample number
T=BRACKET (T,ANF)
NT=JINT (T)                            !Integer
DT=T-NT
NT1=NT+I
IF (NT1.GT.NF) NT1=NT1-NF !Recycle if NT2 out of range (too large)
TC=TC+Y                                !Increment EYE pattern sample counter
EO=(1.-DT)*EOUT (K,NT)+DT*EOUT (K,NT1) !Interpolate
OUTPUT (I,K+1)=EO+EOFF
OUT=AMP*(EO+EOFF)*450.+A0
IF (I.GT.1) GO TO 410
WRITE (21,200) OUT                      !First point is a point, not a line
FORMAT (1X,'p[+2,'G']V[]')
GO TO 400
IF (OUT.GE.0.0) WRITE (21,420) OUT
FORMAT (1X,'V[+2,'G']')
IF (OUT.LT.0.0) WRITE (21,430)
FORMAT (1X,'P[+2,0]')                  !Write a blank for negative display locations
CONTINUE
WRITE (21,500)
FORMAT (1X,'[] (e)')
IF (NEYE.EQ.0) GO TO 300
IF (TC.LT.ANF) GO TO 310 !Finish for EYE pattern.
CONTINUE

```

```

*****
*****LABEL AXES HERE*****
*****

```

```

DO I=1,11
ORD (I)=I-1
ORD (I)=0.1*AFS*ORD (I)-VOFF
ENDDO
WRITE (21,621) ORD
FORMAT (1X,'p[1,28]t(s1,i0)'''F7.2''''/
1 1X,'p[1,73]t(s1,i0)'''F7.2''''/
2 1X,'p[1,118]t(s1,i0)'''F7.2''''/
3 1X,'p[1,163]t(s1,i0)'''F7.2''''/
4 1X,'p[1,208]t(s1,i0)'''F7.2''''/
5 1X,'p[1,253]t(s1,i0)'''F7.2''''/
6 1X,'p[1,298]t(s1,i0)'''F7.2''''/
7 1X,'p[1,343]t(s1,i0)'''F7.2''''/
8 1X,'p[1,388]t(s1,i0)'''F7.2''''/
9 1X,'p[1,433]t(s1,i0)'''F7.2''''/
1 1X,'p[1,478]t(s1,i0)'''F7.2''''/
LABEL ABSCISSA AXIS

```


1 GRAPHICS

```

DO I=1,6
  MM=25+137*(I-1)
  IF (I.EQ.6) MM=714
  TG2=2.*TG*(I-1)-TGO
  if (id.eq.1) tg2=1000.*tg2          !Convert freq to MHz
  WRITE (21,700) MM,TG2
700  FORMAT (1X,'p['13',15]t(s1,i0)'''F7.2''')
  ENDDO
  IF (ID.NE.1) WRITE (21,710) AFILE,ESC          ! PLOT TITLE
710  FORMAT (1X,'p[230,479]t(s0,i0)'''AMPLITUDE VS TIME IN NS.'',
1 " INPUT FILE:","'A32''', 's(a[0,0][767,479])',A1,'\') !EXIT GRAP
  IF (ID.EQ.1) WRITE (21,711) AFILE,ESC
711  FORMAT (1X,'p[230,479]t(s0,i0)'''AMPLITUDE VS FREQ IN MHZ.'',
1 " INPUT FILE:","'A32''', 's(a[0,0][767,479])',A1,'\') !EXIT GRAP

CLOSE (UNIT=21)
! IF (ID.EQ.0) RETURN          !if ID=1, Create an output file DFILE
IN=MOUT+1                    !with the graphed data
OPEN (UNIT=21,FILE=DFILE,STATUS='NEW')
DO I=1,350
WRITE (21,1000) (OUTPUT (I,J),J=1,IN)
1000  FORMAT (1X,F20.3,<MOUT>-(1X,F20.3))
ENDDO
CLOSE (UNIT=21)
RETURN
END

```

!

! GSIN

FUNCTION GSIN(N,W)

RETURNS THE SINEWAVE GENERATOR FREQUENCY COMPONENT
FOR THE Nth NETWORK ELEMENT AT FREQUENCY W (TYPE 8 CIRCUIT)
12/16/86

COMPLEX*16 ABCD(500,2,2),ZERO,GSIN
DOUBLE PRECISION DN,PHI,PI2
COMMON/NET/RLC(500,20),ABCD,PI2,ZERO
COMMON/TIME/PER

WO=RLC(N,16) !Sinewave frequency, 2*pi*F in GRad/s
DN=0.0
IF(W.EQ.0.0) DN=RLC(N,3) !DC Component
IF(W.EQ.WO) DN=RLC(N,2) !Amplitude

GENERATOR DELAY

PHI=-W*RLC(N,4) !Delay
GSIN=DN*DCMLX(DCOS(PHI),DSIN(PHI))

RETURN
END

1 HPF
1 IAMP

```

SUBROUTINE HPF (N,W)
CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC
C   SOLVES FOR THE A,B,C,D PARAMETERS FOR THE HPF CIRCUIT
C   FOR THE Nth NETWORK ELEMENT AT FREQUENCY W (TYPE 10 CIRCUIT)
C   11/3/86 new B coeff
CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC
      COMPLEX*16 ABCD (500,2,2) ,ZERO
      DOUBLE PRECISION P12,GL,GC,D,AR,A1,B
      COMMON/NET/RLC (500,20) ,ABCD,P12,ZERO

      GL=RLC (N,20)    !L/R
      GC=RLC (N,19)   !1/RC
      GLW=W*GL
      D=1./ (1.+GLW**2)
      AR=1.-D*GL*GC
      IF (W.GE.1.E-5) THEN
         A1=-D*GC/W
         B=-1./ (W*RLC (N,1))
      ELSE
         A1=-D*GC*1.E+5
         B=-1.E+5/RLC (N,1)
      ENDIF
      ABCD (N,1,1)=DCMPLX (AR,A1)
      ABCD (N,1,2)=B*DCMPLX (0.0,1.0)
      ABCD (N,2,1)=RLC (N,3) /DCMPLX (1.,GLW)
      ABCD (N,2,2)=DCMPLX (1.0,0.0)
      RETURN
      END

```

IA
IB
IC
ID

```

SUBROUTINE IAMP (N,W)
CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC
C   A VOLTAGE DEPENDENT CURRENT SOURCE WITH EXTERNAL RESISTANCE
C   AND A DEFINED BANDWIDTH (TYPE 9 CIRCUIT)
C   1/20/86
CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC
      COMPLEX*16 ABCD (500,2,2) ,ZERO,D
      DOUBLE PRECISION P12
      COMMON/NET/RLC (500,20) ,ABCD,P12,ZERO

      W0=RLC (N,16)
      D=DCMPLX (1.0,W/W0) /RLC (N,1)
      ABCD (N,1,1)=RLC (N,3) *D
      ABCD (N,1,2)=D
      ABCD (N,2,1)=1.D-6*RLC (N,3) *D
      ABCD (N,2,2)=1.D-6*D

      RETURN
      END

```

! INFILTER

SUBROUTINE INFILTER(NVAR,K,ITEM)

CC

Special Handling for FILTER Network, places the implied "1" after
 the Sine, Cosine, or Gaussian statements

CHARACTER*1 ITEM(130)

```

IF (NVAR.LT.9) RETURN
DO I=K,130
IF (ITEM(I).EQ.' ') GO TO 10
ENDDO
ITEM(I-1)='1'
ITEM(I-2)=' '
K=I-3
    
```

RETURN
 END

INPUT

SUBROUTINE INPUT(AFIL,BFIL,DFIL,FGL,FDL,PGL,PDL,NC,TDIS,TOFF,DC,
1 CRL,FS,VOFF,NEYE,ID,nff,npp)

CC

C DATA ENTRY SUBROUTINE 1/28/87 C

C INPUT reads in the simulation model, interprets network C
C subsections and writes the subsection circuit parameters C
C into array RLC, and calls subroutine CONTROL to interpret C
C simulation model control statements. Input also generates C
C the names for the output files, based on the name of the C
C simulation model. INPUT also passes the control pointers: C
C NEYE, NFF, and NPP to the main routine. C

- C AFIL = file containing simulation model C
- C BFIL = time domain graph output file C
- C DFIL = time domain data output file C
- C FGL = frequency domain graph output file C
- C FDL = frequency domain data output file C
- C PGL = freq. vs. phase graphics output file C
- C PDL = freq. vs phase data output file C
- C NC = number of network sections in simulation model C
- C TDIS = horizontal display scale for time domain C
- C TOFF = vector of horizontal shifts for time displays C
- C DC = vector of vertical shifts for displays C
- C CRL = file containing locations of voltage crossings C
- C FS = vertical axis scale for time domain output C
- C VOFF = vertical offset for time domain graphs C
- C NEYE = 1 if EYE pattern output requested, else=0 C
- C ID = 1 if frequency domain outputs requested, else=0 C
- C NFF = 1 if frequency domain outputs requested C
- C NPP = 1 if phase shift outputs requested C

CC

CHARACTER*1 ITEM(130),SPA,TAB,ICDAT
 CHARACTER*2 NDATA(13,20),ND
 CHARACTER*3 NAME(20),NOW
 CHARACTER*32 AFIL,BFIL,DFIL,FGL,FDL,PGL,PDL,CRL
 COMPLEX*16 ABCD(500,2,2)
 DOUBLE PRECISION PER,PI2
 DIMENSION TOFF(5),DC(5)
 COMMON/NET/RLC(500,20),ABCD,PI2
 COMMON/MISC/KIND(500),RGEN,NOUT(5),MOUT,KD(5),iff

integer Status, size
 integer Str\$trim
 integer time_graph_size, time_data_size
 integer freq_graph_size, freq_data_size
 integer phase_graph_size, phase_data_size, cross_size

! INPUT

DATA NAME/

1	2	3	4	5	6	7	8	9	10	
1	'GEN'	'TEE'	'DEL'	'LIN'	'STU'	'EAM'	'FIB'	'SIN'	'IAM'	'HPF'
11	12	13	14	15	16	17	18	19	20	
2	'XFO'	'FIL'	'SHU'	'SER'	'BPF'	'SRC'	'	'	'	'/'

DATA NDATA/

GEN	1	2	3	4	5	6	7	8	9	10	11	12	13
1	'PW'	'TR'	'PE'	'AM'	'DE'	'BW'	'DC'	'R'	'TF'	'MO'	'MO'	'MO'	'MO'
TEE	1	2	3	4	5	6	7	8	9	10	11	12	13
2	'R1'	'R2'	'G'	'L1'	'L2'	'C'	'F1'	'F2'	'R'	'R'	'R'	'R'	'R'
DELTA	1	2	3	4	5	6	7	8	9	10	11	12	13
3	'G1'	'G2'	'C1'	'C2'	'L'	'R'	'R1'	'R2'	'RP'	'RP'	'RP'	'RP'	'RP'
LINE	1	2	3	4	5	6	7	8	9	10	11	12	13
4	'ZO'	'TD'	'X'	'R'	'TC'	'D'	'K1'	'K2'	'WC'	'Q'	'F1'	'F2'	'F3'
STUB	1	2	3	4	5	6	7	8	9	10	11	12	13
5	'ZO'	'TD'	'X'	'R'	'TC'	'D'	'K1'	'K2'	'GL'	'RL'	'CL'	'WC'	'WC'
EAMP	1	2	3	4	5	6	7	8	9	10	11	12	13
6	'GA'	'R'	'BW'	'BW'	'BW'	'BW'	'BW'	'BW'	'BW'	'BW'	'BW'	'BW'	'BW'
FIBER	1	2	3	4	5	6	7	8	9	10	11	12	13
7	'AD'	'AW'	'FW'	'F1'	'D'	'QM'	'QC'	'X'	'LC'	'LO'	'SO'	'SO'	'SO'
SIN	1	2	3	4	5	6	7	8	9	10	11	12	13
8	'F'	'AM'	'DC'	'DE'	'R'	'R'	'R'	'R'	'R'	'R'	'R'	'R'	'R'
IAMP	1	2	3	4	5	6	7	8	9	10	11	12	13
9	'GA'	'R'	'G'	'BW'	'BW'	'BW'	'BW'	'BW'	'BW'	'BW'	'BW'	'BW'	'BW'
HPF	1	2	3	4	5	6	7	8	9	10	11	12	13
1	'C'	'L'	'G'	'R'	'R'	'R'	'R'	'R'	'R'	'R'	'R'	'R'	'R'
XFORMER	1	2	3	4	5	6	7	8	9	10	11	12	13
2	'L1'	'L2'	'M'	'K'	'N'	'N'	'N'	'N'	'N'	'N'	'N'	'N'	'N'
FILTER	1	2	3	4	5	6	7	8	9	10	11	12	13
3	'TO'	'N'	'R2'	'G1'	'R1'	'FO'	'FO'	'FO'	'S1'	'CO'	'GA'	'GA'	'GA'
SHUNT	1	2	3	4	5	6	7	8	9	10	11	12	13
4	'AO'	'A1'	'A2'	'A3'	'A4'	'B0'	'B1'	'B2'	'B3'	'B4'	'K'	'K'	'K'
SERIES	1	2	3	4	5	6	7	8	9	10	11	12	13
5	'AO'	'A1'	'A2'	'A3'	'A4'	'B0'	'B1'	'B2'	'B3'	'B4'	'K'	'K'	'K'
BPF	1	2	3	4	5	6	7	8	9	10	11	12	13
6	'F1'	'F2'	'S1'	'S2'	'K'	'PH'	'PH'	'PH'	'PH'	'PH'	'PH'	'PH'	'PH'
SRC	1	2	3	4	5	6	7	8	9	10	11	12	13
7	'EF'	'PW'	'PE'	'AM'	'DE'	'NT'	'DT'	'TY'	'R'	'R'	'R'	'R'	'R'
UNASS	1	2	3	4	5	6	7	8	9	10	11	12	13
8	'	'	'	'	'	'	'	'	'	'	'	'	'
UNASS	1	2	3	4	5	6	7	8	9	10	11	12	13
9	'	'	'	'	'	'	'	'	'	'	'	'	'
UNASS	1	2	3	4	5	6	7	8	9	10	11	12	13
1	'	'	'	'	'	'	'	'	'	'	'	'	'
UNASS	1	2	3	4	5	6	7	8	9	10	11	12	13
2	'	'	'	'	'	'	'	'	'	'	'	'	'/'

DATA SPA/ ' ',TAB/ ' ',IERR/O/,MOUT/O/,KD/5#0/

NEND=0 !End of the model pointer
 NEYE=0 !EYE pattern pointer
 nff=0 !frequency domain output pointer
 npp=0 !Phase angle output pointer
 NC=0 !Number of circuit subsections counter
 NKIND=16 !Number different circuit subsection types
 ID=1 !Pointer for freq. domain outputs in GRAPHICS

```

TYPE 20
  FORMAT(' Enter the input file name: ',S)
  ACCEPT 40,AFIL
40
  FORMAT(A32)

C
  Code to generate file names for output files
    ! trim out extra spaces
  status = str$trim( AFIL , AFIL , size )
  If (.NOT. Status) Call lib$signal( %val ( Status ))
    ! make filenames by adding file type
  BFIL = AFIL (1:size) // 'tg' ! filetype
  DFIL = AFIL (1:size) // 'td'
  FDL  = AFIL (1:size) // 'fd'
  FGL  = AFIL (1:size) // 'fg'
  PGL  = AFIL (1:size) // 'pg'
  PDL  = afil (1:size) // 'pd'
  CRL  = afil (1:size) // 'dat'

  ! get string size
  status = str$trim( BFIL,
2          BFIL, time_graph_size )
  status = str$trim( DFIL,
2          DFIL, time_data_size )
  status = str$trim( FGL,
2          FGL, freq_graph_size )
  status = str$trim( FDL,
2          FDL, freq_data_size )
  status = str$trim( PGL,
2          PGL, phase_graph_size )
  status = str$trim( PDL,
2          PDL, phase_data_size )
  status = str$trim( CRL,
2          CRL, cross_size )

TYPE *, '*****'
OPEN(UNIT=1,FILE=AFIL,STATUS='OLD')

DO 500 I=1,1000 !Will read 1000 statements in simulation model
DO J=1,130
ITEM(J)=SPA !Initialize input array
ENDDO
120 READ(1,120)(ITEM(K),K=1,130) !Read in an input line
FORMAT(130A1)
TYPE 125,(ITEM(J),J=1,80) !Type it back
125 FORMAT(1X,80A1)
DO J=1,130
IT=ICHAR(ITEM(J))
IF(IT.GE.97.AND.IT.LE.122)IT=IT-32 !Capitalize everything
ITEM(J)=CHAR(IT)
IF(ITEM(J).EQ.','.OR.ITEM(J).EQ.'='.OR.ITEM(J).EQ.TAB)
1 ITEM(J)=SPA !Replace , = or tab with space character
ENDDO
!

```

! INPUT

```

DO 140 M=1,130
IF (ITEM(M).EQ.SPA) GO TO 140      !Look for special first characters
IF (ITEM(M).EQ.'!') GO TO 500     !Space
IF (ITEM(M).EQ.'.') GO TO 350     !Comment
IF (ITEM(M).EQ.'+') GO TO 190     !Control Line
GO TO 160                          !Continuation Line
CONTINUE                            !Must be a Circuit
GO TO 500
140
160 NOW(1:1)=ITEM(M)
NOW(2:2)=ITEM(M+1)
NOW(3:3)=ITEM(M+2)
DO LT=1,NKIND
IF (NOW.EQ.NAME(LT)) GO TO 170     !Match with circuit NAME
ENDDO
TYPE *, '***** UNDEFINED CIRCUIT NAME *****'
IERR=1
GO TO 500
170 NC=NC+1
IF (NC.LE.500) GO TO 175           !Only 500 Networks allowed
TYPE *, '***** TOO MANY NETWORKS SPECIFIED *****'
STOP
175 KIND(NC)=LT
M=M+2
M=M+1
90 DO K=M,130
IF (ITEM(K).EQ.'!') GO TO 500     !Next line if Comment
IF (ITEM(K).EQ.SPA) GO TO 185     !Look for space character
ENDDO
185 M=K

DO 300 J=1,13
DO 310 K=M,130
IF (ITEM(K).EQ.'!') GO TO 500     !Next line if Comment
IF (ITEM(K).EQ.SPA) GO TO 310     !Passover space characters
ND(1:1)=ITEM(K)
ND(2:2)=ITEM(K+1)
DO NVAR=1,13
IF (ND.EQ.NDATA(NVAR,LT)) GO TO 195 !Match data variable names
ENDDO

TYPE *, '***** UNDEFINED VARIABLE NAME *****'
IERR=1
GO TO 500
CONTINUE
IF (K.GE.130) GO TO 500
95 IF (LT.EQ.12) CALL INFILTER(NVAR,K,ITEM) !Special FILTER Handling
M=K+1
DO K=M,130
IF (ITEM(K).EQ.SPA.AND.ITEM(K+1).NE.SPA) GO TO 200

```


! INPUT

```

ENDDO
TYPE *, '***** INCORRECT DATA FORMAT *****'
IERR=1
GO TO 500

200  M=K+1
      RLC(NC,NVAR)=DIGIT(M,ITEM,IERR) !Convert the numeric value
      IF(IERR.LT.2)GO TO 300           !Bad input
      TYPE *, '***** VARIABLE VALUE ERROR *****'
      GO TO 500

300  CONTINUE
350  CALL CONTROL(M,NC,ITEM,TDIS,TOFF,DC,FS,VOFF,IERR,NEND,NEYE,nff,npp)
      IF(NEND.GT.0)GO TO 600

500  CONTINUE      !Comment or blank line, no processing, next line

600  CLOSE(UNIT=1) !.END, end of input file
      IF(MOUT.GT.0)GO TO 1000
      TYPE *, '***** NO OUTPUT WAS SPECIFIED *****'
1000 TYPE *, '*****'
      TYPE *, ' '
      TYPE *, ' THE FOLLOWING OUTPUT FILES WILL BE CREATED: '
      type *, ' '
      type *,BFIL(1:time_graph_size ), ' = time domain REGIS file'
      type *,DFIL(1:time_data_size ), ' = time domain data file'
      IF(NFF.EQ.1)type *,FGL(1:freq_graph_size ), ' = freq. domain REGIS'
      IF(NFF.EQ.1)type *,FDL(1:freq_data_size ), ' = freq. domain data'
      IF(NPP.EQ.1)type *,PGL(1:phase_graph_size ), '= phase vs. freq. REGIS'
      IF(NPP.EQ.1)type *,PDL(1:phase_data_size ), '= phase vs. freq. data fi
      type *, ' '
      TYPE *, '*****'

      IF(IERR.GT.0)STOP
      RETURN
      END

```

SUBROUTINE LINE (N,W)

```

CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC
C SOLVES FOR THE A,B,C,D PARAMETERS FOR A TRANSMISSION LINE C
C FOR THE Nth NETWORK ELEMENT AT FREQUENCY W (TYPE 4 CIRCUIT) C
CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC
COMPLEX*16 ABCD (500,2,2),ZERO,Z,Y,ZO,GAM,GAMJ,A,B,C,Q,CON
DOUBLE PRECISION PI2,X
COMMON/NET/RLC (500,20),ABCD,PI2,ZERO
CON=DCMPLX (0.0,-1.0)
BC=W*RLC (N,20)
W1=RLC (N,11)
IF (W1.EQ.0.0) GO TO 5
IF (W.LE.W1) GO TO 5
W2=RLC (N,12)
IF (W2.EQ.0.0.OR.W.LE.W2) THEN
    BC=BC*(1.+RLC (N,17)*ALOG (W/W1))
ELSE
    BC=BC*(1.+RLC (N,7)+RLC (N,18)*ALOG (W/W2))
ENDIF
5 XL=W*RLC (N,19)
RT=RLC (N,4)
QD=RLC (N,16)
IF (RT.EQ.0.0.OR.QD.EQ.0.0) GO TO 20 !Skin effect calculation
QD=QD*SQRT (W)
IF (QD.LT.1.E-5) GO TO 20
R=0.5*QD*RT
IF (QD.GT.15) GO TO 10
R=R/(COSH (QD)-COS (QD))
RT=R*(SINH (QD)+SIN (QD))
XL=XL+R*(SINH (QD)-SIN (QD))
GO TO 20
10 RT=R
XL=XL+R
20 X=RLC (N,3)
D=RLC (N,6)
Z=DCMPLX (RT,XL)
Y=BC*DCMPLX (D,1.)
GAM=X*CDSQRT (Z*Y)
GAMJ=CON*GAM
A=CDCOS (GAMJ)
IF (W.GT.1.E-5) GO TO 50
Q=1.0+GAM**2/6.0
B=Z*X*Q
C=Y*X*Q
GO TO 100
50 Q=CDSIN (GAMJ)/CON
ZO=CDSQRT (Z/Y)
B=ZO*Q
C=Q/ZO
100 ABCD (N,1,1)=A
ABCD (N,1,2)=B
ABCD (N,2,1)=C
ABCD (N,2,2)=A
RETURN
END
```

! Check for even or odd ZEN
 ! Check for 1st GEN

```
! MABCD
! MOVE
```

```

SUBROUTINE MABCD (A,M,IFLAG)
CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC
C
C SOLVES THE MATRIX MULTIPLICATION A(I,J) = ABCD(M,I,J)*A(I,J)
C 2/16/87
C
CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC
COMPLEX*16 ABCD(500,2,2),ZERO,A(2,2),B(2,2)
DOUBLE PRECISION P12
COMMON/NET/RLC(500,20),ABCD,P12,ZERO
N=2
DO 100 I=1,N
DO 100 J=1,N
B(I,J)=ZERO
DO 100 K=1,N
B(I,J)=B(I,J)+A(I,K)*ABCD(M,K,J)
C=CDABS(B(I,J))
IF(C.GT.1.E34)IFLAG=1
100 CONTINUE
DO 200 I=1,N
DO 200 J=1,N
200 A(I,J)=B(I,J)
RETURN
END

```

```

SUBROUTINE MOVE (A,M)
CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC
C
C MOVES ABCD(M,I,J) TO A(I,J)
C 12/31/85
C
CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC
COMPLEX*16 ABCD(500,2,2),ZERO,A(2,2),B(2,2)
DOUBLE PRECISION P12
COMMON/NET/RLC(500,20),ABCD,P12,ZERO
N=2
DO 100 I=1,N
DO 100 J=1,N
100 A(I,J)=ABCD(M,I,J)
RETURN
END

```

!

1 PRECALC

SUBROUTINE PRECALC (NC,NF,TDIS,TOFF,NGEN)

CALCULATION OF CERTAIN CIRCUIT PARAMETERS TO REDUCE
RECURRENT CALCULATIONS, INITIALIZATIONS, & OVERLAPPING 1
CORRECTION

NC = number of network sections (circuits) in AFILE
NF = number of frequency terms computed in Fourier Transform
NGEN = total number of generators in model in AFILE
TDIS = amount of time to be graphed
EDGE = array of frequency terms for a SRC model
XX = array of time dependent voltages for SRC input;
also for real part of Fourier transform of time waveform
YY = array for imaginary part of Fourier Transform of SRC
time domain waveform

COMPLEX*16 ABCD (500,2,2),ZERO

complex*16 edge (8192)

DOUBLE PRECISION PER,PI2

DIMENSION TOFF (5),dc1 (500)

dimension xx (8192),yy (8192)

COMMON/NET/RLC (500,20),ABCD,PI2,ZERO

COMMON/MISC/KIND (500),RGEN,NOUT (5),MOUT,KD (5),iff

COMMON/TIME/PER

common/expt1/edge

DATA XX/8192*0/,YY/8192*0/

DEL=1000000.

NGEN=0

ndgen=0 !Count the number of GEN networks

nsrcc=0 !Count number of SRC networks

TRM=1.E10

IDF=0

1 Setup for default for dual slope case

tr1=rlc (1,2)

tr2=rlc (2,2)

if ((tr2.ne.tr1).and.(tr2.ne.0.0))idf=1 !Dual slope case

DO 50 I=1,NC

GO TO (120, 200, 300, 400, 500, 600, 700, 800, 900,1000,

1 1100,1200,1300,1400,1500,1600,1700,1800,1900,2000),KIND (I)

GEN-GENERATOR INPUT PROCESSING

Upon exiting PRECALC the GEN entires in RLC are:

10-90% rise and fall times in elements 2 and 9 respectively

50% pulse width in element 1

IF (RLC (I,1).NE.0.0) GO TO 125

TYPE #,' ***** GENERATOR PULSE WIDTH NOT SPECIFIED *****'

STOP

25 NGEN=NGEN+1

ndgen=ndgen+1

dc1 (I)=rlc (I,7)

ick=jmod (I,2)

IF (I.EQ.1) GO TO 130

!Check for even or odd GEN

! Check for 1st GEN

! PRECALC

```

DO J=1,100
TRO=RISE(WO,T,0.8)-RISE(WO,T,0.2)
IF (ABS(TRO-TF).LT.1.E-4) GO TO 160
T=T+1.2*(TF-TRO)
ENDDO

150 T=0.0
TYPE 115,1,T
160 T=T/0.6
TDF50=RISE(WO,(0.6*T),0.5)-T/2.
! convert "ideal" pulse width (instantaneous rise and fall
! times) to real 50% crossing width assumed for the rest of GEN
! setup and calculations
RLC(1,1)=RLC(1,1)+(5.*(TF-TR)/6.)
RLC(1,1)=RLC(1,1)+TDR50-TDF50 !Modify pulse width
170 RLC(1,19)=T !Place T in location 19
IF (RLC(1,4).EQ.0.0) RLC(1,4)=1.0 !Amplitude defaults to one
RLC(1,5)=RLC(1,5)+RLC(1,1)/2.-TDR50
RGEN=RGEN+RLC(1,8) !Generator resistance sum
GO TO 50

C Type 2 Circuit (TEE)
200 RLC(1,16)=PI2*RLC(1,7)
RLC(1,17)=PI2*RLC(1,8)
IF (RLC(1,9).NE.0.0) RLC(1,3)=1./RLC(1,9) !G = 1./R
GO TO 50

C Type 3 circuit (DELTA)
300 IF (RLC(1,7).NE.0.0) RLC(1,1)=1./RLC(1,7) !G1 = 1./R1
IF (RLC(1,8).NE.0.0) RLC(1,2)=1./RLC(1,8) !G2 = 1./R2
GO TO 50

C Type 4 Circuit (LINE)
400 IF (RLC(1,1).NE.0.0.AND.RLC(1,2).NE.0.0.AND.RLC(1,3).NE.0.0)
1 GO TO 410
405 TYPE *, ' ***** LINE IMPROPERLY SPECIFIED ***** '
STOP
410 IF (RLC(1,10).EQ.0.0) RLC(1,10)=1.0 !Q defaults to one
RLC(1,11)=PI2*RLC(1,11) !F1 -> 2 PI F1
RLC(1,12)=PI2*RLC(1,12) !F2 -> 2 PI F2
RLC(1,13)=PI2*RLC(1,13) !F3 -> 2 PI F3
IF (RLC(1,12).EQ.0.0.AND.RLC(1,11).NE.0.0) GO TO 405
IF (RLC(1,11).NE.0.0) RLC(1,17)=RLC(1,7)/ALOG(RLC(1,12)/RLC(1,11))
IF (RLC(1,13).NE.0.0.AND.RLC(1,12).NE.0.0)
1 RLC(1,18)=RLC(1,8)/ALOG(RLC(1,13)/RLC(1,12))
RLC(1,19)=RLC(1,1)*RLC(1,2) !Place Inductance in location 19
RLC(1,20)=RLC(1,2)/RLC(1,1) !Place Capacitance in location 20
IF (RLC(1,9).EQ.0.0) GO TO 420 !WC = 0
IF (RLC(1,5).GT.0.0) GO TO 425 !TC > 0, WC > 0
TYPE *, ' ***** WC CANNOT BE SPECIFIED IF TC IS NOT SPECIFIED ***** '
STOP
25 RLC(1,4)=1.7241E-6*RLC(1,10)/(RLC(1,9)*RLC(1,5)) !Compute Rdc
20 RLC(1,16)=RLC(1,5)*SQRT(PI2/(1.7241E-6*RLC(1,10))) !Compute const
GO TO 50
!

```

1 PRECALC

```

C Type 5 Circuit (STUB)
500 IF (RLC(1,1).NE.0.0.AND.RLC(1,2).NE.0.0.AND.RLC(1,3).NE.0.0)
      1 GO TO 510
      TYPE *, '***** STUB IMPROPERLY SPECIFIED *****'
      STOP
510 RLC(1,17)=RLC(1,7)/PI2           !K1 > K1/2 PI
      RLC(1,18)=RLC(1,8)/PI2           !K2 > K2/2 PI
      RLC(1,19)=RLC(1,1)*RLC(1,2)       !Place Inductance in location 19
      RLC(1,20)=RLC(1,2)/RLC(1,1)       !Place Capacitance in location 20
      IF (RLC(1,10).NE.0.0) RLC(1,9)=1./RLC(1,10)           !GL = 1./RL
      IF (RLC(1,12).EQ.0.0) GO TO 520     !WC = 0
      IF (RLC(1,5).GT.0.0) GO TO 525      !TC > 0, WC > 0
      TYPE *, '***** WC CANNOT BE SPECIFIED IF TC IS NOT SPECIFIED *****'
      STOP
525 RLC(1,4)=1.7241E-6/(RLC(1,12)*RLC(1,5)) !Compute Rdc
520 RLC(1,16)=RLC(1,5)*SQRT(PI2/1.7241E-6) !Compute const
      GO TO 50

C Type 6 Circuit (EAMP)
600 IF (RLC(1,1).EQ.0.0) RLC(1,1)=1.0           ! GAIN defaults to 1
      IF (RLC(1,2).LT.1.E-6) RLC(1,2)=1.E-6     ! R defaults to 1E-6
      BW=RLC(1,3)
      WO=PI2*BW
      IF (WO.EQ.0.0) WO=50./TRM                 !F defaults to MAX
      RLC(1,16)=WO                               !Place WO in location 16
      GO TO 50

C Type 7 Circuit (FIBER)
700 IF (RLC(1,6).EQ.0.0) RLC(1,6)=1.0           !QM defaults to one
      IF (RLC(1,7).EQ.0.0) RLC(1,7)=1.0           !QC defaults to one
C Code for Siecor Dispersion Model
      IF (RLC(1,5).NE.0.0) GO TO 720
      ALC=RLC(1,9)                               !D not zero
      IF (ALC.NE.0.0) GO TO 710
      TYPE *, '***** NO LC SPECIFIED *****'
      STOP
710 DP=0.25*RLC(1,11)*(ALC-RLC(1,10)**4/ALC**3)
      rlc(i,5)=.85*rlc(i,3)*rlc(i,11)           !0.85*FWHM*SO
      rlc(i,5)=rlc(i,5)**2
      rlc(i,5)=rlc(i,5)/8.
      rlc(i,5)=(dp**2)+rlc(i,5)
      RLC(1,5)=SQRT(rlc(i,5))
      RLC(1,5)=1.176*RLC(1,5)
720 XM=RLC(1,8)**RLC(1,6)
      XC=RLC(1,8)**RLC(1,7)
      A=ALOG(10.)/10.
      B=4.*SQRT(ALOG(2.))
      C=A/B*RLC(1,2)*RLC(1,3)*XC
      RLC(1,18)=C
      RLC(1,16)=A*ABS(RLC(1,1))*XC-C**2
      RLC(1,17)=RLC(1,3)*RLC(1,5)*XC/B
      IF (RLC(1,4).LT.1.E-4) RLC(1,19)=0.0
      IF (RLC(1,4).GE.1.E-4) RLC(1,19)=B*XM/RLC(1,4)/(4.*PI2) !TM, material disp
      GO TO 50
!Intermodal length correction
!Chromatic length correction
!Wavelength dependent atten
!Total attenuation
!TC, chromatic dispersion
!F1 defaults to infinity

```

C Type 8 Circuit (GSIN)

```

800 IF (RLC(1,1).NE.0.0) GO TO 810
    TYPE *, '***** GENERATOR FREQUENCY NOT SPECIFIED *****'
    STOP

810 RLC(1,16)=PI2*RLC(1,1)           !F > WO
    IF (RLC(1,2).EQ.0.0) RLC(1,2)=1.0 !AMP defaults to one
    NGEN=NGEN+1                       !Increase number of generators
    RGEN=RGEN+RLC(1,5)                 !Generator resistance sum
    P=1./RLC(1,1)
    IF (P.GT.PER) PER=P
    IF (TRM.GT.P) TRM=P
    RLC(1,4)=RLC(1,4)+1./(4.*RLC(1,1)) !Add quadrature phase shift
    GO TO 50
  
```

C Type 9 Circuit (IAMP)

```

900 IF (RLC(1,1).EQ.0.0) RLC(1,1)=1.0           ! GAIN defaults to 1
    IF (RLC(1,2).NE.0.0) RLC(1,3)=1./RLC(1,2) !If R not zero, G=1/R
    IF (RLC(1,3).LT.1.E-6) RLC(1,3)=1.E-6      !G defaults 1.E-6
    BW=RLC(1,4)
    WO=PI2*BW
    IF (WO.EQ.0.0) WO=50./TRM                   !F > MAX
    RLC(1,16)=WO                                 !Place WO in location 16
    GO TO 50
  
```

C Type 10 Circuit (HPF)

```

000 IF (RLC(1,1).GT.0.0) GO TO 1010
    TYPE *, '***** C MISSING FROM HPF NETWORK *****'
    STOP

1010 IF (RLC(1,4).NE.0.0) RLC(1,3)=1./RLC(1,4)
    IF (RLC(1,3).GT.0.0) GO TO 1020
    TYPE *, '***** SHUNT CONDUCTANCE MISSING FROM HPF NETWORK *****'
    STOP

1020 RLC(1,20)=RLC(1,2)*RLC(1,3)               !G * L
    RLC(1,19)=RLC(1,3)/RLC(1,1)                !G / C
    GO TO 50
  
```

C Type 11 Circuit (XFORMER)

```

1100 IF (RLC(1,1).EQ.0.0.AND.RLC(1,5).EQ.0.0) GO TO 1110 !L1 = 0 & N = 0
    IF (RLC(1,2).EQ.0.0.AND.RLC(1,5).EQ.0.0) GO TO 1110 !L2 = 0 & N = 0
    IF (RLC(1,1).EQ.0.0.AND.RLC(1,2).EQ.0.0) GO TO 1110 !L1 = 0 & L2 = 0
    IF (RLC(1,1).EQ.0.0) RLC(1,1)=RLC(1,2)/RLC(1,5)**2 !L1 = 1 / N**2
    IF (RLC(1,2).EQ.0.0) RLC(1,2)=RLC(1,1)*RLC(1,5)**2 !L2 = L1 N**2
    IF (RLC(1,4).GT.0.0) RLC(1,3)=RLC(1,4)*SQRT(RLC(1,1)*RLC(1,2))
    IF (RLC(1,3).EQ.0.0) RLC(1,3)=SQRT(RLC(1,1)*RLC(1,2)) !K > 1
    RLC(1,16)=RLC(1,1)/RLC(1,3)                 !L1/M
    RLC(1,17)=RLC(1,2)/RLC(1,3)                 !L2/M
    RLC(1,18)=RLC(1,1)*RLC(1,2)/RLC(1,3)-RLC(1,3) !L1 L2 / M - M
    GO TO 50

1110 TYPE *, '***** INDUCTOR MISSING FROM TRANSFORMER *****'
    STOP
  
```



```

C Type 12 Circuit (FILTER)
1200 IF (RLC(1,5) .NE.0.0) RLC(1,4)=1.0/RLC(1,5) !G1 > 1 / R1
      IF (RLC(1,2) .EQ.0.0) RLC(1,2)=1.0
      IF (RLC(1,6) .NE.0.0) RLC(1,1)=0.25/RLC(1,6) !First Zero
      IF (RLC(1,1) .EQ.0.0) GO TO 1210
      GO TO 50

1210 TYPE *, '***** To MUST BE SPECIFIED IN FILTER *****'
      STOP

C Type 13 Circuit (SHUNT)
1300 IF (ABS(RLC(1,11)) .EQ.0.0) RLC(1,11)=1.0 !K defaults to One
      bt=abs(rlc(i,6))+abs(rlc(i,7))+abs(rlc(i,8))+abs(rlc(i,9))
      bt=bt+abs(rlc(i,10))
      if(bt.eq.0) rlc(i,6)=1.      !set B0=1 if no B's spec'd
      GO TO 50

C Type 14 Circuit (SERIES)
1400 IF (ABS(RLC(1,11)) .EQ.0.0) RLC(1,11)=1.0 !K defaults to One
      bt=abs(rlc(i,6))+abs(rlc(i,7))+abs(rlc(i,8))+abs(rlc(i,9))
      bt=bt+abs(rlc(i,10))
      if(bt.eq.0) rlc(i,6)=1.      !set B0=1 if no B's spec'd
      GO TO 50

C Type 15 Circuit (BPF)
1500 rlc(i,1)=pi2*rlc(i,1)          !w1
      rlc(i,2)=pi2*rlc(i,2)          !w2
      if(rlc(i,3) .eq.0) rlc(i,3)=6.  !S1 default = 6 dB/octave
      if(rlc(i,4) .eq.0) rlc(i,4)=6.  !S2 default = 6 dB/octave
      rlc(i,3)=abs(rlc(i,3))
      rlc(i,4)=abs(rlc(i,4))
      if(rlc(i,5) .eq.0) rlc(i,5)=1.  !K default = 1. (gain)
      rlc(i,16)=rlc(i,1)*(1.+3./rlc(i,3))  !AA
      rlc(i,17)=rlc(i,2)/((1.+3./rlc(i,4))) !BB
      rlc(i,18)=1-(46.99/rlc(i,3))
      rlc(i,18)=rlc(i,1)*rlc(i,18)
      rlc(i,19)=1+(50/rlc(i,4))
      rlc(i,19)=rlc(i,17)*rlc(i,19)
      RLC(1,20)=0.5*(RLC(1,1)+RLC(1,2))
      rlc(i,14)=rlc(i,1)-rlc(i,2)
      rlc(i,14)=(2*rlc(i,6))/rlc(i,14)
      GO TO 50
      !ps=slope of phase line

C Type 16 Circuit (SOURCE)
1600 ngen=ngen+1
      nsrc=nsrc+1
      if(nsrc.le.1) go to 1601
      type *, ' ONLY 1 SRC ALLOWED PER MODEL '
      stop

1601 rgen=rgen+rlc(i,9)
      P=rlc(i,3)
      if(p.gt.per) per=p
      OPEN(UNIT=1, FILE='SOURCE.DTA', STATUS='OLD')
      TYPE=RLC(1,8)
      IF((TYPE.EQ.1) .OR. (TYPE.EQ.2)) GO TO 5002
      TYPE *, '

```

```

TYPE *, ' DATA FILE TYPE INCORRECT '
STOP
5002 nt=RLC(1,6)      !Number of points, must be .le. 8192
if(nt.le.8192)go to 6002
type *, ' TOO MANY POINTS IN SOURCE.DTA '
stop
6002 do ii=1,11
if(nt.lt.2**ii)go to 5071
enddo
5071 nt=2**(ii-1)    !Reset number of points to be a power of 2
if(rlc(i,6).eq.nt)go to 5051
type *, '
type *, ' NOT ALL POINTS IN SOURCE.DTA WILL BE READ; reduced '
TYPE *, ' to a power of 2 as required '
TYPE *, '
dt=rlc(i,7)        !Time increment between points
per=(nt-1)*dt     !Correct period for truncated waveform
5051 IF (TYPE.EQ.1)GO TO 5003    !SOURCE.DTA already has frequency data
!
! Read time domain data
DO II=1,NT
5662 READ(1,5662)XX(II)
FORMAT(F20.3)
ENDDO
CLOSE(UNIT=1)
! Calculate frequency spectrum of input waveform and place it
! in complex COMMON array EDGE
CALL FFTB42(1,NT,XX,YY)
DO II=1,NT
EDGE(II)=CMPLX(XX(II),YY(II))
edge(ii)=dconjg(edge(ii))
ENDDO
5003 GO TO 50
dt=per/nt
rlc(i,7)=amax1(dt,rlc(i,7))
do ii=1,nt
5772 read(1,5772)edge(ii)
format(2e16.2) !May be changed for different data format
enddo
CLOSE(UNIT=1)
go to 50
C Type 17 Circuit
1700 CONTINUE
C Type 18 Circuit
1800 CONTINUE
C Type 19 Circuit
1900 CONTINUE
C Type 20 Circuit
2000 CONTINUE
50 CONTINUE

```

```

!
! Correction for overlapping 1's so that source level does not
! exceed normalized "ON" level amplitude
!
IF (KIND(1).NE.1) GO TO 555 !Skip overlapping 1's correction
!if 1st circuit not a GEN

N1=NGEN
IGEN=1 !One or two trapezoids per edge
IF (RLC(1,2).NE.RLC(2,2)) N1=NGEN-1 !Two trapezoid case
F1=5*RLC(1,9)/4. !0-100% falltime
R1=5*RLC(1,2)/4. !0-100% risetime
F2=5*RLC(2,9)/4.
R2=5*RLC(2,2)/4.
PN1=RLC(N1,1)+0.5*(R1-F1) !ideal width (next to) last GEN
PN=RLC(NGEN,1)+.5*(R2-F2) !ideal width for last GEN
P1=RLC(1,1)+0.5*(R1-F1) !ideal pulse width for 1st
P2=RLC(2,2)+0.5*(R2-F2) !ideal width for 2nd GEN

IF (N1.NE.NGEN) GO TO 9556
TDR52=TDR51
9556 IF (N1.EQ.NGEN) GO TO 9559
IGEN=2
9559 DN1=RLC(N1,5)-RLC(N1,1)/2.+TDR51
DN=RLC(NGEN,5)-RLC(NGEN,1)/2.+TDR52
CORR=ABS(RLC(N1,3)+DEL1-DN1) !CORR=|PER+DEL1-DN1|
IF ((RLC(1,3)+DEL1).LE.(DN1+PN1)) GO TO 9555

IF ((RLC(1,3)+DEL1).GE.(DN1+PN1+F1)) GO TO 555
IF ((DN1+PN1+F1).LE.(RLC(1,3)+DEL1+R1)) GO TO 555

9555 RLC(N1,1)=RLC(1,1)+CORR !correct width (next to) last GEN
RLC(NGEN,1)=RLC(IGEN,1)+CORR !correct width last GEN
RLC(N1,5)=DN1+RLC(N1,1)/2.-TDR51
RLC(NGEN,5)=DN+RLC(NGEN,1)/2.-TDR52
IBG=IGEN+1

DO I=1BG,NC ! drop extra GEN and reset pointers
DO J=1,20
RLC(I-IGEN,J)=RLC(I,J)
KIND(I-IGEN)=KIND(I)
ENDDO
ENDDO
NGEN=NGEN-IGEN
NC=NC-IGEN

do i=2,(ndgen-2) !replace DC offsets of GEN networks after correctio
rlc(i-1,7)=dcl(i)
enddo
rlc(ndgen-1,7)=dcl(1)

DO I=1,MOUT
NOUT(I)=NOUT(I)-IGEN
ENDDO

```

! PRECALC

```

555 ABCD (NGEN,1,1) =DCMPLX (1.0,0.0)
ABCD (NGEN,1,2) =DCMPLX (RGEN,0.0)
ABCD (NGEN,2,1) =ZERO
ABCD (NGEN,2,2) =DCMPLX (1.0,0.0)

C DISPLAY PROCESSING
IF (TDIS.EQ.0.0) TDIS=PER !Default value for horizontal display scale

DO K=1,MOUT !Make sure delay is in range
DO I=1,100
IF (TOFF (K) .LT. (I*PER)) GO TO 9999
ENDDO
9999 TOFF (K) =TOFF (K) - (I-1) *PER
ENDDO

CALCULATING THE NUMBER OF FREQUENCY TERMS TO BE COMPUTED

if (kind(1) .ne.16) go to 5114 !SRC determines number of terms
nf=nt
go to 5115

1114 IF (32768.*TRM.LT.PER) TRM=PER/32768.
X=PER/TRM
K=ALOG10 (10.*X) /ALOG10 (2.)
IF (K.LT.8) K=8
IF (K.GT.13) K=15
NF=2**K
ict=0
5225 anf=anf
err=per/anf !Check time increment magnitude=error in timing jitter
if (err.le.0.125) go to 5115
if (nf.lt.32768) nf=2*nf
ict=ict+1
if (ict.ge.2) go to 5115
go to 5225
5115 if ((iff.eq.1) .and. (nf.gt.8192)) nf=8192
RETURN
END

```

I RISE
I SERIES

FUNCTION RISE (WGEN, TR, A)

```

CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC
C
C COMPUTES THE DELAY FOR THE INPUT WAVEFORM
C WGEN IS THE GENERATOR FILTER 3DB FREQUENCY (RAD/SEC),
C TR IS THE 20 % TO 80 % GENERATOR RISETIME
C A IS THE FRACTIONAL AMPLITUDE
C
CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC

```

```

      C=WGEN*TR/0.6
      X=A*C
      DO I=1,100
      Y=EXP(-X)
      IF (X.GT.C) THEN                               !Delay > TR
        F=C*(1.-A)+Y*(1.-EXP(C))
        DF=Y*(EXP(C)-1.)
      ELSE                                           !Delay <= TR
        F=X-1.0-A*C+Y
        DF=1.-Y
      END IF
      IF (ABS(F).LT.1.E-6) GO TO 200
      X=X-F/DF
      ENDDO
      TYPE *, 'NON-CONVERGENT'
200    RISE=X/WGEN
      RETURN
      END

```

SUBROUTINE SERIES (N,W)

```

CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC
C
C POLYNOMIAL MODEL FOR A SERIES IMPEDANCE
C UP TO FOUR ZEROS AND FOUR POLES
C (TYPE 14 CIRCUIT)
C 12/17/86
CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC
      COMPLEX*16 ABCD(500,2,2), ZERO
      DOUBLE PRECISION P12, PR, P1, ZR, Z1
      COMMON/NET/RLC(500,20), ABCD, P12, ZERO
      W2=W**2
      W3=W**3
      W4=W**4
      ZR=RLC(N,1)-W2*RLC(N,3)+W4*RLC(N,5)
      Z1=W*RLC(N,2)-W3*RLC(N,4)
      PR=RLC(N,6)-W2*RLC(N,8)+W4*RLC(N,10)
      P1=W*RLC(N,7)-W3*RLC(N,9)
      IF (DABS(PR).LT.1.E-16) PR=1.E-16
      ABCD(N,1,1)=1.0
      ABCD(N,1,2)=DCMPLX(ZR,Z1)/(RLC(N,11)*DCMPLX(PR,P1))
      ABCD(N,2,1)=ZERO
      ABCD(N,2,2)=1.0
      RETURN
      END

```

! SHUNT
! SOURCE

SUBROUTINE SHUNT(N,W)

POLYNOMIAL MODEL FOR A SHUNT IMPEDANCE
UP TO FOUR ZEROS AND FOUR POLES
(TYPE 13 CIRCUIT)
5/21/86

COMPLEX*16 ABCD(500,2,2),ZERO
DOUBLE PRECISION P12,PR,P1,ZR,Z1
COMMON/NET/RLC(500,20),ABCD,P12,ZERO

W2=W**2
W3=W**3
W4=W**4

ZR=RLC(N,1)-W2*RLC(N,3)+W4*RLC(N,5)
Z1=W*RLC(N,2)-W3*RLC(N,4)
PR=RLC(N,6)-W2*RLC(N,8)+W4*RLC(N,10)
P1=W*RLC(N,7)-W3*RLC(N,9)
IF(DABS(ZR).LT.1.E-16)ZR=1.E-16

ABCD(N,1,1)=1.0
ABCD(N,1,2)=ZERO
ABCD(N,2,1)=DCMLPX(PR,P1)/(RLC(N,11)*DCMLPX(ZR,Z1))
ABCD(N,2,2)=1.0
RETURN
END

FUNCTION SOURCE(N,W)

RETURNS THE SOURCE GENERATOR FREQUENCY COMPONENT
FOR THE Nth NETWORK ELEMENT AT FREQUENCY W
TYPE 16 CIRCUIT - SRC, experimental data
3/11/87

COMPLEX*16 ABCD(500,2,2),ZERO,source
COMPLEX*16 EDGE(8192)
COMMON/NET/RLC(500,20),ABCD,P12,ZERO
COMMON/TIME/PER
COMMON/EXPTL/EDGE

return source(n,w) from table lookup and fold spectrum abt NF/2

AX=((W*PER)/P12)+1 ! index of frequency term
ix=ax
SOURCE=EDGE(ix)
ifold=rlc(n,6)/2 ! fold spectrum abt NF/2
if(ix.ne.1.and.ix.le.ifold)source=2*source
if(ix.gt.ifold)source=0.
return
end

```

SUBROUTINE STUB (N,W)
CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC
C   SOLVES FOR THE A,B,C,D PARAMETERS FOR A TRANSMISSION LINE STUB C
C   FOR THE Nth NETWORK ELEMENT AT FREQUENCY W (TYPE 5 CIRCUIT) C
C   12/30/85 C
CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC
COMPLEX*16 ABCD (500,2,2) ,ZERO,Z,Y,ZO,GAM,GAMJ,A,C,Q,CON,YL
DOUBLE PRECISION PI2,X
COMMON/NET/RLC (500,20) ,ABCD,PI2,ZERO

CON=DCMLPX (0.0,-1.0)
BC=W*RLC (N,20) *(1.0+ALOG (1.0+W*(RLC (N,17)+W*RLC (N,18))))
RT=RLC (N,4)
XL=W*RLC (N,19)

QO=RLC (N,16)
IF (RT.EQ.0.0.OR.QO.EQ.0.0) GO TO 20      !Skin effect calculation
QO=QO*SQRT (W)
IF (QO.LT.1.E-5) GO TO 20
R=0.5*QO*RT
IF (QO.GT.15) GO TO 10
R=R/(COSH (QO)-COS (QO))
RT=R*(SINH (QO)+SIN (QO))
XL=XL+R*(SINH (QO)-SIN (QO))
GO TO 20
10  RT=R
    XL=XL+R

20  X=RLC (N,3)
    D=RLC (N,6)

    YL=CMPLX (RLC (N,9) ,W*RLC (N,11))
    Z=DCMLPX (RT,XL)
    Y=BC*DCMLPX (D,1.)
    GAM=X*CDSQRT (Z*Y)
    GAMJ=CON*GAM
    A=CD COS (GAMJ)

    IF (W.GT.1.E-5) GO TO 50
    Q=1.0+GAM**2/6.0
    C=(YL*A+Y*X*Q)/(A+YL*Z*X*Q)
    GO TO 100

50  Q=CDSIN (GAMJ)/CON
    ZO=CDSQRT (Z/Y)
    C=(YL*A+Q/ZO)/(A+ZO*YL*Q)
100 ABCD (N,1,1)=DCMLPX (1.0,0.0)
    ABCD (N,1,2)=ZERO
    ABCD (N,2,1)=C
    ABCD (N,2,2)=DCMLPX (1.0,0.0)

RETURN
END

```

! TEE

SUBROUTINE TEE (N,W)

SOLVES FOR THE A,B,C,D PARAMETERS FOR THE TEE CIRCUIT
FOR THE Nth NETWORK ELEMENT AT FREQUENCY W (TYPE 2 CIRCUIT)
12/11/85

COMPLEX*16 ABCD (500,2,2),ZERO
DOUBLE PRECISION P12
COMMON/NET/RLC (500,20),ABCD,P12,ZERO

R1=RLC (N,1)
R2=RLC (N,2)
G=RLC (N,3)
XL1=W*RLC (N,4)
XL2=W*RLC (N,5)
BC=W*RLC (N,6)
W1=RLC (N,16)
W2=RLC (N,17)

IF (W1.EQ.0.0) GO TO 10
RT=R1*(1+(W/W1)**2)**.25
RAC=RT-R1
R1=RT

XL1=XL1+RAC
IF (W2.EQ.0.0) GO TO 20
RT=R2*(1+(W/W1)**2)**.25
RAC=RT-R2
R2=RT

XL2=XL2+RAC
ABCD (N,1,1)=DCMLX ((1.+R1*G-XL1*BC), (XL1*G+R1*BC)) !A
ABCD (N,1,2)=DCMLX ((R1+R2+G*R1*R2-G*XL1*XL2-BC*(R1*XL2+R2*XL1)),
1 (XL1+XL2+G*(R1*XL2+R2*XL1)+BC*R1*R2-XL1*XL2*BC)) !B
ABCD (N,2,1)=DCMLX (G,BC) !C
ABCD (N,2,2)=DCMLX ((1.+R2*G-XL2*BC), (XL2*G+R2*BC)) !D
RETURN
END


```

SUBROUTINE TRANSFORM(NC,NF,NGEN)
CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC
C
C   COMPUTES THE FREQUENCY DOMAIN RESPONSE OF THE NETWORK
C
C
CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC
  DOUBLE PRECISION PER,PI2
  COMPLEX*16 ABCD(500,2,2),ZERO,A(2,2),B(2,2),EW(5,32768),
  1 EG,GEN,GSIN,SOURCE
  CHARACTER*6 NODENAME(54)
  COMMON/NET/RLC(500,20),ABCD,PI2,ZERO
  COMMON/MISC/KIND(500),RGEN,NOUT(5),MOUT,KD(5),iff
  COMMON/WAVES/EW,EOUT(5,32768)
  COMMON/TIME/PER
  DATA IFLAG/O/
  WD=PI2/PER
  TYPE *, ' THERE ARE ',NF,' FREQ. TERMS TO COMPUTE'      !busy msg
  type *, ' '
  DO 100 M=1,NF
  MM=MOD(M,300)      !busy msg processing
  IF((NF.GT.300).AND.(MM.EQ.0))TYPE 990,M
  990  FORMAT(' FOCAS IS ALIVE & WELL . . . freq term: ',16)
  W=WO*(M-1)
  EG=ZERO
  DO 200 K=1,NC
  GO TO (220,240,260,280,300,320,330,340,350,360,
  1      380,400,420,440,460,480,500,520,540,560),KIND(K)
  220  EG=EG+GEN(K,W)      !Type 1 Circuit
  GO TO 200
  240  CALL TEE(K,W)      !Type 2 Circuit
  GO TO 200
  260  CALL DELTA(K,W)   !Type 3 Circuit
  GO TO 200
  280  CALL LINE(K,W)   !Type 4 Circuit
  GO TO 200
  300  CALL STUB(K,W)   !Type 5 Circuit
  GO TO 200
  320  CALL EAMP(K,W)   !Type 6 Circuit
  GO TO 200
  330  CALL FIBER(K,W)  !Type 7 Circuit
  GO TO 200
  340  EG=EG+GSIN(K,W)  !Type 8 Circuit
  GO TO 200
  350  CALL IAMP(K,W)   !Type 9 Circuit
  GO TO 200
  360  CALL HPF(K,W)   !Type 10 Circuit
  GO TO 200
  380  CALL XFORMER(K,W) !Type 11 Circuit
  GO TO 200
  400  CALL FILTER(K,W) !Type 12 Circuit
  GO TO 200

```

! TRANSFORM

```

420 CALL SHUNT (K,W) !Type 13 Circuit
GO TO 200
440 CALL SERIES (K,W) !Type 14 Circuit
GO TO 200
460 CALL BPF (K,W) !Type 15 Circuit
GO TO 200
480 eg=eg+SOURCE (K,W) !Type 16 Circuit
GO TO 200
500 CONTINUE !Type 17 Circuit
520 CONTINUE !Type 18 Circuit
540 CONTINUE !Type 19 Circuit
560 CONTINUE !Type 20 Circuit
200 CONTINUE
N1=NGEN
CALL MOVE (A,N1)
DO 5000 K=1,MOUT
KOUT=NOUT (K)
IF (KOUT.EQ.N1) GO TO 5200
NS=N1+1
DO 5100 I=NS,KOUT
CALL MABCD (A,I,IFLAG)
5200 IF (KOUT.EQ.NC) GO TO 5300
CALL MOVE (B,(KOUT+1))
N2=KOUT+2
IF (N2.GT.NC) GO TO 5400
DO 5500 I=N2,NC
CALL MABCD (B,I,IFLAG)
5400 EW (K,M)=EG/(A (1,1)+A (1,2)*B (2,1)/B (1,1))
GO TO 5000
5300 EW (K,M)=EG/A (1,1)
5000 N1=KOUT
IF (IFLAG.EQ.1) GO TO 6000
100 CONTINUE

RETURN
6000 DO I=1,14
IF (M.LT.2**I) GO TO 6100
ENDDO
6100 NF=2** (I-1)
RETURN
END

```


SUBROUTINE: FFT842
 FAST FOURIER TRANSFORM FOR N=2**M
 COMPLEX INPUT

SUBROUTINE FFT842 (IN, N, X, Y)

THIS PROGRAM REPLACES THE VECTOR $Z=X+iy$ BY ITS FINITE DISCRETE,
 COMPLEX FOURIER TRANSFORM IF $IN=0$ (FREQUENCY DOMAIN TO TIME DOMAIN).
 THE INVERSE TRANSFORM IS CALCULATED FOR $IN=1$. IT PERFORMS AS MANY BASE
 8 ITERATIONS AS POSSIBLE AND THEN FINISHES WITH A BASE 4 ITERATION
 OR A BASE 2 ITERATION IF NEEDED.

THE SUBROUTINE IS CALLED AS SUBROUTINE FFT842 (IN,N,X,Y).
 THE INTEGER N (A POWER OF 2), THE N REAL LOCATION ARRAY X, AND
 THE N REAL LOCATION ARRAY Y MUST BE SUPPLIED TO THE SUBROUTINE.

DIMENSION X(2), Y(2), L(15)

COMMON /CON2/ PI2, P7

EQUIVALENCE (L15,L(1)), (L14,L(2)), (L13,L(3)), (L12,L(4)),

* (L11,L(5)), (L10,L(6)), (L9,L(7)), (L8,L(8)), (L7,L(9)),

* (L6,L(10)), (L5,L(11)), (L4,L(12)), (L3,L(13)), (L2,L(14)),

* (L1,L(15))

PI2 = 8.*ATAN(1.)

P7 = 1./SQRT(2.)

DO 10 I=1,15

M = I

NT = 2**I

IF (N.EQ.NT) GO TO 20

10 CONTINUE

TYPE *, ' N IS NOT A POWER OF TWO FOR FFT842'

STOP

20 N2POW = M

NTHPO = N

FN = NTHPO

IF (IN.EQ.1) GO TO 40

DO 30 I=1,NTHPO

Y(I) = -Y(I)

30 CONTINUE

40 NBPOW = N2POW/3

IF (NBPOW.EQ.0) GO TO 60

RADIX 8 PASSES, IF ANY.

DO 50 IPASS=1,NBPOW

NXTLT = 2**(N2POW-3*IPASS)

LENGT = 8*NXTLT

CALL R8TX(NXTLT, NTHPO, LENGT, X(1), X(NXTLT+1), X(2*NXTLT+1),

* X(3*NXTLT+1), X(4*NXTLT+1), X(5*NXTLT+1), X(6*NXTLT+1),

* X(7*NXTLT+1), Y(1), Y(NXTLT+1), Y(2*NXTLT+1), Y(3*NXTLT+1),

* Y(4*NXTLT+1), Y(5*NXTLT+1), Y(6*NXTLT+1), Y(7*NXTLT+1))

50 CONTINUE

```

C IS THERE A FOUR FACTOR LEFT
60 IF (N2POW-3*N8POW-1) 90, 70, 80
! GO THROUGH THE BASE 2 ITERATION
70 CALL R2TX(NTHPO, X(1), X(2), Y(1), Y(2))
GO TO 90
C
C GO THROUGH THE BASE 4 ITERATION
C
80 CALL R4TX(NTHPO, X(1), X(2), X(3), X(4), Y(1), Y(2), Y(3), Y(4))
C
90 DO 110 J=1,15
L(J) = 1
IF (J-N2POW) 100, 100, 110
100 L(J) = 2** (N2POW+1-J)
110 CONTINUE
IJ = 1
DO 130 J1=1,L1
DO 130 J2=J1,L2,L1
DO 130 J3=J2,L3,L2
DO 130 J4=J3,L4,L3
DO 130 J5=J4,L5,L4
DO 130 J6=J5,L6,L5
DO 130 J7=J6,L7,L6
DO 130 J8=J7,L8,L7
DO 130 J9=J8,L9,L8
DO 130 J10=J9,L10,L9
DO 130 J11=J10,L11,L10
DO 130 J12=J11,L12,L11
DO 130 J13=J12,L13,L12
DO 130 J14=J13,L14,L13
DO 130 J1=J14,L15,L14
IF (IJ-J1) 120, 130, 130
120 R = X(IJ)
X(IJ) = X(J1)
X(J1) = R
FI = Y(IJ)
Y(IJ) = Y(J1)
Y(J1) = FI
130 IJ = IJ + 1
IF (IN.EQ.1) GO TO 150
DO 140 I=1,NTHPO
Y(I) = -Y(I)
140 CONTINUE
GO TO 170
150 DO 160 I=1,NTHPO
X(I) = X(I)/FN
Y(I) = Y(I)/FN
160 CONTINUE
170 RETURN
END

```

! R2TX
! R4TX

C SUBROUTINE: R2TX
C RADIX 2 ITERATION SUBROUTINE

SUBROUTINE R2TX (NTHPO, CRO, CR1, C10, C11)
DIMENSION CRO (2), CR1 (2), C10 (2), C11 (2)

DO 10 K=1, NTHPO, 2

R1 = CRO (K) + CR1 (K)

CR1 (K) = CRO (K) - CR1 (K)

CRO (K) = R1

F11 = C10 (K) + C11 (K)

C11 (K) = C10 (K) - C11 (K)

C10 (K) = F11

10 CONTINUE

RETURN

END

C SUBROUTINE: R4TX
C RADIX 4 ITERATION SUBROUTINE

SUBROUTINE R4TX (NTHPO, CRO, CR1, CR2, CR3, C10, C11, C12, C13)
DIMENSION CRO (2), CR1 (2), CR2 (2), CR3 (2), C10 (2), C11 (2), C12 (2),

* C13 (2)

DO 10 K=1, NTHPO, 4

R1 = CRO (K) + CR2 (K)

R2 = CRO (K) - CR2 (K)

R3 = CR1 (K) + CR3 (K)

R4 = CR1 (K) - CR3 (K)

F11 = C10 (K) + C12 (K)

F12 = C10 (K) - C12 (K)

F13 = C11 (K) + C13 (K)

F14 = C11 (K) - C13 (K)

CRO (K) = R1 + R3

C10 (K) = F11 + F13

CR1 (K) = R1 - R3

C11 (K) = F11 - F13

CR2 (K) = R2 - F14

C12 (K) = F12 + R4

CR3 (K) = R2 + F14

C13 (K) = F12 - R4

10 CONTINUE

RETURN

END

C
C-----
C SUBROUTINE: R8TX
C RADIX 8 ITERATION SUBROUTINE
C-----
C

```

SUBROUTINE R8TX(NXTLT, NTHPO, LENGT, CRO, CR1, CR2, CR3, CR4,
* CR5, CR6, CR7, C10, C11, C12, C13, C14, C15, C16, C17)
DIMENSION CRO(2), CR1(2), CR2(2), CR3(2), CR4(2), CR5(2), CR6(2),
* CR7(2), C11(2), C12(2), C13(2), C14(2), C15(2), C16(2),
* C17(2), C10(2)
COMMON /CON2/ P12, P7
SCALE = P12/FLOAT(LENGT)
DO 30 J=1,NXTLT
  ARG = FLOAT(J-1)*SCALE
  C1 = COS(ARG)
  S1 = SIN(ARG)
  C2 = C1**2 - S1**2
  S2 = C1*S1 + C1*S1
  C3 = C1*C2 - S1*S2
  S3 = C2*S1 + S2*C1
  C4 = C2**2 - S2**2
  S4 = C2*S2 + C2*S2
  C5 = C2*C3 - S2*S3
  S5 = C3*S2 + S3*C2
  C6 = C3**2 - S3**2
  S6 = C3*S3 + C3*S3
  C7 = C3*C4 - S3*S4
  S7 = C4*S3 + S4*C3
DO 20 K=J,NTHPO,LENGT
  ARO = CRO(K) + CR4(K)
  AR1 = CR1(K) + CR5(K)
  AR2 = CR2(K) + CR6(K)
  AR3 = CR3(K) + CR7(K)
  AR4 = CRO(K) - CR4(K)
  AR5 = CR1(K) - CR5(K)
  AR6 = CR2(K) - CR6(K)
  AR7 = CR3(K) - CR7(K)
  A10 = C10(K) + C14(K)
  A11 = C11(K) + C15(K)
  A12 = C12(K) + C16(K)
  A13 = C13(K) + C17(K)
  A14 = C10(K) - C14(K)
  A15 = C11(K) - C15(K)
  A16 = C12(K) - C16(K)
  A17 = C13(K) - C17(K)
  BRO = ARO + AR2
  BR1 = AR1 + AR3
  BR2 = ARO - AR2
  BR3 = AR1 - AR3
  BR4 = AR4 - A16
  BR5 = AR5 - A17
  BR6 = AR4 + A16
  BR7 = AR5 + A17

```

```

B10 = A10 + A12
B11 = A11 + A13
B12 = A10 - A12
B13 = A11 - A13
B14 = A14 + AR6
B15 = A15 + AR7
B16 = A14 - AR6
B17 = A15 - AR7
CRO (K) = BRO + BR1
C10 (K) = B10 + B11
IF (J.LE.1) GO TO 10
CR1 (K) = C4*(BRO-BR1) - S4*(B10-B11)
C11 (K) = C4*(B10-B11) + S4*(BRO-BR1)
CR2 (K) = C2*(BR2-B13) - S2*(B12+BR3)
C12 (K) = C2*(B12+BR3) + S2*(BR2-B13)
CR3 (K) = C6*(BR2+B13) - S6*(B12-BR3)
C13 (K) = C6*(B12-BR3) + S6*(BR2+B13)
TR = P7*(BR5-B15)
T1 = P7*(BR5+B15)
CR4 (K) = C1*(BR4+TR) - S1*(B14+T1)
C14 (K) = C1*(B14+T1) + S1*(BR4+TR)
CR5 (K) = C5*(BR4-TR) - S5*(B14-T1)
C15 (K) = C5*(B14-T1) + S5*(BR4-TR)
TR = -P7*(BR7+B17)
T1 = P7*(BR7-B17)
CR6 (K) = C3*(BR6+TR) - S3*(B16+T1)
C16 (K) = C3*(B16+T1) + S3*(BR6+TR)
CR7 (K) = C7*(BR6-TR) - S7*(B16-T1)
C17 (K) = C7*(B16-T1) + S7*(BR6-TR)
GO TO 20
10 CR1 (K) = BRO - BR1
C11 (K) = B10 - B11
CR2 (K) = BR2 - B13
C12 (K) = B12 + BR3
CR3 (K) = BR2 + B13
C13 (K) = B12 - BR3
TR = P7*(BR5-B15)
T1 = P7*(BR5+B15)
CR4 (K) = BR4 + TR
C14 (K) = B14 + T1
CR5 (K) = BR4 - TR
C15 (K) = B14 - T1
TR = -P7*(BR7+B17)
T1 = P7*(BR7-B17)
CR6 (K) = BR6 + TR
C16 (K) = B16 + T1
CR7 (K) = BR6 - TR
C17 (K) = B16 - T1

```

20 CONTINUE

30 CONTINUE

RETURN

END

VI. ADDING A NEW NETWORK TYPE

In order to add a new network type to FOCAS, one must first calculate the chain matrix for the new network section as described in section 11 of this document. The definitions of A, B, C, and D thus derived are coded into a new subroutine, SUBROUTINE NEWNAME, the purpose of which is to compute the chain matrix for this new network subsection type. The subroutine will compute the ABCD parameters for the network at the frequency W based on the subsection's circuit parameters defined in array RLC.

In general, SUBROUTINE NEWNAME must have as a minimum:

```
SUBROUTINE XXXX(N,W)
COMPLEX*16 ABCD(500,2,2),ZERO
DOUBLE PRECISION PI2          ! = 2.*PI
COMMON/NET/RLC(500,20),ABCD,PI2,ZERO
```

```

      |
      | Programmer Defined
VAR=RLC(N, ) ----- "
      |
      | "
      | "
      | "
      | "
ABCD(N,1,1) = ----- "
ABCD(N,1,2) = ----- "
ABCD(N,2,1) = ----- "
ABCD(N,2,2) = ----- "
RETURN
END
```

The variable N indicates which of the input model's subsections is being analyzed, and the location of its parameters in the RLC array. The variable W is the radian frequency in Gigaradians/second.

The chain matrix for the Nth network section is represented in FOCAS as ABCD(N,2,2), with the following definitions.

```
ABCD(N,1,1) = A
ABCD(N,1,2) = B
ABCD(N,2,1) = C
ABCD(N,2,2) = D .
```

Once this subroutine has been coded, several changes are made to the existing FOCAS code.

A name for the two-terminal-pair network with the first three characters being unique must be defined, e.g. NEW. Next, the input parameters must each be defined by two character names. Present array space will hold up to 20 network parameters. The following changes are now to be made:

1. SUBROUTINE INPUT

- a. Add the new Network name (three characters) to the NAME array.
- b. Add the two character parameter names to the NDATA array. More than two characters are ignored so the first two must be unique. The parameter names must be on the line number corresponding to the location of Network name in the NAME array. If less than 12 parameters are specified, fill to the left with the last parameter name. This maps the parameter values into the RLC array.
- c. Increase NKIND by one.

2. SUBROUTINE PRECALC

- a. Perform any preliminary calculations which are desired to manipulate the input data, set default values, or reduce computational overhead during later frequency dependent calculations.

3. SUBROUTINE TRANSFORM

- a. Add a call statement to this subroutine for the Network named above. The value of KIND is the same as the location of the three character name in the NAME array in the INPUT subroutine.

4. Recompile all programs and subroutines and relink, including SUBROUTINE NEWNAME. The new version of FOCAS, supporting the new network subsection type is now ready to run.

VI. ENHANCEMENTS

A. Complex Filters & Resonant Circuits

In the present formulation of FOCAS, one may encounter an arithmetic overflow condition:

```
%SYSTEM-F-FLTOVF_F, arithmetic fault, floating overflow at PC=0030626;
PSL=03C00000
```

when simulating networks in which more than two subsections have poles and/or zeroes at the same frequency. An example of such a simulation model is given below:

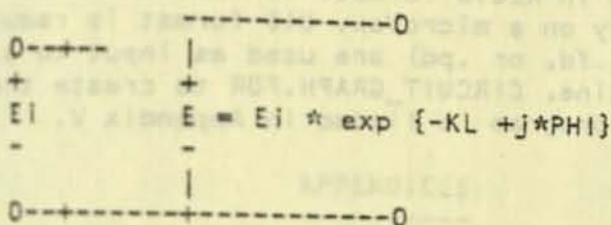
```
GEN TR=.05 PW=.0625 PER=1000 AMP=2 r=1000
series a0=1 a2=25 b1=.16
shunt a1=.012 b0=1 b2=25
series a0=1 a2=25355.98 b1=.0000163
shunt a0=1000 b0=1
.out
.fdg ff
.END TX=6 DC=.5 FS=3.
```

The overflow occurs in subroutine MABCD line 20, where the complex magnitude of each element of the chain matrix is calculated. There are two options to be investigated for overcoming this limitation.

By converting all floating point calculations to the G_FLOATING format the maximum magnitude is increased from 1.7×10^{38} to 9×10^{307} , but the accuracy of double precision is lost. Alternatively, one could change the divide by zero checking in all the subsection subroutines (BPF, SERIES, SHUNT, etc) so that any denominator less than 10^{-10} (rather than the present 10^{-16}) is set equal to 10^{-10} (as opposed to 10^{-16}). This change would mean that the response around resonances would be somewhat flattened, but overall accuracy is not altered as much. If it becomes necessary to simulate complex filter sections, both options should be investigated carefully before a choice is made. FOCAS is not intended to support detailed filter design. There are specialized software packages for that purpose.

B. Optical Cable Splices

A splice in the optical fiber can be modeled simply as a section which introduces a loss, L db, and a phase shift, PHI radians. This type of model will neglect the effects of reflections at the near end of the fiber, but will provide a useful approximation to the behavior at the far end (beyond the splice). The two-terminal-pair network for such a model would appear as:



$$K = \text{db conversion factor} = \frac{\ln 10}{10} = 0.23026$$

The chain matrix elements for this splice model are:

$$\begin{aligned} A &= \exp \{KL - j*PHI\} & B &= 0 \\ C &= 0 & D &= 0 \end{aligned}$$

C. Reduced Timing Uncertainty

The uncertainty in voltage crossing times and timing jitter calculations can be reduced in cases in which fewer than 32,768 frequency terms (NF) are calculated by increasing the value of NF in subroutine PRECALC to 32,768. However, this change will require substantial changes to the numerical algorithms used in subroutine CROSS to find zeroes and extrema of the output waveforms.

The most serious changes will be required in the extremum computations which presently use empirically determined values to discriminate between real extrema and slight variations in the voltage levels due to the compounding of numerical rounding. It was not felt that such a major overhaul is required at present.

D. Fiber Models

More general fiber models are presented in section III and Appendices II and III of this report.

E. Graphics Support on MicroVax

The standard FOCAS output graphics files with the .tg, .fg, or .pg extensions are in REGIS format. In order to view the output graphically on a microVax, UIS format is required. The data files (.td, .fd, or .pd) are used as input to a separate FORTRAN graphing routine, CIRCUIT_GRAPH.FOR to create the UIS format graphs. This program is listed in Appendix V.

APPENDIX L - LED/CABLE REFERENCE DATA FOR FIGURES 1 & 2

LED SPEC'S

LED No.	1	2	3	4	5	6	7
WAVE	131	125	130	127	126	114	105 nm
λ	1284	1300	APPENDICES		1297	1286	1340 nm

CABLE SPEC'S

	CABLE A	CABLE B
Length	3.7	3.38 km
Transmission	0.82	1.127 dB/km
Attenuation	1.17	0.74 dB/km
LD	1297	1245 nm
LD	9.27×10^{-5}	9.90×10^{-5} ns/(cm ²) - km

21 22

Faint, illegible text, possibly bleed-through from the reverse side of the page.

23 24

APPENDIX 1. - LED/CABLE REFERENCE DATA FOR FIGURES 1 & 2

LED SPEC'S:

LED #:	1	2	3	4	5	6	7
FWHM	137	125	130	127	126	114	159 nm
LC	1284	1306	1280	1298	1291	1296	1340 nm

CABLE SPEC'S:

	CABLE 1	CABLE 2
Length =	2.7	3.38 km
Intermodal Bandwidth	0.82	1.127 GHz-km
Attenuation	1.11	0.74 db/km
LD	1357	1349 nm
SD	9.27×10^{-5}	9.90×10^{-5} ns/(nm**2) -km

Page 118

TABLE 1 - THE TABLE WITH ENGLISH DATA FOR TABLE 1 & 2

	1	2	3	4	5	6	7	8
100	100	100	100	100	100	100	100	100
100	100	100	100	100	100	100	100	100

TABLE 2

100	100
100	100
100	100
100	100

TABLE 3

100	100
100	100
100	100
100	100

2-20-10-2 (continued) - m

2-20-10-2

APPENDIX 11. FIBER RESPONSE WITH CONSTANT D

Equation (18) is used to define a constant D, to approximate the value of the dispersion coefficient over the source spectrum. The fiber response of equation (31) then becomes:

$$(11-1) \quad P_e(w) = \frac{e^{-d^2} e^{-kr_0 X^{Qa}}}{z \sqrt{\pi}} \int_{-\infty}^{\infty} e^{-(L-L_c)^2/z^2 + j\omega X^{Qc} D(L-L_c) - Kr_w(L-L_c) X^{Qc}} dL$$

where $z = \text{FWHM}/(2\sqrt{\ln 2})$. (11-1) is simplified by making the substitutions:

$$(11-2) \quad \alpha = Kr_0 * (X^{**}Qa)$$

$$(11-3) \quad \beta = Kr_w * (X^{**}Qa) - jwD * (X^{**}Qc)$$

$$(11-4) \quad \lambda = L-L_c$$

$$(11-5) \quad P_e(w) = \frac{e^{-d^2} e^{-\alpha}}{z \sqrt{\pi}} \int_{-\infty}^{\infty} e^{-\lambda^2/z^2} e^{-\beta\lambda} d\lambda$$

Defining $a=1/(z^{**2})$ and $b=\beta/2$, (11-5) takes the form of the integral (IV-9).

$$(11-6) \quad P_e(w) = \frac{e^{-d^2} e^{-\alpha}}{z \sqrt{\pi}} \int_{-\infty}^{\infty} e^{-(a\lambda^2 + 2b\lambda)} d\lambda$$

$$(11-7) \quad P_e(w) = \frac{e^{-d^2} e^{-\alpha}}{z \sqrt{\pi}} \sqrt{\pi z^2} e^{-z^2 \beta^2 / 4}$$

$$(11-8) \quad P_e(w) = e^{-d^2 - \alpha + z^2 \beta^2 / 4}$$

$$(11-9) \quad d = 0.1325 * w * (X^{**}Qm) / F1$$

$$(11-10) \quad \alpha = Kr_0 * (X^{**}Qa)$$

$$(11-11) \quad \frac{z^2 \beta^2}{4} = \frac{(F^{**2})}{16(\ln 2)} * \{ Kr_w * (X^{**}Qa) - jwD * (X^{**}Qc) \}^2$$

The FIBER model implemented in FOCAS sets Qa equal to Qc in equation (11-11) as a further restriction.

APPENDIX III. FIBER RESPONSE FUNCTION FOR $D=50*(L-L_c)$

Substituting equation (22) into equation (31), the fiber response function becomes:

$$(III-1) \quad P_e(w) = \frac{e^{-d^2 - K r_0 X^{Q_a}}}{Z \sqrt{\pi}} \int_{-\infty}^{\infty} e^{-\frac{(L-L_c)^2}{Z^2} + j\omega X^{Q_c} S_0 (L-L_0)(L-L_c) - K r_w (L-L_c) \lambda} d\lambda$$

where $Z = \text{FWHM} / (2\sqrt{\ln 2})$. (III-1) is simplified by substituting:

$$(III-2) \quad Z_0 = K r_0 * (X^{**} Q_a)$$

$$(III-3) \quad \lambda = L - L_c; \quad \lambda_0 = L_0; \quad \lambda_c = L_c$$

$$(III-4) \quad Z_1 = K r_w * (X^{**} Q_a)$$

$$(III-5) \quad P_e(w) = \frac{e^{-d^2 - Z_0}}{Z \sqrt{\pi}} \int_{-\infty}^{\infty} e^{-\lambda^2/Z^2 - \alpha \lambda + j\omega X^{Q_c} S_0 \lambda (\lambda + \lambda_c - \lambda_0)} d\lambda$$

Letting I_1 be the integral in (III-5), $P_e(w)$ becomes

$$(III-6) \quad P_e(w) = \frac{\exp\{-(d^{**2}) - Z_0\}}{Z * \text{SQRT}(\pi)} * I_1$$

The expression for I_1 can be simplified by the substitutions:

$$(III-7) \quad \alpha = j\omega * (X^{**} Q_c) * S_0$$

$$(III-8) \quad \delta = \lambda_c - \lambda_0$$

$$(III-9) \quad I_1 = \int_{-\infty}^{\infty} e^{-\lambda^2/Z^2 - \alpha \lambda + \alpha \lambda (\lambda + \delta)} d\lambda = \int_{-\infty}^{\infty} e^{-\lambda^2(\frac{1}{Z^2} - \alpha) - \lambda(Z_1 - \alpha \delta)} d\lambda$$

Again substitutions can make the integral more manageable:

$$(III-10) \quad k = [1/(Z^{**2})] - \alpha = \frac{4(\ln 2)}{(F^{**2})} - j\omega * (X^{**} Q_c) * S_0$$

$$(III-11) \quad l = \frac{Z_1 - \alpha \delta}{2} = \frac{K r_w * (X^{**} Q_a) - j\omega * (X^{**} Q_c) * S_0 * (\lambda_c - \lambda_0)}{2}$$

It now has the form (IV-9):

$$(III-12) \quad I_1 = \int_{-\infty}^{\infty} e^{-(k\lambda^2 + 2l\lambda)} d\lambda = \sqrt{\frac{\pi}{k}} e^{l^2/k}$$

Making the substitutions:

$$(III-13) \quad Z_2 = S_0 * (X^{**}Q_c) * \delta$$

$$(III-14) \quad Z_3 = S_0 * (X^{**}Q_c) * (F^{**2})$$

$$(III-15) \quad T_0 = 4 * (F^{**2}) * (Z_1^{**2}) * (\ln 2)$$

$$(III-16) \quad T_1 = 2 * (F^{**2}) * \{ Z_1 * Z_2 * Z_3 - 2 * (Z_1^{**2}) * (\ln 2) \}$$

$$(III-17) \quad T_2 = (F^{**2}) * \{ (Z_1^{**2}) * Z_3 - 8 * Z_1 * Z_2 * (\ln 2) \}$$

$$(III-18) \quad T_3 = (F^{**2}) * (Z_2^{**2}) * Z_3$$

$$(III-19) \quad T_4 = 64 * [(\ln 2)^{**2}]$$

$$(III-20) \quad T_5 = 4 * (Z_3^{**2})$$

$$(III-21) \quad T_6 = T_2 - T_3$$

Equation (III-6) now becomes:

$$(III-22) \quad P_e(w) = \frac{F * \exp\{-(d^{**2}) - Z_0\}}{Z} * \sqrt{\frac{1}{4 \ln 2 - j \omega Z_3}} * \exp\left\{ \frac{T_0 + T_1 * (w^{**2}) + j \omega * T_6}{T_4 + T_5 * (w^{**2})} \right\}$$

with $d = 0.01325 * (X^{**}Q_m) * w / F_1.$

Equation (III-22), along with the definitions of the intermediate terms can be used to define a somewhat more generalized fiber cable model than that used in FIBER.

121 2000

Handwritten notes and a small diagram or table at the top of the page.

Main body of handwritten text, appearing to be a list or series of entries, possibly related to a survey or data collection.

A section of text with a horizontal line and a small diagram or box, possibly a signature or a specific note.

Additional handwritten text at the bottom of the page, including some numbers and possibly a date.

APPENDIX IV. INTEGRALS & MATHEMATICAL IDENTITIES
(Reference 1)

$$(IV-1) \quad \int_{-\infty}^{\infty} e^{-t^2/2} dt = \sqrt{2\pi}$$

$$(IV-2) \quad \int_0^z e^{-t^2/2} dt = \frac{\sqrt{\pi}}{2} \operatorname{erf} z$$

$$(IV-3) \quad \int_0^{\infty} e^{-(at^2+2bt)} dt = \frac{1}{2} \sqrt{\frac{\pi}{a}} e^{b^2/a} \operatorname{erfc} \frac{b}{\sqrt{a}}$$

$$(IV-4) \quad \frac{2}{\sqrt{\pi}} \int_z^{\infty} e^{-t^2} dt = \operatorname{erfc} z = 1 - \operatorname{erf} z$$

$$(IV-5) \quad \operatorname{erf}(-z) = -\operatorname{erf}(z)$$

$$(IV-6) \quad \frac{1}{\sigma\sqrt{2\pi}} \int_{-\infty}^x e^{-(t-m)^2/2\sigma^2} dt = \frac{1}{2} \left[1 + \operatorname{erf} \left(\frac{x-m}{\sigma\sqrt{2}} \right) \right]$$

$$(IV-7) \quad \operatorname{erfc}(-z) = 1 + \operatorname{erf}(z)$$

$$(IV-8) \quad \operatorname{erfc}(z) + \operatorname{erfc}(-z) = 2$$

$$(IV-9) \quad \int_{-\infty}^{\infty} e^{-(at^2+2bt)} dt = \sqrt{\frac{\pi}{a}} e^{b^2/a}$$

1941-42

BRITISH CAPITAL MARKET & INVESTMENT IN INDIA
(1941-42)

$$\frac{1}{2} \pi \frac{d^2 r}{dt^2} = \frac{1}{2} \pi \frac{d^2 r}{dt^2}$$

$$\frac{1}{2} \pi \frac{d^2 r}{dt^2} = \frac{1}{2} \pi \frac{d^2 r}{dt^2}$$

$$\frac{1}{2} \pi \frac{d^2 r}{dt^2} = \frac{1}{2} \pi \frac{d^2 r}{dt^2}$$

$$\frac{1}{2} \pi \frac{d^2 r}{dt^2} = \frac{1}{2} \pi \frac{d^2 r}{dt^2}$$

$$\frac{1}{2} \pi \frac{d^2 r}{dt^2} = \frac{1}{2} \pi \frac{d^2 r}{dt^2}$$

$$\left[\frac{(m-x)}{2t} \right] \frac{1}{2} = \frac{1}{2}$$

$$\frac{1}{2} \pi \frac{d^2 r}{dt^2} = \frac{1}{2} \pi \frac{d^2 r}{dt^2}$$

$$\frac{1}{2} \pi \frac{d^2 r}{dt^2} = \frac{1}{2} \pi \frac{d^2 r}{dt^2}$$

$$\frac{1}{2} \pi \frac{d^2 r}{dt^2} = \frac{1}{2} \pi \frac{d^2 r}{dt^2}$$

APPENDIX V. CIRCUIT_GRAPH.FOR LISTING

```

program circuit_graph
! plot circuit output
! 8-OCT-1986 Mike Taylor
! Modified 3-MARCH-1987 Ann Ewalt
! based on an examples in the
! MicroVMS Workstation Graphics Programming Guide
! Generates a graph of the two input

implicit integer(a-z)
external circuit_graph_cmd ! cld file
external lib$get_foreign !
external cli$dcl_parse, cli$present, cli$get_value

include 'sys$library:uisentry'
include 'sys$library:uisusrdef'

character*4 string
real array1(350), array2(350)
real xx(350),yy(350)
real largest_x, smallest_x, largest_y, smallest_y
real world_x1, world_x2, world_y1, world_y2
real scale_x, scale_y
real x, x2, height, y

CHARACTER*32 DFILE
CHARACTER*255 command_line
CHARACTER*255 verb
CHARACTER*255 title, x_title, y_title

! get the name of the input file from the user.

status = cli$get_value( '$line', command_line, command_line_size)
if ( .NOT. status ) call lib$signal( %val( status ))
! type *, command_line(1:command_line_size)

status = cli$get_value( '$verb', verb, verb_size)
if ( .NOT. status ) call lib$signal( %val( status ))
! type *, verb (1:verb_size)

status = cli$dcl_parse( command_line , circuit_graph_cmd )
if ( .NOT. status ) call lib$signal( %val( status ))

if ( cli$present( 'file_name' ) ) then
  call cli$get_value( 'file_name' , DFILE )
else
  TYPE 90
  ACCEPT 95,DFILE
endif
  
```



```

90     FORMAT(' Enter the graph's data file name: ',S)
95     FORMAT(A32)
99     continue

    If (cli$present('title')) then
        call cli$get_value('title', title, title_size)
    else
        title = ' '
        title_size = 1
    endif

    If (cli$present('x_title')) then
        call cli$get_value('x_title', x_title, x_title_size)
    else
        x_title = ' Time '
        x_title_size = 6
    endif

    If (cli$present('y_title')) then
        call cli$get_value('y_title', y_title, y_title_size)
    else
        y_title = ' Voltage '
        y_title_size = 9
    endif

! the first col in file is time in nano-seconds or frequency
! in MHz (plot on X)
! the second col in file is volts or phase shift in radians,
! or normalized power (plot on y)
! find the largest and smallest of array

largest_x = 0.0
smallest_x = 0.0
largest_y = 0.0
smallest_y = 0.0
MOUT = 1 ! number of graphs
points = 350

OPEN( UNIT=21, FILE=DFILE, STATUS='OLD', READONLY)
DO I=1,Points
    read (21,1000) array1(I), array2(I) ! ann had a 2D array
    IF ( largest_x .LT. array1(I) ) largest_x = array1(I)
    IF ( smallest_x .GT. array1(I) ) smallest_x = array1(I)
    IF ( largest_y .LT. array2(I) ) largest_y = array2(I)
    IF ( smallest_y .GT. array2(I) ) smallest_y = array2(I)
ENDDO
1000 FORMAT(1X,F20.3,<MOUT>(1X,F20.3)) ! <mout>?
CLOSE(UNIT=21)

```

```

! type *, largest_x, smallest_x, largest_y, smallest_y
! create the display and window
vd_id = uis$create_display( -15.0,-15.0,410.0,350.0,240.0,200.0 )
! the first four numbers are the world coordinates
! the window size in centimeters

wd_id = uis$create_window(vd_id,'Sys$workstation',
2' Focus Output ')

! make the axis 5 times wider than normal
Call uis$set_line_width( vd_id, 0,16,5.0 )
Call uis$plot( vd_id, 16,0,0, 0,330.0 ) ! draw Y axis
Call uis$plot( vd_id, 16,0,0, 400.0,0 ) ! draw X axis

! the legend of the graph
Call uis$text( vd_id, 0, title ,
2 190.0,345.0 )

! set the scaling factors
! 330 is the coordinat distance
scale_x = (ABS(largest_x) + ABS(smallest_x)) / 400.0
scale_y = (ABS(largest_y) + ABS(smallest_y)) / 330.0
! type *, scale_x, scale_y

! information along the Y axis
! Y is always between -1 and 4 (for now!)
do 20 i = 33,330,33
    y = (float(i) * scale_y) + smallest_y
    if ( y .GE. 1.0 ) then
        encode (4,11,string) y
    else
        encode (4,10,string) y
    endif
    y = float(i)
10 format (F3.1)
11 format (F3.1)
    Call uis$plot( vd_id, 16, 0,y,-3.0,y ) ! tick mark

20    Call uis$text( vd_id, 0, string, -14.0, Y ) ! value
Call uis$text( vd_id, 0, Y_title , -7.0, 340.0 )
! information along the X axis
do 40 i = 40, 400 , 40
    y = (float(i) * scale_x) + smallest_x
    n = y
    encode (4,30,string) n
    y = float(i)
30 format (F10.2)
    Call uis$plot( vd_id, 16, y,0,y,-3.0 ) ! tick mark
    y = y - 8 ! center the value under the tick
40 Call uis$text( vd_id, 0, string, y, -7.0 ) ! value

```

```
Call uis$text( vd_id, 0, x_title , 200.0, -12.0 )
! convert the input points to plotting coordinates
do 50 i = 1,Points
    xx(i) = (array1(i) + ABS(smallest_x)) /scale_x
    yy(i) = (array2(i) + ABS(smallest_y)) /scale_y
50 continue
! set the line width back to normal
Call uis$set_line_width( vd_id, 0,16,1.0 )
do 100 i = 2,Points ! do the plot
    Call uis$Plot( vd_id, 16, xx(i-1), yy(i-1), xx(i), yy(i) )
100 continue
pause ! back to DCL, use continue to retrun to the program
! if you run this from inside a TPU subprocess
! the pause doesn't work.
end
```

ACKNOWLEDGEMENTS

FOCAS is based on an early version of SIMER, a network analysis tool developed by Don Marshall in the HPS technology group. Without Don's patience and the running start SIMER provided, FOCAS would not have been completed as quickly or in as useable a form.

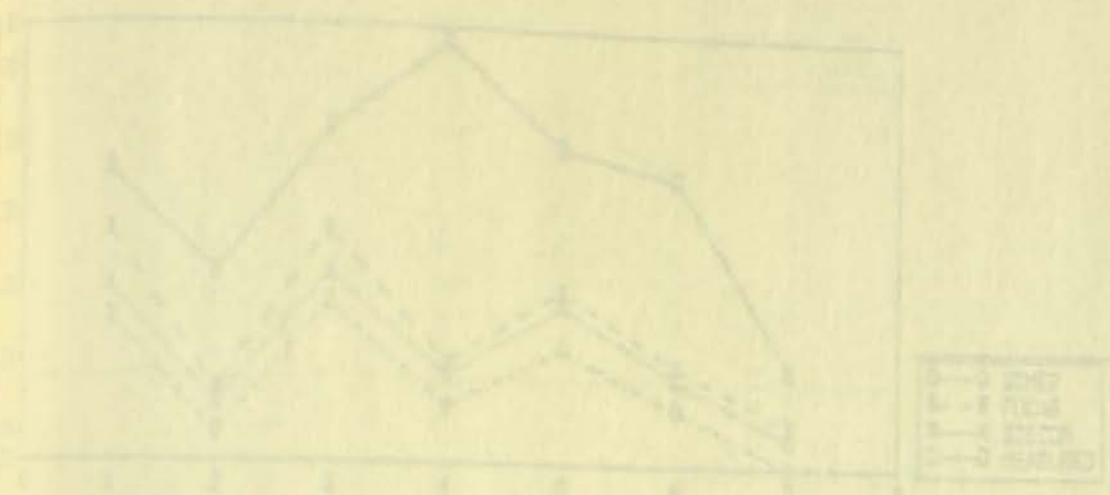
Mike Taylor of the SARA Technical Office helped as a general software consultant throughout the project. He made several improvements to the user interface that are impossible to get when restricted to FORTRAN. In addition, he coded the UIS graphics routine for microVax support and served as a sounding board and sanity check for many enhancements and corrections to the code. Mike spent many late nights working on code improvements.

Seymon Ginzburg and Don Knudson (DS A/D) provided on-going sanity checks for the modelling and results and made frequent suggestions for improvements. Bruce Schofield, Hans Rolla, and Bill Gover (P&DS) provided experimental support to validate these checks, and, when necessary, to redirect the efforts.

BIBLIOGRAPHY

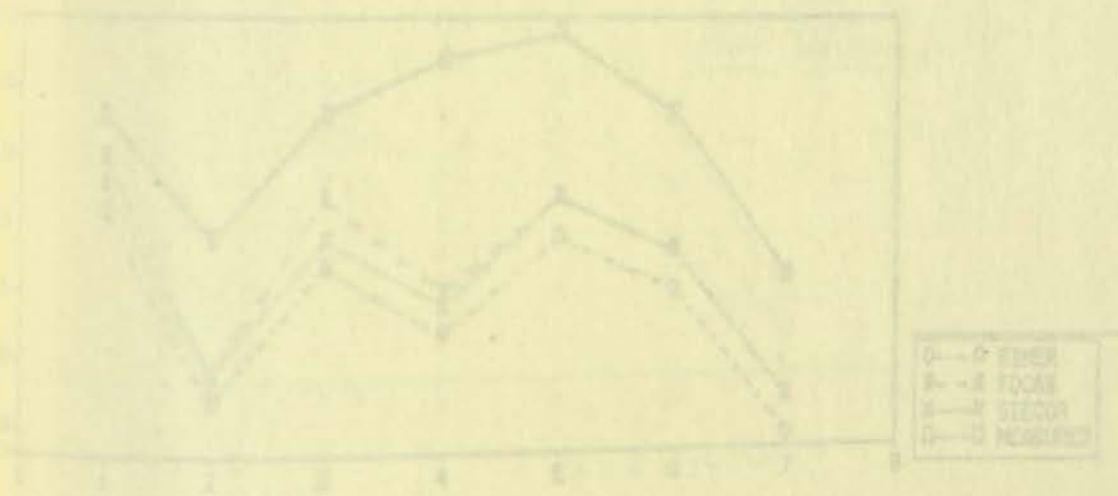
1. Abramowitz, Milton & Stegun, Irene, "Handbook of Mathematical Functions", Dover Publications, Inc., New York, N. Y., 1972
2. Button, Leslie; Love, Walter; & Hawk, Robert, "Chromatic Dispersion in Optical Fibers: Comparison of Measurements with Models", Corning Glass Works, Corning, N. Y.
3. Keiser, Gerd, "Optical Fiber Communications", McGraw-Hill Book Company, 1983
4. Marshall, Don, "SIMER Application Notes", REV 6, 3/20/86
5. Schickentanz, Dieter, notes from private communication with Seymon Ginzburg, Jerry Hutchison, Don Knudson, & Bruce Schofield
6. Weinberg, Louis, "Network Analysis and Synthesis", Robert E. Kreiger Publishing Co., Inc., Huntington, N.Y., 1975

RISE TIME VS. MODEL
Data 1



FIGURES
----- LED NUMBER

FALL TIME VS. MODEL
Data 1



LED NUMBER

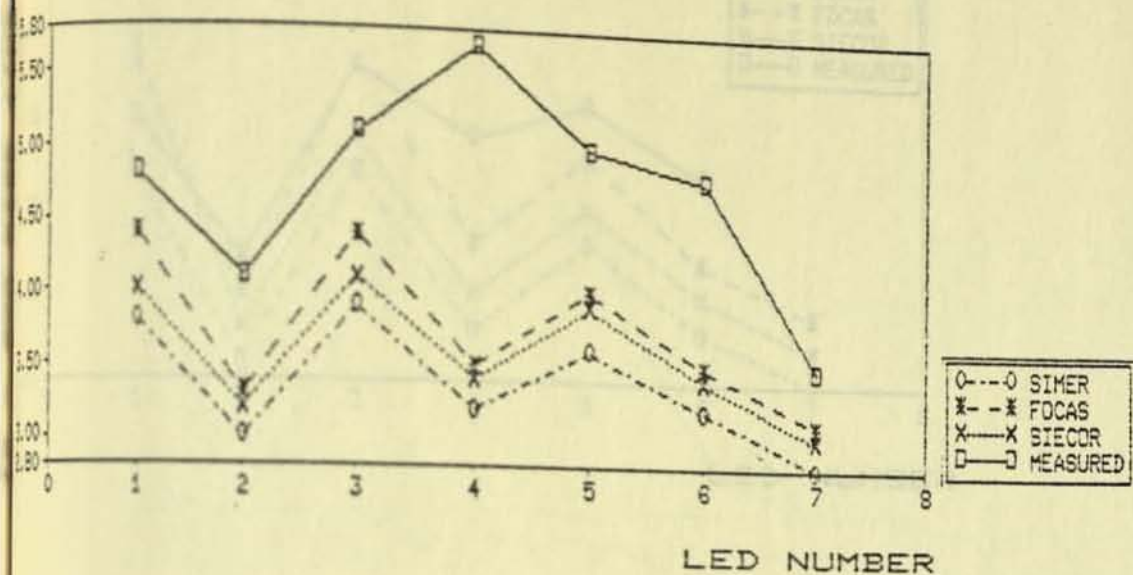
The first part of the paper is devoted to a general discussion of the problem. It is shown that the problem is equivalent to the problem of finding the minimum of a certain functional. This functional is defined as the sum of the squares of the differences between the observed values and the values calculated from the model. The minimum of this functional is found by the method of least squares.

The second part of the paper is devoted to the derivation of the equations for the parameters of the model. These equations are obtained by differentiating the functional with respect to the parameters and setting the derivatives equal to zero. The resulting equations are solved by the method of successive approximations.

The third part of the paper is devoted to the numerical solution of the equations. The equations are solved by the method of successive approximations. The results of the calculations are presented in the form of a table.

The fourth part of the paper is devoted to the discussion of the results. It is shown that the results of the calculations are in good agreement with the observed values. This indicates that the model is a good representation of the physical process.

RISE TIME VS. MODEL
Cable 1



FALL TIME VS. MODEL
Cable 1

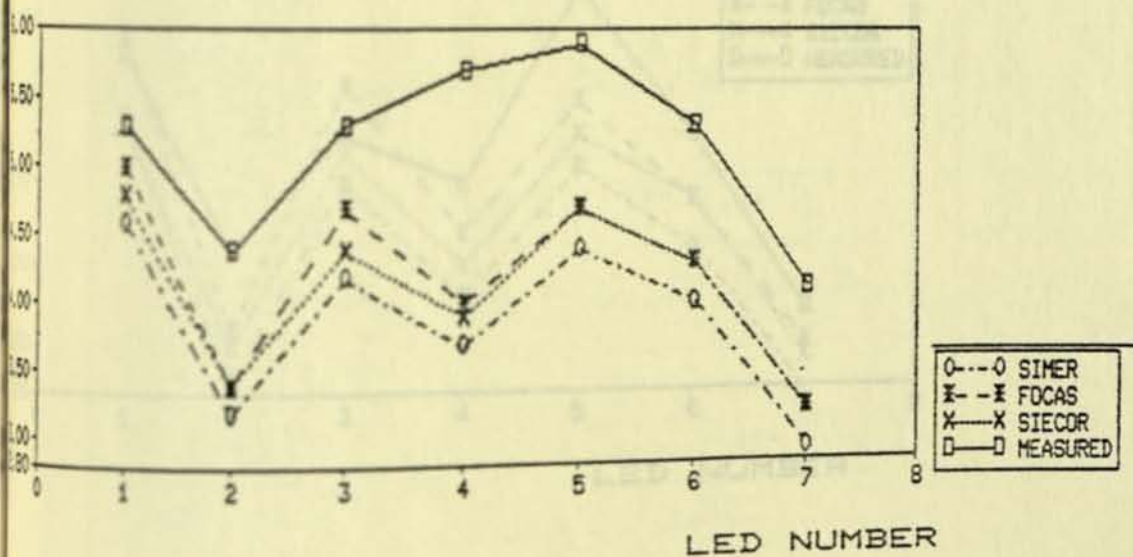
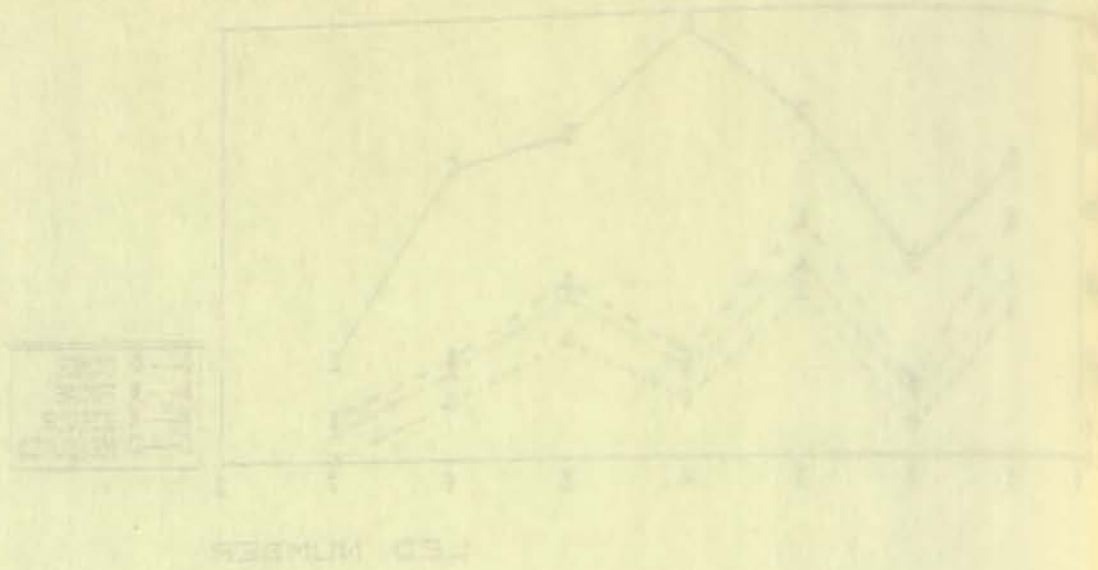
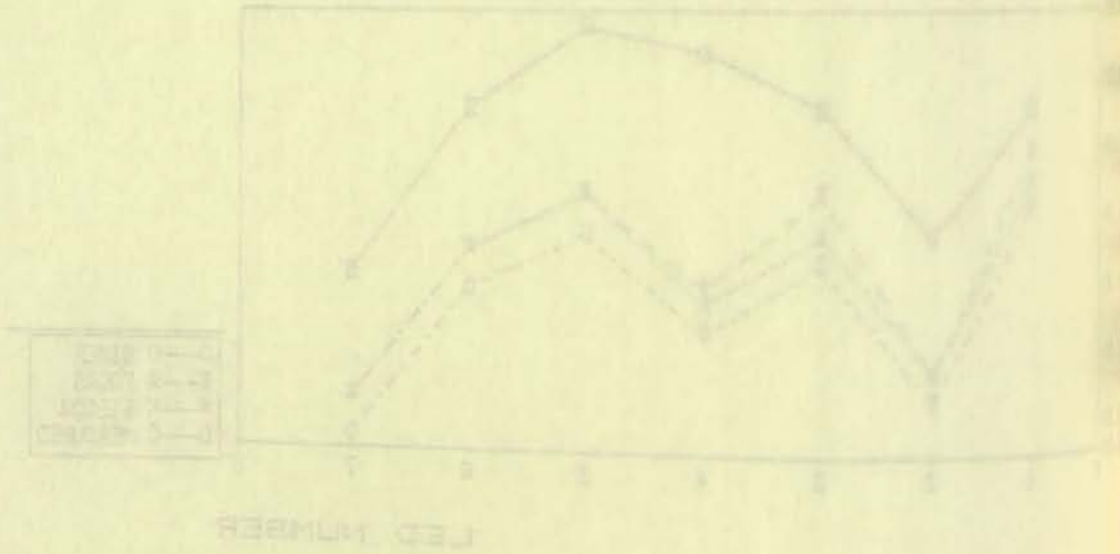


FIGURE 1

1930 TIME VS. MODEL
Chart 1

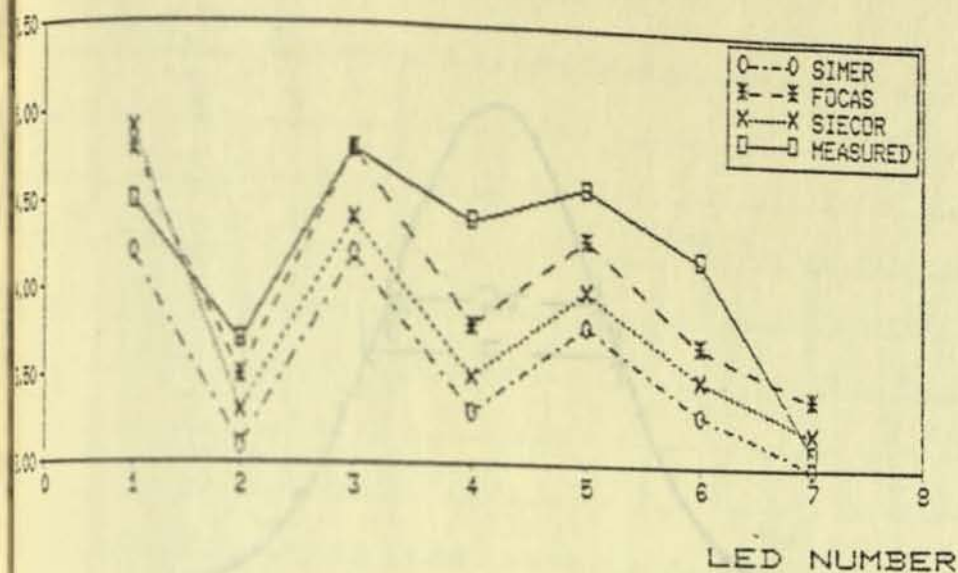


1931 TIME VS. MODEL
Chart 2



RISE TIME VS. MODEL

Cable 2



FALL TIME VS. MODEL

Cable 2

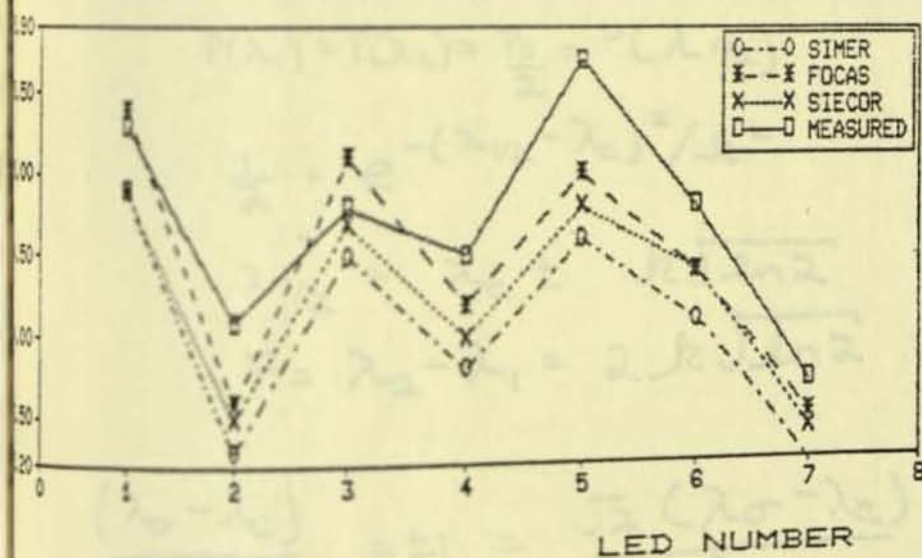
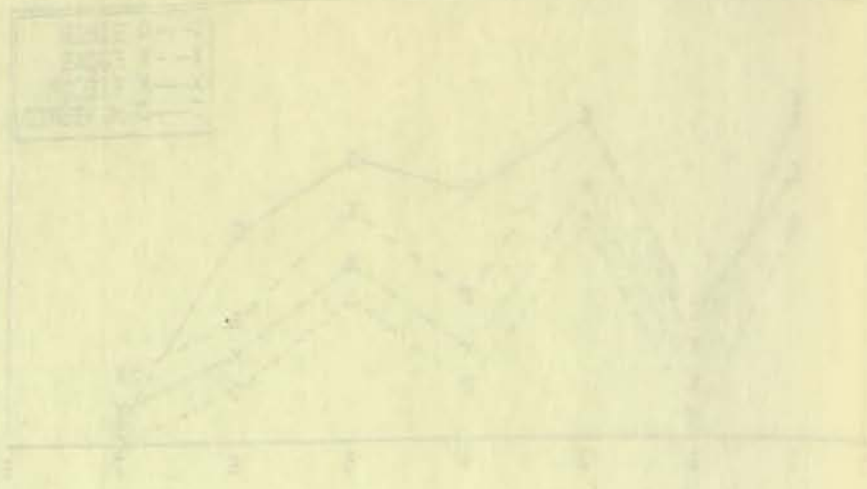


FIGURE 2

$$\lambda_0 = \lambda_c \pm k\sqrt{2}/2$$

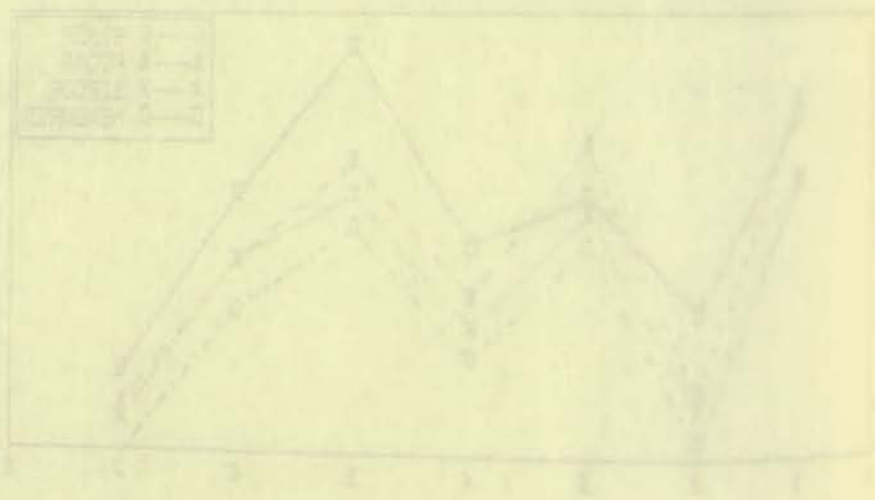
$$2\theta = \frac{F}{\sqrt{2}m^2} \sqrt{2} = \frac{F}{\sqrt{2}m^2} = 0.899F$$

WIRE TIME VS. MODEL
DAY 2



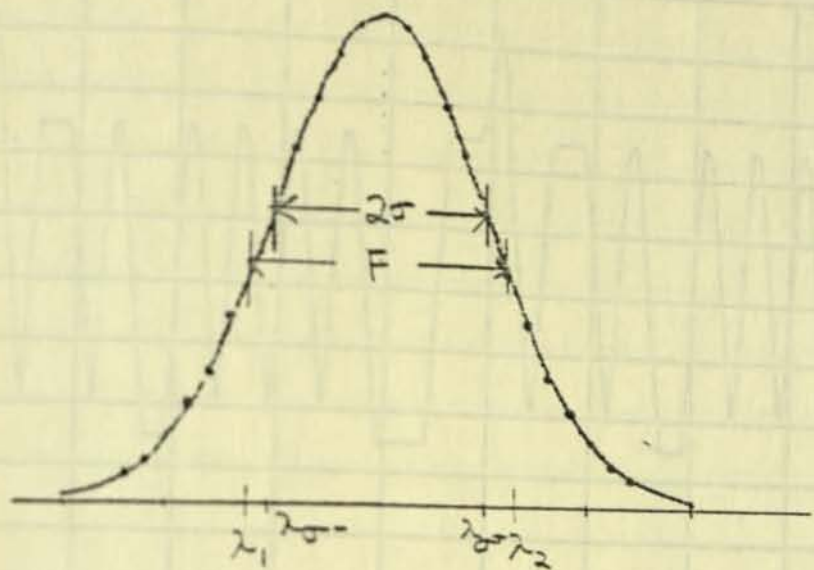
LED NUMBER

WIRE TIME VS. MODEL
DAY 1



LED NUMBER

WIRE 1



$$P_{\text{SD}}(\lambda) = P_0 e^{-(\lambda - \lambda_0)/k^2} = P_0 e^{-(\lambda - \lambda_0)^2/2\sigma^2}$$

$$P(\lambda_1) = P(\lambda_2) = \frac{P_0}{2} = P(\lambda_{1/2})$$

$$\frac{1}{2} = e^{-(\lambda_{1/2} - \lambda_0)^2/k^2}$$

$$\lambda_{1/2} = \lambda_0 \pm k\sqrt{\ln 2}$$

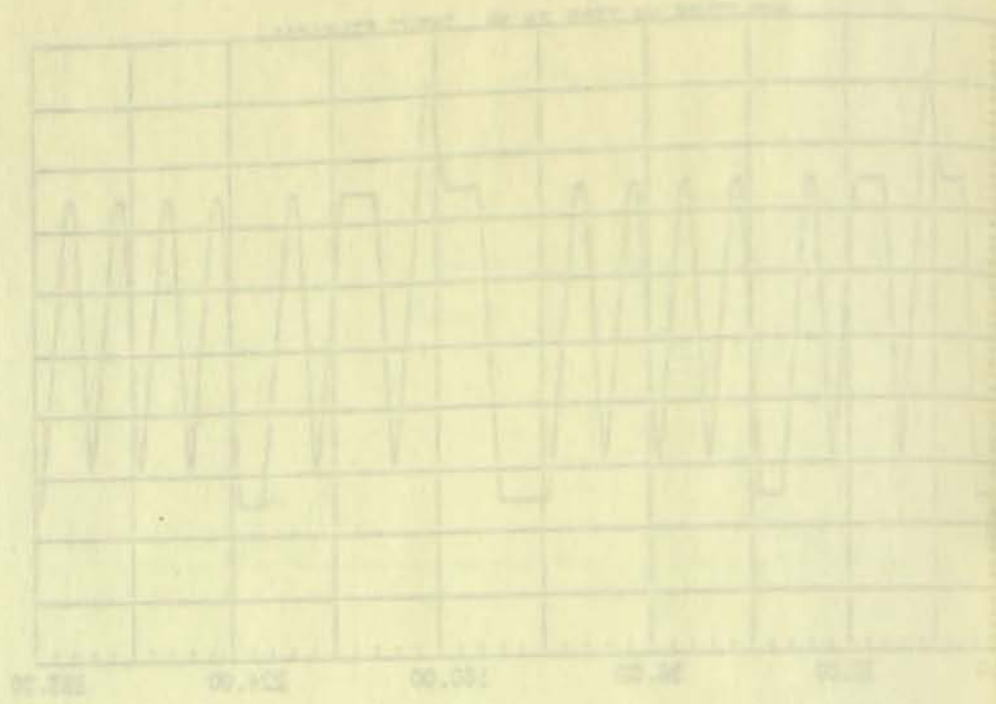
$$F = \lambda_2 - \lambda_1 = 2k\sqrt{\ln 2}$$

$$\frac{(\lambda_0 - \lambda_c)}{\sigma} = \pm 1 = \frac{\sqrt{2}(\lambda_0 - \lambda_c)}{k}$$

$$\lambda_0 = \lambda_c \pm k\sqrt{2}/2$$

$$2\sigma = \lambda_{0+} - \lambda_{0-} = k\sqrt{2}$$

$$2\sigma = \frac{F}{2\sqrt{\ln 2}} \sqrt{2} = \frac{F}{\sqrt{2\ln 2}} = 0.849F$$

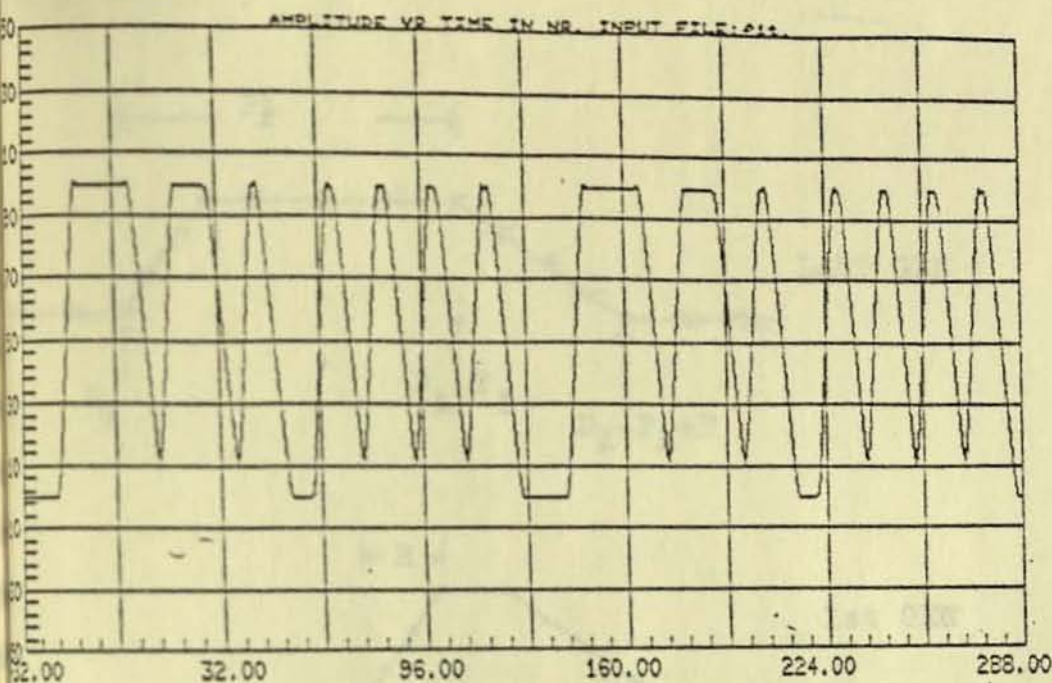


APPROXIMATE PERIOD OF THE SIGNAL IS 0.001 SECONDS

THE SIGNAL IS A COMPLEX PERIODIC WAVEFORM

DATE: 10/10/60
 TIME: 10:00 AM
 LOCATION: LAB 101
 OPERATOR: J. SMITH
 INSTRUMENT: OSCILLOSCOPE
 MODEL: 5000

THE SIGNAL IS A COMPLEX PERIODIC WAVEFORM



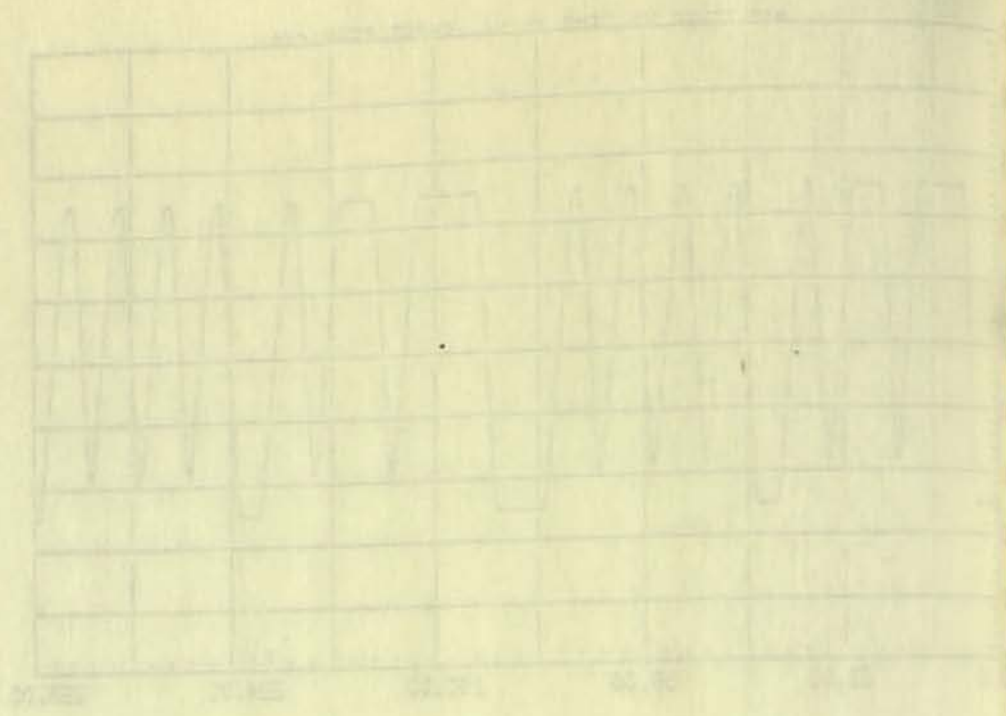
ULATION OF 4 5b PATTERN

First and last pulse generator overlap to give overshoot
for the electronic case where voltage sources add, but
only result in a single wider pulse for the optical case.

PW=8 PER=160 TR=4.4 TF=7
 PW=16 DEL=16
 PW=8 DEL=40
 PW=8 DEL=64
 PW=8 DEL=80
 PW=8 DEL=96
 PW=8 DEL=112
 PW=16 DEL=144

TX=320 DC=.5 FS=2

FIGURE 5. CORRECTED 16110 10010 10011



... and the ...
 ... of the ...
 ... for the ...

...
 ...
 ...
 ...
 ...
 ...
 ...

...
 ...
 ...

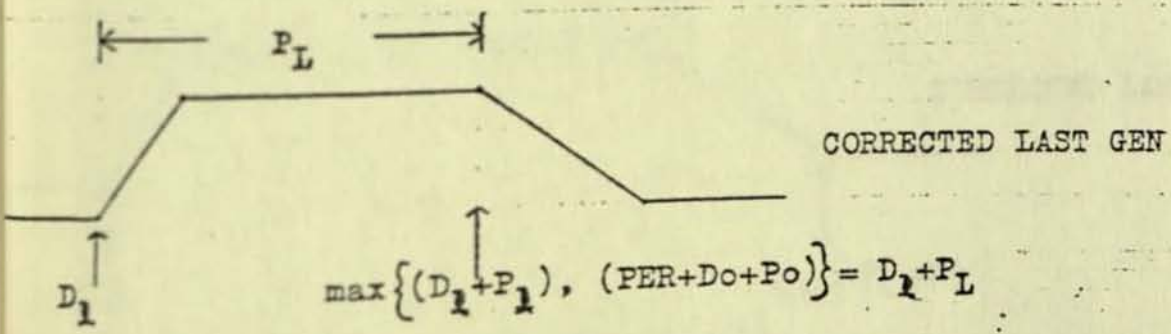
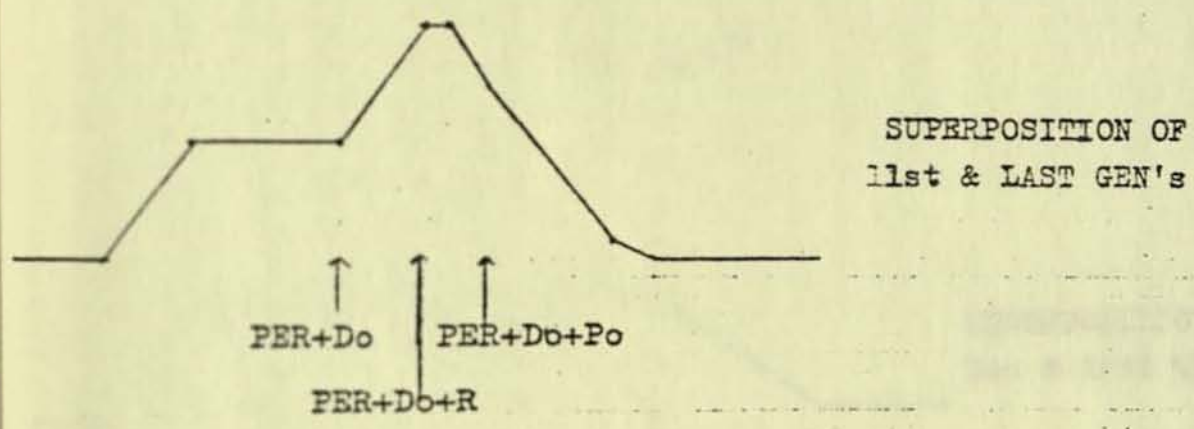
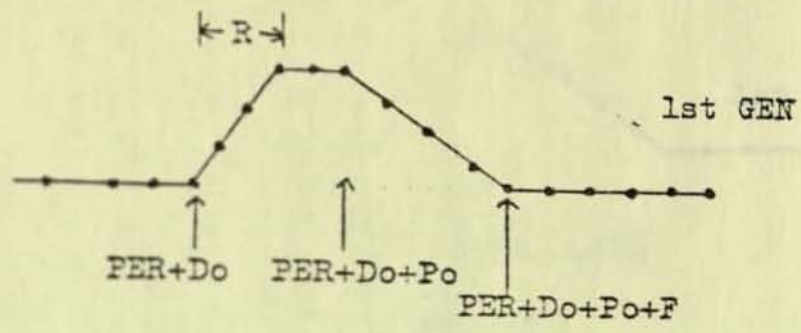
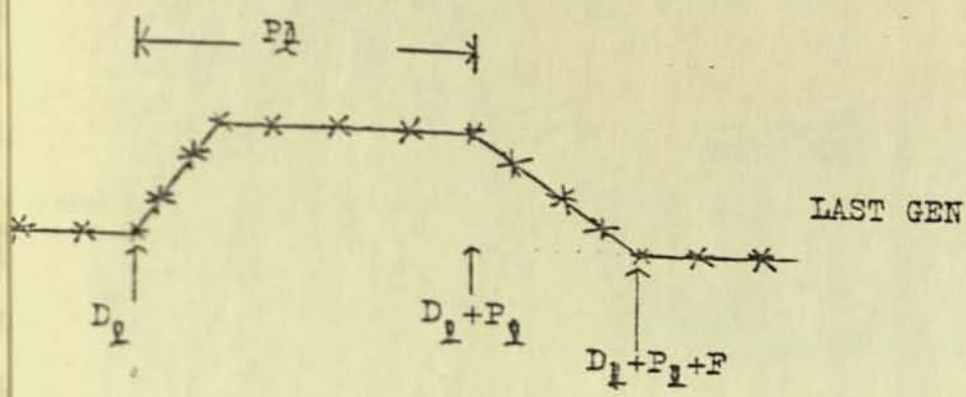
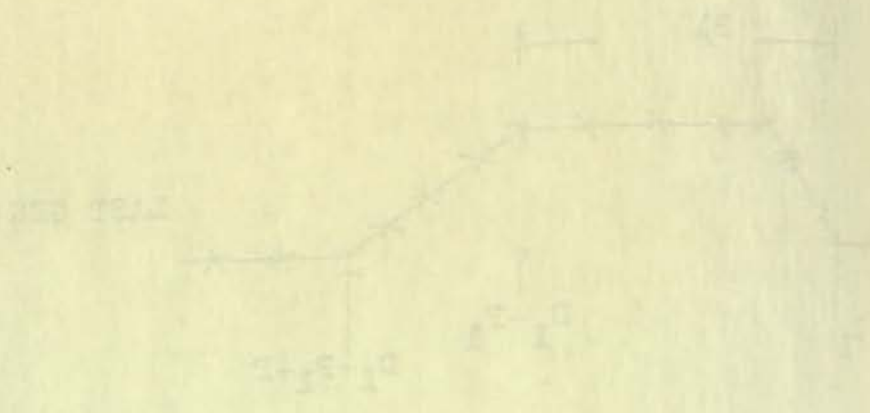
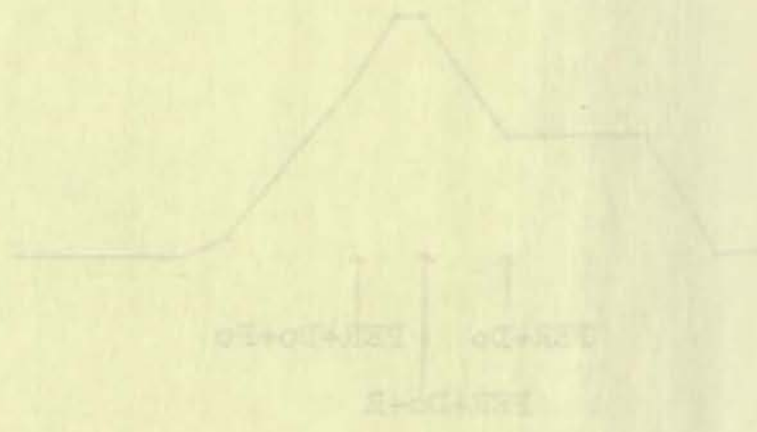


FIGURE 6. OVERLAPPING 1's CORRECTION $PER + Do < D_1 + P_1$

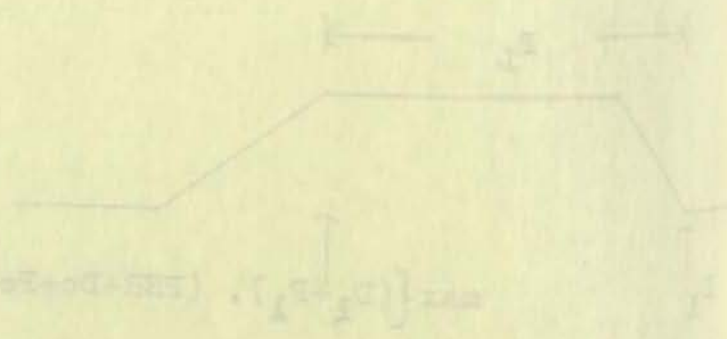


REPRODUCTION OF
LAST GEN'S



COMBINED LAST GEN

$$D_1^2 + 2D_1 + 1 = D_1^2 + 2D_1 + 1$$



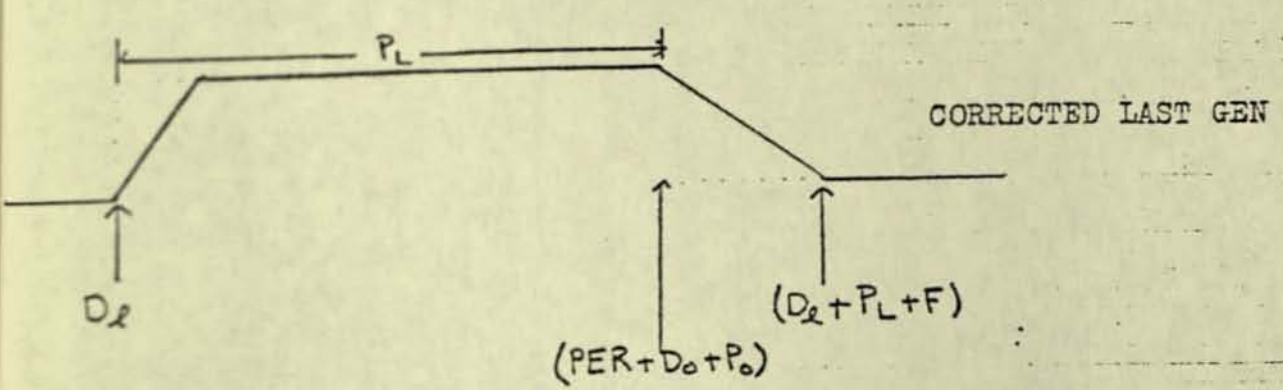
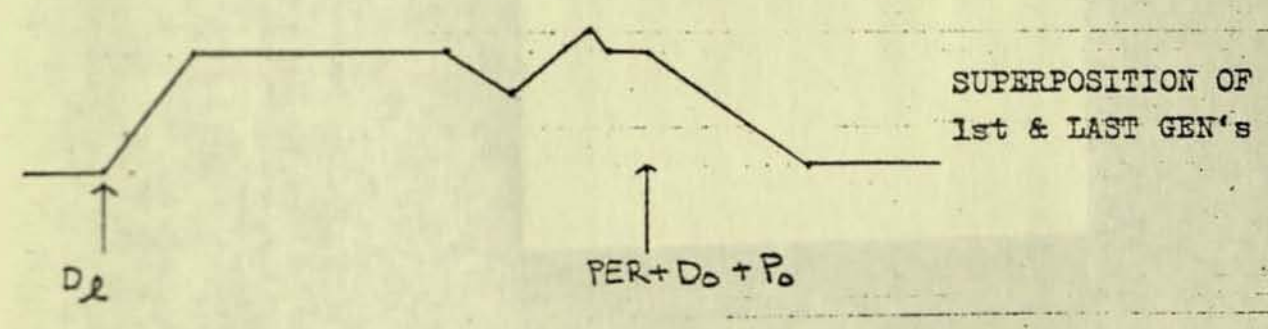
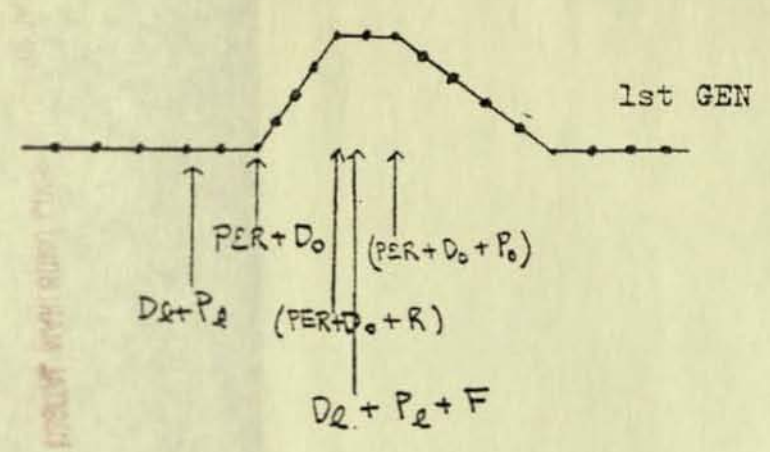
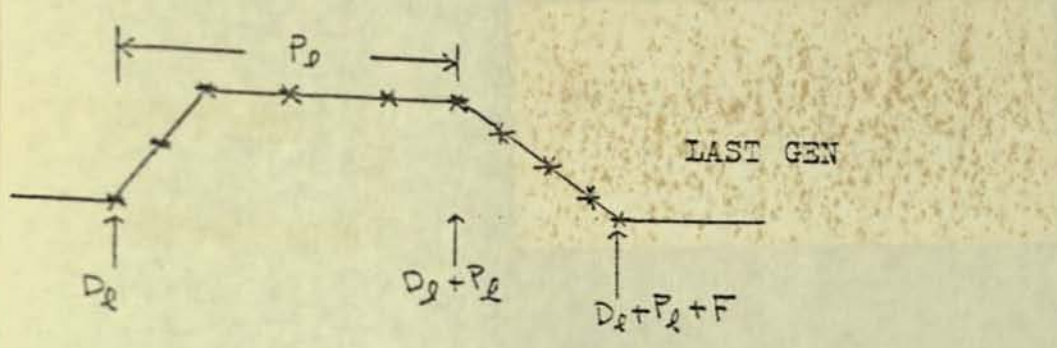


FIGURE 7. OVERLAPPING 1's $(D_L + P_L < PER + D_0 < D_L + P_L + F) \cap (D_L + P_L + F > PER + D_0 + R)$

